# TECHNOLOGY - VIDEO - STEREO

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# May 1992



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# **ON THE COVER**



The digital multimeter is the backbone of the modern test bench, and today's DMM's provide better performance and more features than ever. How do you go about selecting the instrument that best suits your testing needs? Your first step should be to read our Buyer's Guide from start to finish. You'll learn how DMM's work, and how to put them to work for you. Read about the latest models from the major manufacturers. Once you know how they differ-and what those differences mean to your work and to your wallet-you'll be ready to make an informed decision. Turn to page 31 for the details.

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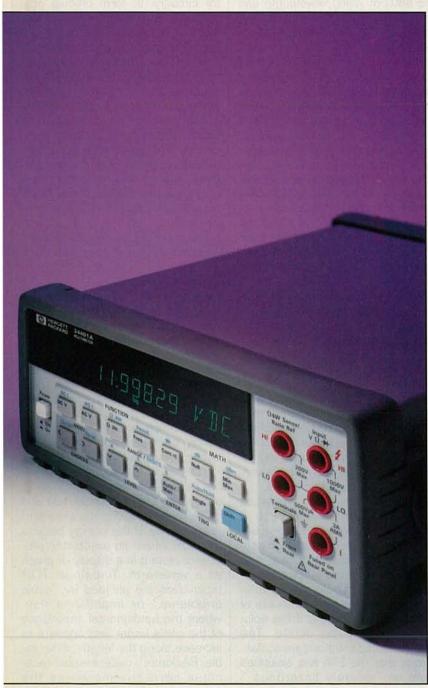
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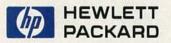
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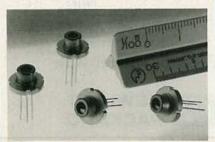
# WHAT'S NEWS

A review of the latest happenings in electronics.

#### High-power semiconductor laser

A series of reliable, 150-mW laser diodes that oscillate at wavelengths between 800 and 870 nanometers has been developed at Sanyo Electric Company's Semiconductor Research Center in Allendale, NJ. The high-powered lasers are expected to enhance processing speeds in erasable optical-disc memories and image-processing equipment and to be used in satellite communication. In addition, they can be used as pumping sources for blue laser-light generation when they are used with a second harmonic generation (SHG) device.

Blue-spectrum lasers with sufficient power are critical to information-intensive applications as longplaying, high-definition, moving-image video storage as well as for fullcolor image processing. An SHG device doubles the frequency of an infrared-spectrum laser when the light from the laser passes through a special crystal in the device, halving the wavelength of the beam and moving it from infrared to visible blue on the color spectrum. A blue laser's beam illuminates about 25% of the area on a receptor surfacequadrupling the recording density on a laser disc and significantly improving resolution in laser imageprocessing applications. The Sanyo development overcomes previous



OSCILLATING AT WAVELENGTHS between 800 and 870 nm, Sanyo's highpower lasers will have applications in optical-disc memories, satellite communications, and blue-laser light generation. obstacles to blue-laser light generation by providing a high output power to compensate for the power loss inherent in the SHG process. The 860-nm lasing wavelength can be produced for phase-matching conditions at room temperature using a typical SHG device.

The laser diodes are fabricated using a relatively uncomplicated two-step, liquid-phase epitaxy process. Adjustments made to laver thickness, lasing cavity length, and the crystalline active layer insure reliable high-power output and suppress temperature rise, which helps prevent degradation or catastrophic damage. Stable continuous wave operation at 150 mW has been confirmed for more than 5000 hours at room temperature, and for over 2000 hours at 50°C. PIN photodiodes for monitoring light-output are built into the assembly.

Three models operating at 800, 830, and 860 nm will initially be produced, with volume production expected by mid-1992.

#### Safer gold electroplating method

A process allowing intricate gold electroplating on microelectronic devices, developed by Researchers at Sandia National Laboratories (Albuquerque, NM), uses a plating solution that is safer than the conventional cyanide-based solutions.

While gold has many properties that make it attractive for use in manufacturing microelectronics high corrosion resistance, high conductivity, high melting point, and ability to form good electrical contacts—standard gold-plating solutions can release large amounts of poisonous cyanide gas if the solution becomes too acidic. The cyanide-based solutions are so dangerous that the EPA has classified them as ''acutely hazardous;'' therefore, they require special safety precautions.

Sandia has applied a gold sul solution, which was developed the 1970's and has been used sir then for protective coverings but for circuitry, to form precise g patterns for semiconductor ( vices. The method has been s cessfully used to plate extrem fine lines on substrates. Sandia searchers have also used the sul solution, which contains no cvan and is not dangerous, to make m ature gold bridges that fo crossovers on gallium arsenide s strates. The crossovers allow c ductors to cross on the surface the IC without touching or short out adjacent conductors. Te: have shown the plating efficiency be close to 100% in such appli tions, and the plated gold's dens approached that of pure gold.

#### Transmission-line impedance measurement

Beckman Industrial has be awarded a patent for developing time-domain reflectometry (TE technique for the accurate misurement of transmission line pedance. The technique has be incorporated in the compan model TMT-1 transmission me tester, which is used to test a certify LAN and telecom phys layer cabling systems.

TDR could be described as " ble radar"-an electrical pulse sent along the LAN under test a cable faults (impedance chang reflect some of the energy back the TDR where an associated r cessor plots it in a display or gra as a waveform. Traditional T techniques are plagued by "ca dribble-up" or impedance ri where the fundamental impedar of the cable under test appears increase along the length of the ble. Beckman's measurement te nique correctly measures 1 impedance along the entire len of cables under test.

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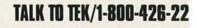
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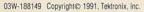


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## VIDEO NEWS

What's new in the fast-changing video industry.

DAVID LACHENBRUCH

• Multimedia. That single word seems to be the hottest topic in both the television and the computer worlds today. With the development of digital video and bandwidth compression, the computer and the TV set are coming increasingly close together, and there are many forecasts that the two will eventually merge. Bets are even being placed as to which will be the survivor.

Of course, that is a ridiculous question, reminiscent of the days when people talked about the computer-controlled home, in which a central computer operated the washing machine, refrigerator, stove, dishwasher, furnace, and everything else. As we all know, that never happened because microprocessors, or mini-computers, became so cheap that they could be incorporated into the appliances themselves.

Since computers and television sets are used in different rooms of the house, the speculation as to which will survive reflects a misunderstanding of the potential of digital technology. We would hope that the computer and the TV set will speak the same language so that their software can be intermixed and combined, but it's ridiculous to think we will attach a keyboard to our TV set of the future to write a nasty letter to the fiber-optic company for overbilling us, while the rest of the family is deprived of watching their favorite sitcom. By the same token, who will want to sit at the 12inch computer monitor to watch the Olympics? Certainly not anyone who has the option of a 100-inch wall-screen (which, incidentally, has no value in word processing).

#### TV-Computer Compati-

**bility.** Capitalizing on the hope that the TV set and the computer will be able to speak with one another without too much translation is Thomson Consumer Electronics (RCA and GE brands in the United States). Thomson says that all of its efforts in advanced TV will be focused on standardization with computer multimedia. Thomson recently won the contract to supply the digital-compression system and consumer receiving boxes and antennas for Hughes Communications' 100-channel high-power direct-to-home DirecTV satellite system, due to start in 1994. Thomson is also participating in a proposed digital HDTV system with Philips, NBC, Sarnoff Research Center, and Compression Labs. Their HDTV system and Thomson's satellite transmission signal use a standard based on the Motion Picture Experts Group (MPEG) digital-compression formula for full-motion video on a CD-ROM disc. Thomson calls its system "MPEG + +," because it has much higher resolution than the current MPEG standard. Actually, the MPEG + + system is one of 32 different proposals for a movie-quality digital-compression system being considered for the title of MPEG 2" standard.

Thomson notes that it expects the MPEG + + system DirecTV to be directly compatible with multimedia systems, computers, video recording systems of the future, and all other encoded digital-compression media. They don't say what will happen if MPEG doesn't choose MPEG + + for MPEG 2.

• How Far Off? Although we're hearing optimistic forecasts that the multimedia age is upon us, one major Japanese manufacturer doesn't see it that way. Toshiba, which, with C. Itoh and Company, paid \$1 billion for part of Time Warner's TV-related business, says its preparing for the multimedia age, but doesn't agree that it's just around the corner. Rather than go off chasing a nonexistent standard, they prefer to wait for the industry to set a standard. Says Kojo Hase, senior manager of Toshiba Media and Communications Group, "Our dream is a five-inch disc with 120 minutes re cording and playback time, erasable and with high-definition quality. A higher density disc and the next MPEG standard for full-motior Imovie quality] video will provide this. But the standards negotiations are only beginning, and there are 32 contenders. There cannot be a stan dard until 1995."

Asked about CD-I, now on the market and hailed by its principle sponsor, Philips, as a success Hase said, "We do not think tha CD-I is fully interactive. The fact tha there are only around 50 titles sug gests that the tooling is wrong, tha the platform is not sufficiently flexi ble. We are studying very carefully and asking what is the right materi al. Certainly not encyclopedias. We need something sparkling tha makes people say I must have it'.

• Small Camcorders Domi nate. Last year, 60% of all camcorders sold in the U.S. were "compacts"—that is, 8mm or VHS C. That is an increase from just ove half in 1990 and only 37% in 1985 The figures are supplied by the Electronic Industries Association (EIA: which doesn't break down the compacts by format, but a good est mate is that 8mm won by a larg margin, comprising 63% of all compacts and 43% of total camcorde sales, compared to 32% full-siz VHS and 25% VHS-C.

• Ghostbusting Winner. As w go to press, the National Associa tion of Broadcasters has declared winner in its field tests of TV ghos canceling systems (**Radio-Elec tronics**, April 1992). It is the sys tem developed by Philips. All five of the proposed systems require that the TV station transmit a pilot sign during the vertical blanking interva and therefore FCC approval is re quired. Action on a standard system is expected as early as June. **R**-



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> ocedure Figure 16 shows the critical diagram for the experiment. The those in membras for the 42 areas included on the diagram. For those, which is the 17, which shows pertinent 94151A data. For the care may result use, an exploreable 10P exists in the configuration, which is 450 are resulted. For the Select and Streebe lines, finality, you'll use the second of the select and streebe lines. for this experiment. You II r

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to the H shipper.

- how ZathbA Fig. 17 Pin Kings

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#### SIMPLE CHARGER

I've been trying to come up with a battery-charger circuit that can maintain an older alarm system. It has two 12-volt 1.2-AH batteries connected in series. I'd like something that charges them in about two hours and then drops the charging current to a trickle charge. I don't need a lot of bells and whistles, but a having a charge/discharge indicator would be nice. Got any tricks up your sleeves that can help me out?—C. Peterson, Van Nuys, CA

It seems that this sort of request shows up every year. Even though rechargeable batteries, of various chemistries are in common use just about everywhere, there's a real shortage of intelligent charging circuits. I ran across the problem a few years ago and designed the circuit in Fig. 1. It has all the features you asked about and can be adapted for use with a wide range of batteries.

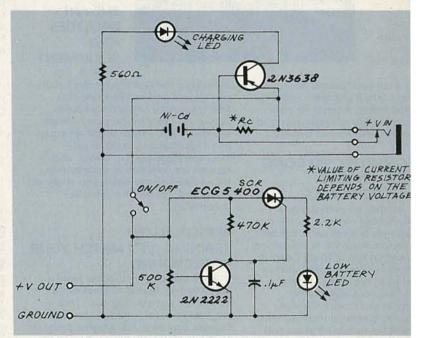
The circuit has two basic parts. The first monitors and controls the charging of the batteries and the second monitors the amount of charge left in the batteries when they are being used to power the circuit.

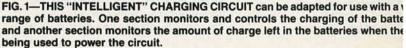
When you're recharging a battery, the most important choice is the value of the current-limiting resistor. Too large a value will result in no charging action, and too small a value will allow so much current to run through the batteries that they'll either be damaged or, if you're not lucky, be destroyed.

The PNP transistor has its baseemitter junction sitting across the current limiter and, when current flows through it, a voltage drop appears across the resistor. That causes current to flow through the collector-emitter junction of the transistor and lights the LED to show that the battery is charging.

As the battery voltage increases, the battery's impedance increases and the current flow gets less and less. That results in a constantly lowering voltage drop across the resistor, and the LED will get dimmer and dimmer until it finally goes out completely. If the intensity of the LED doesn't change, that's giving you some good information as well. Constant high intensity means the batteries aren't taking a charge, and a low intensity (when you first plug in the charger) means the batteries were already charged. low a level determined by the pc tiometer setting, the transiturns off and the battery voltage pears at the collector. That trigg the SCR and lights the LED to v you that the battery has to be charged. The SCR latches and LED stays lit until you turn off power.

The critical element in monitor circuit is the setting of potentiometer, so you should u multiturn unit. The circuit can be





The second part of the circuit monitors the charge on the battery while the batteries are being used. The full battery voltage is put across the potentiometer and a certain value appears at the potentiometer's wiper. The NPN transistor is set up as a switch and, as long as the voltage at its base is high enough to keep it turned on, the collector-base junction conducts and keeps the collector at close to ground level.

When the battery voltage falls be-

ibrated by hooking it up to a vari power supply and setting the age to the value you want for trigger voltage.

The schematic shows the or switch for the circuit being pow and uses the single-pole, sir throw switch in the jack to cha the power source from the batter to the charger. If you trace thro the connections, you'll see that charger recharges the batteries powers the circuit at the same t

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#### DRIVEWAY ALARM

I have a summer cabin that's located at the end of a long driveway and I'm looking for some way to have an alarm that goes off when a vehicle pulls into the driveway. There are animals around so a photoelectric cell isn't practical. Because there are sensing circuits used on highways all the time, there must be a simple and reliable way to get the job done. I've been toying with the idea of an RF-based circuit that has coils under the road and uses the presence of a vehicle to transfer a signal from one coil to another. If you have any ideas, it would help me a great deal.-D. Ingebright, Seattle WA

I have a house in the country and was faced with the same problem a few years ago. I have to admit though, that the idea of solving it by burying coils in the driveway never occurred to me. Maybe that's because the mere thought of the work involved in digging up part of the driveway just seems like too much effort. The method I used is a lot easier, much more straightforward, and can be done without having to use a shovel.

I contacted the local road maintenance people, and bought a length of the same sort of wire they string across the road when they want to do a traffic count. It's just two lengths of wire, separated by foam, and sealed in a tough rubber overcoat. When a car goes across it, the weight compresses the foam and the two wires inside the cable touch each other—a simple on/off switch. I don't know who makes the stuff, but I'd be willing to bet you can get a length of that wire the same way I did.

Some of the road departments use a different detection method in which the cable they lay across the road is really nothing more than a hollow tube filled with air. When a car crosses over it, the increase in pressure forces an air switch to close. Different method, same idea. It doesn't matter which of the two materials you get since either of them will do the job for you—although an air switch may not be such an easy part to get a hold of. The signal from the wire acm my driveway triggers a simple circuit and that, in turn, sound bell in the house. The whole syst took about an hour to set up and been working reliably ever sin then.

#### CALL-WAITING DILEMMA

My wife is using a compu and modem on her job, a she's having a problem with c waiting on the line. Whenever incoming call shows up on line, the computer locks up a to make matters worse, the coming call is lost. Is there a thing that can be done abo that without doing away with call-waiting service co pletely?—T. Edmonson, Bir ingham, AL

This is a fairly common probl and, if you can't afford to hav separate data line, it can be a pain in the neck when you're usir modem. It's bad enough losing call, but imagine what it's like wl call waiting causes the carrier drop just at the tail end of downlo ing a huge file. It's practica enough to make you start using U.S. mail again!

There are ways around the pr lem without discontinuing call w ing and, believe it or not, the answ is really simple. All that's needed bit of software that's so minor y can even write it yourself us BASIC or whatever language prefer.

The basic approach is to te porarily change the values in a of the modem's control registe The exact method is going to v with different modems, but basic principle will be the sar Just about all popular modems h one register that controls amount of time the modem will y to disconnect, once carrier l been lost. In the Hayes series modems it's the S10 register the time can be set in tenth-c second increments. I'm not fam with the modem you have, but sure that if you go through the m ual (or call the manufacturer), yo be able to get the information need to make the modification. I ran across a small public-don continued on page

14

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## LETTERS

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#### **KEEP IT SIMPLE**

Regarding the article "Energy Consumption Monitor" (**Radio-Electronics**, December 1991): Sometimes we get so engrossed with today's technology that we lose sight of the simple way to do things. Although certainly not as elegant or convenient, the same thing can be done using a surplus watthour meter, which can be purchased at a flea market for a few dollars. (It's the same sort as the electric company uses at your house to bill you.)

The formula for finding power consumed at any given time is:

Watts = rev.  $\times$  kWh  $\times$  360/sec (The kWh rating can be found on the name plate.) You can count the number of revolutions by the black spot on the disk. The meter can be left in the circuit for any length of time—a week, a month, or whatever. You can read the consumption in kWh and use your electric company's rate to arrive at a monetary value.

JOHN L. KURSCHNER Toms River, NJ

#### MOTOR-SPEED CONTROLLER CORRECTION

I have found the motor-speed controller circuit described in *Ask R-E* (**Radio-Electronics**, February, 1992) to be very handy. However, there was a schematic error showing the oscillator circuit to be grounded. I've enclosed a drawing





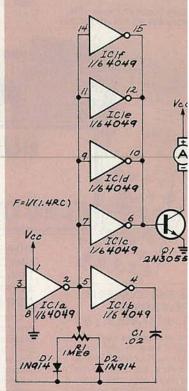


Fig. 1. This is the corrected schema for the Motor-Speed Controller that a peared in February's Ask R-E.

of the corrected circuit (Fig. 1). CALVIN D. KAUFMAN *Highland, MI* 

Thanks to you and all the oth who pointed out our error.—Ed.

#### FLIGHTS OF FANCY

For the last few years I have re **Radio-Electronics** without mu interest in the "Letters" section, I over the past few issues I've reac I noted a letter in the February 19 issue that is very wrong. It concer the December 1992 *Drawing Boa* column, written by Robe Grossblatt.

In the letter, the writer states the "high-speed aircraft all used ger ators and the electromechani voltage regulators they required have spent the last 20 years as Air Force airborne radar navigat technician working on KC-135

Radio-Electronics, May 1992

B-52's, F-5/T-38's, U-2/TR-1's, and even the fastest of them all, the SR-71. We had engine-driven devices that were called generators on board, but all of them were 240volt AC, 400-Hz, 3-phase machines. Voltage regulation was electronically controlled, and charging power for the aircraft battery was provided by a 28-volt DC transformer rectifier that had its own internal electronic controller. The oldstyle generator and the associated inverters that the writer described left the Air Force inventory with the old T-28 and its sister aircraft, which were piston-engine powered. All modern jet-powered aircraft have gone over to the alternator type of power generator, but the old name has stuck nonetheless.

RICHARD J. GOULET, MSgt, USAF

Wright-Patterson AFB, OH

#### **ASK R-E ANSWERS**

I read about the "Commercial Limiter" in Ask R-E (**Radio-Electronics**, January 1992) with great interest. I have some additional information that I think would be helpful to anyone undertaking such a project.

The article indicates that the audio signal can be obtained at the volume control or the speaker. While the latter is always true, the former might not be.

Many recent model TV's using conventional (potentiometer) volume controls use electronic volume attenuators that are generally incorporated into the sound IC. Those IC's are also used in sets with pushbutton and remote volume controls. In this type of system, no audio will be found at the volume control.

One reason for adopting this method is that it eliminates hum pick-up, which results from routing the audio signal to and from the volume control.

If a set uses an electronic attenuator and if the concept of the Commercial Limiter is understood, the circuit could be modified to take control of the attenuator line that controls the volume. That would preclude the necessity of using an additional amplifier stage, which could degrade the audio, particularly if the set has a good audio section like those found in stereo TV's.

In some of these sets, an increase in control voltage causes a decrease in volume, while in others it is just the opposite. Depending on the system, the photo resistor could be placed so that it will shunt the attenuator line to ground, pull it up to some positive value, or it could be placed in series with the line. The LED could be driven by a simple circuit that would cause it to light when the volume exceeds some value. An LM3915 (such as the one shown in the same Ask R-E column in the "Audio Light" circuit) would be ideal for that purpose.

If the limiter's input comes from the output of the power amp or any point after the volume control or attenuator, it will have to be readjusted each time the volume setting is changed. If the input comes from some point ahead of the volume control or attenuator, then the limiter's action will be independent of the volume setting.

In many of the sound chips that use electronic attenuators, there is a pin that is connected tot he output of the FM detector. From there, the audio is sent to the attenuator. That would be a good take-off point. In sets that do have audio at the volume control, audio can be taken from the "hot" side of the control and the photo resistor can be connected between the wiper and ground. All of those points can be easily found if the service literature is available.

By the way, I'd like to point out that there might be a problem with the circuit diagram for the Audio Limiter, shown on page 82. The PNP transistor must have a base voltage of about -0.6V with respect to the emitter for turn-on. Since the base is tied to V+, it will always be positive with respect to the emitter. Under these conditions it will never turn on. Perhaps it would if the emitter and the base were reversed.

The writer also indicated that the photo detector is wired in the feedback loop where a change in resistance will change the gain of the amp. Actually, the LM386 is operating in a fixed gain mode. The photo continued on page 22 Test Fine-Pitch Devices to 208 Pins

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#### LETTERS

continued from page 17

detector is wired across the source (the secondary windi the transformer) and not in a back loop. A change in resis acts to reduce the audio lev shunting it to ground before ap tion to the "gain adjust," wh actually a level-adjust control doesn't involve the same pri as adjusting the gain of a stag

The "Commercial Limiter the "Audio Link" illustrate an i tant concept. Projects suc these help us to understand individual building blocks can I sembled to perform a task tribute a large portion c knowledge of electronics t projects and articles found in **dio-Electronics**. Keep u good work.

STEVE BABBERT Worthington, OH

#### ASK R-E

continued from page 14

program that addresses this lem and I've put it on the RE (516-293-2283) for you to ( load. The file is called ( WAIT.ZIP. It has a small B program that was designed Hayes 1200-baud modem, should show you what has done and how to do it. The pro is well-commented and all really have to do to make it with your modem is change numbers.

It really works. I've tried i modems from Racal Vadic Robotics, and Hayes (after su patching), and never had a prc If you find that it doesn't wor best advice I can give you is to tact the modem manufacture let them come up with an ar Call waiting is a common se and you can be sure that you' the only one who's had this lem. If that fails, you'll just ha spring for a second phone | drop call waiting-whichever od you prefer. Your phone cor should have a way to disabl waiting temporarily as well.

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V-212 - 20MHz Scope \$425

Hitachi Portable Scopes

DC to 50MHz, 2-Channel, DC offset func-

tion. Alternate magnifier function

V-525 - CRT Readout, Cursor Meas. \$995

\$975

\$875

\$775

\$695

\$625

V-523 - Delayed Sweep

V-223 - 20MHz delayed sweep

V-522 - Basic Model

V-222 - 20MHz deluxe

V-422 - 40MHz

VC-6023 - 20MHz, 20MS/s

VC-6024 - 50MHz, 20MS/s

VC-6045 - 100MHz, 40MS/s

VC-6145 - 100MHz, 100MS/s

comfort of analog and the power fo digital.

RSO's from Hitachi feature roll mode, averaging, save

memory, smoothing, interpolation, pretriggering, cursor

measurements. These scopes enable more accurate,

simplier observation of complex waveforms, in addition to

such functions as hardcopy via a plotter interface and waveform transfer via the RS-232C interface. Enjoy the

25MHz Elenco Oscilloscope

VC-6025 - 50MHz, 20MS/s

\$1,695

\$1.995

\$2,195

\$2,995

\$4,495

This series provides many new functions such as CRT Readout, Cursor measurements (V-1085/1065A/665A), Frequency Ctr. (V-1085), Sweeptime Autoranging, Delayed sweep and Tripper Lock using a 6-inch CRT.You don't feel the compactness in terms of performance and operation.

V-660 - 60MHz, Dual Trace	\$1,149
V-665A - 60MHz, DT, w/cursor	\$1,345
V-1060 - 100MHz, Dual Trace	\$1,395
V-1065A - 100MHz, DT, w/cursor	\$1,649
V-1085 - 100MHz, QT, w/cursor	\$1,995
V-1100A - 100MHz, Quad Trace	\$2,195
V-1150 - 150MHz, Quad Trace	\$2,695

#### Elenco 40MHz Dual Trace



CIRCLE 109 ON FREE INFORMATION CARD

## NEW PRODUCTS

Use the Free Information Card for more details on these products.

#### VIDEO FRAME GRABBER.

You can take a video image and incorporate it in a variety of word-processing and desktop publishing applications-creating a customized letterhead including your photograph, for example-using MacroHard's Video Frame Grabber package. With the help of a VCR or camcorder, the Video Frame Grabber will digitize a video signal, which you can then modify PageMaker are provided. to suit your needs and save



**CIRCLE 16 ON FREE INFORMATION CARD** 

The package, which conit in any of five different for- sists of the Video Frame mats for export to various Grabber board, a 51/4-inch third-party software. The floppy disk, and a manual, images can be scaled to allows you to capture realsize and positioned any- time video at up to two where in your document, frames per second. Pause and formats for popular and freeze modes are availprograms including Micro- able for viewing and modisoft Paint and Windows fying the image. The PBrush, Corel Draw, Ven- package features scalable tura Publisher, and Aldus printing, user-defined mar-

gins, auto-exposure compensation, and anti-aliasing filtering for lower noise and less snow

The Video Frame Grabber, which carries a oneyear warranty, has a suggested retail price of \$149.95.—MacroHard. 755 Tatum Street, Woodbury, NJ 08096-3431; Phone: 609-853-4680: Fax: 609-853-0677.

**RF DETECTOR/FREQUENCY** COUNTERS. Two new pocket-sized counters from Startek each feature a 2-inch, 10-segment LED signal-strength bar graph that functions independently of the digital frequency counter. The bar graph, which can be used to locate or adjust an RF signal, will indicate the relative strength of an input signal at any frequency from 500 kHz to 3.5 GHz. The user can switch between dot-graph and bar-graph operation, and the sensitivity is adjustable.

The frequency-counter function has a range of 1 MHz to 1.5 GHz on the model 15-BG and 1 MHz to 3.2 GHz on the model 35-



**CIRCLE 17 ON FREE INFORMATION CARD** 

BG. Other than the frequency bandwidth, the only difference between the two models is that the 35-BG is more sensitive above 500 MHz. Both counters have a display-hold switch with indicator and three switchselectable gate times. Resolution is 1 kHz at 0.25 second, 100 Hz at 2.5 seconds, and 10 Hz at 25 seconds, over the entire range. The display consists of eight red LED digits. A 1-PPM TCXO time base is standard, and there are provisions for an optional, ultra-high-stability TCXO. With the Ni-Cd battery pack fully charged, the counter will operate for 3-5 hours. An AC adaptor/ charger is standard.

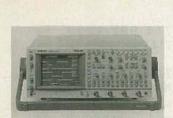
Models 15-BG and 35-BG bar-graph/frequency counters cost \$220 and \$265, respectively.-Startek International Inc., 398 NE 38th Street, Fort Lauderdale, FL 33334; Phone: 305-561-221 or 80-638-8050 for orders only; Fax: 305-561-9133.

ANALOG/DIGITAL STORAG SCOPES. A series of 10 and 200-MHz analog/div ital oscilloscopes fro Fluke and Philips combine analog familiarity and digit performance. It offers tru four-channel operation a well as models with 2+ input-channel configur. tions. Additional feature include direct-access co trols, closed-case calibr tion, serial interface for printer/plotter, full pre grammability, automat pass/fail testing, sign math and analysis, sign averaging, and peak-dete mode for HDTV tri-lev sync. A touch-hold-an measure feature is used f initiating measuremen from a probe-mounted bu ton. The units have a wic range of applications in r dio/TV, medical ele tronics, and the new 110 line resolution video.

The four Fluke/Philip oscilloscopes-PM 338 PM 3384, PM 3392, ar PM 3394-can each t considered two scopes one, with the easy oper tion and realtime 100-200-MHz bandwidth of a analog scope, and digit sampling rates of 200 m lion samples per secon According to Fluke, th analog portion is not redu dant; there continue to I many applications that d mand the "live" signal re resentation and infini resolution of analog instr ments. Switching betwee analog and digital oper tion is simple. The uni have full digital capture a signal analysis capabilitie providing powerful stora and measurement c

Radio-Electronics. May 1992

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**CIRCLE 18 ON FREE** INFORMATION CARD

pabilities on demand. Instant results are obtained by selecting the required function and channel from a set of built-in, fully automated voltage and timing measurement functions.

Triggering modes include logic state and pattern triggering, and glitch triggering. In addition to pre- and post-trigger signal display, the scopes have event delay and time-afterevent delay.

Models PM 3384 (100 MHz) and PM 3394 (200 MHz) offer true four-channel operation, which provides full sensitivity with complete attenuation ranges on each channel. The 100-MHz PM 3382 and the 200-MHz PM 3392 offer 2+2 channel operation, a configuration that provides a cost-effective alternative to four fully featured inputs.

Models PM 3382, PM 3384. PM 3392, and PM 3394 have list prices of \$4490, \$5490, \$5990, and \$6490, respectively .-John Fluke Mfg. Co., Inc., P.O. Box 9090, Everett, WA 98206; Phone: 800-44-FLUKE.

SELF-TUNING TEST RE-**CEIVER.** For testing the modulation quality of mobile voice transmitters and verifying the accuracy of audio signalling tones such as DTMF and CTCSS, Optoelectronics' R-10 communications test receiver is a quick, easy, and inexpensive alternative to costly modulation meters. The unit can automatically accommodate

the enormous difference ratio between an unknown carrier of up to 1000 MHz and a modulating frequency as low as 50 Hz, without any manual tuning.

The R-10 automatically determines and locks onto whatever carrier signal is the strongest within a 100foot range. Once the signal is locked in, the R-10 demodulates whatever FM audio is present on the carrier and outputs that audio to an internal power amplifier/speaker and the BNC output jack. For DTMF and CTCSS testing, the audio output is evaluated on a frequency counter or oscilloscope. For countersurveillance operations. the R-10's demodulated audio can be monitored to determine if a surreptitious transmitter is in the area.

MULTI-PUBPOSE INTER-FERENCE FILTER. Ace

Communications' MPIF-1 receiver filter eliminates a variety of unwanted signals. The compact  $(3 \times 2 \times$ 11/2-inch) external filter eliminates most of the sources of interference common in broad-banded receivers. Unwanted signals are filtered from the 54-108 MHz, 174-220 MHz, and 512-806 MHz ranges, as well as the range above 869 MHz. A switchable notch will also eliminate interference on the 150-153 MHz range, which is a common source of interference in many areas. The use of BNC connectors makes the filter quite versatile; the MPIF-1 can even be used on handheld receivers.

The MPIF-1 multi-purpose interference filter has



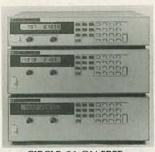
**CIRCLE 20 ON FREE** INFORMATION CARD

a suggested retail price of \$59.—Ace Communications. Monitor Division, 10707 East 106th Street. Fishers, IN 46038; Phone: 817-842-7115: Fax: 317-849-8794.

POWER SUPPLIES. The "smart" front panels on the 15 instruments in Hewlett-Packard's new line of DC



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CIRCLE 21 ON FREE INFORMATION CARD

power supplies help provide fast, flexible, and precise output control. With outputs ranging from 200 to 2000 watts, the HP 6000 series power supplies provide a variety of choices for benchtop or system applications that do not require HP-IB control of the power supply. The series consists of five power supplies (which vary by voltage ratings) in each of three power ranges: 200, 500, and 2000 watts.

All models have low output noise, extensive loadprotection features, and the "smart" front panel. The peak-to-peak ripple and noise spans from a low of 3 mV p-p on the lowpower, low-voltage models to a high of just 16 mV on the 2-kW, 120-volt model. Overcurrent, overvoltage, and over-temperature protection are provided for the device under test and for the power supply. Those features protect the device under test by disabling the output voltage of the power supply when potentially dangerous conditions occur. "Smart" front-panel controls provide three methods for setting the output voltage and current. A numeric-entry keypad allows the user to set the voltage quickly and precisely, while up/down buttons and rotary pulse generators permit the user to quickly and conveniently change the voltage and current settings in small incre-

26 ments. For repetitive

Radio-Electronics, May 1992

benchtop tests, up to five states or sets of powersupply settings can be stored and recalled for easy sequencing among states. Front-panel controls also allow the user to calibrate the power supply. In addition, the output voltage can be controlled via an external voltage signal, allowing for computer control or analog modulation.

List prices for the *HP* 6500 series of power supplies range from \$1650 to \$1750 for the 200-watt models, \$2100 to \$2300 for the 500-watt models, and \$3650 to \$3800 for the 2000-watt models.— **Hewlett-Packard Company**, Inquiries, 19210 Pruneridge Avenue, Cupertino, CA 95014; Phone: 800-752-0900.

HAND-HELD DIGITAL MUL-

TIMETERS. Aimed squarely the field service technician, Fieldpiece's digital multimeter models HB75 and HB77 are professionalgrade instruments that are durable and easy to use in the field. They include a built-in logic probe and a variable-pitch tone. The tone's pitch varies proportionally from high to low with high to low readings. In the field, that feature has two primary uses. First, intermittents can easily be found by listening for discontinuities in the tone while wiggling suspect connections. Second, one "odd" test point in a series



CIRCLE 22 ON FREE INFORMATION CARD

of test points that are nearly the same can be found without waiting for the meter to display a number; instead, the technician can quickly scan the test points and "hear" the one that is different.

Both models have a builtin logic probe that responds up to 20 MHz. "Hi" and "lo" are indicated both in the display (with up/ down arrows) and by a beeper with two different tones. Both DMM's also have a built-in capacitance meter that measures capacitors up to 200  $\mu$ F in the circuit using the test leads.

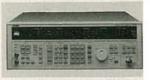
Model *HB77* measures the true-RMS value of AC voltages and currents. Current ranges go from a low of 200  $\mu$ A to 20 A. Because the meter is manual ranging, the voltage burden (the voltage across the device when current is flowing through it) is low-0.25 volts for the 200- $\mu$ A range.

Both heavy-duty meters feature a drop-resistant housing, O-ring seals to protect against contaminants, MOV's to protect against transients, and full 600-V fusing on all current jacks. They come with a tilt stand and a hanger on the back, test leads, and fuses and batteries installed. Both DMM's have 24 ranges in AC and DC volts, AC and DC amps, and ohms, and both include a high-voltage indicator that warns the user when touching anything over 28 volts. A single rotary dial, with the "menu" of functions printed around it, makes the meters easy to understand and use.

The models *HB75* and *HB77* digital multimeters have suggested list prices of \$139 and \$179, respectively.—**Fieldpiece Instruments, Inc.**, 8322B Artesia Blvd., Buena Park, CA 90621; Phone:

714-992-1239; Fax 714-992-1239.

2.7-GHz SYNTHESIZED SIG NAL GENERATOR. De signed for use in researc and development, mar ufacturing, and servicin electronic products, Leac er's model 3221 syr thesized signal generatc



CIRCLE 23 ON FREE INFORMATION CARD

offers high-stability an high-purity outputs to 2. GHz. Its extensive modula tion capabilities includ seven modes with 14 simu taneous combination mod ulation modes-pulse, loc ic, DC-FM, and internal c external AM and FM. RI output level engineerin units are selectable be tween dBm and dBµ wit 0.1-dB resolution. Thre convenient presets are in cluded for commonly use RF-output levels. One hur dred preset memories a low storage and recall of a front-panel setting cond tions. RF output on/of cor trol and 50-watt revers power-protection reset ar accomplished with a sing key. A continuous, variab RF-output mode allow ±5-dB variation in 0.1-d increments for squelch a justments. Edit functior for frequency, output leve and modulation make easy to change operatir parameters. Other star dard features include GPI and a GaAs FET puls modulator.

The model 3221 syi thesized signal generato costs \$12,300.—Leadi Instruments Corporatio 380 Oser Avenue, Hau pauge, NY 11788; Phon 800-645-5104 c 516-231-6900.





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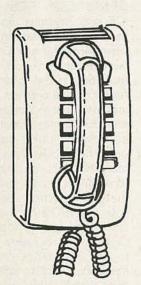
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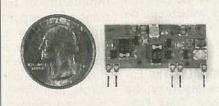




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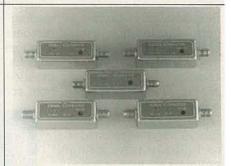
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## NEW LIT

Use The Free Information Card for fast response.

RADIO FREQUENCY INTER-FERENCE: How to Find it and Fix it; book edited by Ed Hare, KA1CV, and Robert Schetgen, KU7G; The American Radio Relay League; 225 Main Street, Newington, CT 06111; \$15.00 plus \$3.00 shipping and handling (\$4.00 for UPS).

If you've ever experienced black bars flashing across your TV picture in a rhythmic pattern, a garage door opening or closing by itself, a buzz that drowns out AM stations, a touchcontrolled lamp with a mind



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of its own, or anonymous conversations that interrupt music on your stereo you've experienced radiofrequency or electromagnetic interference (RFI/ EMI). This 256-page book not only explains the mechanics of RFI/EMI, but also provides practical cures for the problem. Opening chapters offer general information on how to find the interacting equipment, locate help, and resolve conflicts. Subsequent chapters discuss RFI/EMI problems and provide cures for specific electronic systems, including transmitters, televisions, telephones, amateur-radio equipment, stereos and other audio gear, power lines and electrical devices. computers, and automobiles. The chapter on filter performance explains how to select a filter and provides test results and performance tables for dozens of low-pass, highpass, power-line, and miscellaneous filters. In addition, the book also explains **RFI/EMI** regulations and standards, and provides a copy of the ARRL EMI/RFI report form, which can be used to file official EMI/RFI complaints.

SECRET SIGNALS: The Euronumbers Mystery; by Simon Mason. Tiare Publications, P.O. Box 493, Lake Geneva, WI 53147; Phone: 414-248-4845; \$9.95 plus \$2.00 shipping and handling (\$3.00 outside the U.S.).

If you've ever tuned to the frequencies outside the shortwave and ham bands to the regions allotted to "fixed" stations, you might have heard broadcasts consisting of a single voice reading out long lists of numbers in four- or five-digit groups. While explanations for those mysterious broadcasts have ranged from commodity prices to



CIRCLE 39 ON FREE INFORMATION CARD

lottery numbers, the most likely explanation is that they are coded messages sent to espionage agents. The end of the Cold War hasn't ended the transmission of coded numbers: they're still being sent, day and night, over all the shortwave-radio bands. This book, written by a man who has studied the European numbers stations for years and has monitored thousands of transmissions in the process, is an indispensable tool for tracking those stations from North America. It includes numerous traffic excerpts, identifiers, schedules, and clues turned up by hearing mistakes in transmissions. Descriptions of transmissions from "Bulgarian Betty," "Papa November," "The Russian Man," "Swedish Rhap-sody," and "The Lincolnshire Poacher" are included, as is a complete frequency log with more than 300 entries that are listed by frequency and contain notes on formats and schedules.

THE HARD DRIVE ENCYCLO-PEDIA: The Guide to PC-Compatible Hard Drives; by Adrian Alting-Mees. Annabooks, 12145 Alta Carmel Court, Suite 250-262, San Diego, CA 92128; Phone: 800-462-1042 or 619-271-9526; Fax: 619-592-0061; \$89.00.

Consisting of more than 600 pages in a three-ring binder and a companion diskette of utilities, this book provides a complete reference on PC-compatible hard-disk drives. The



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book includes sections ST-506, ESDI, SCSI, a IDE specifications, as v as a section covering f other interfaces. Oth sections cover the physi and electrical chara teristics of hard drives, lo ical encoding scheme and file formats. Sectic covering controller para eters, hard-disk drives, a manufacturers include i tensive lists of related inf mation. Another secti lists the BIOS hard-dr tables for many popu BIOS's, and explains he to use the utilities on t companion disk to see t drive-type tables in yc own BIOS. More than 16 hard-drive model numbe are listed by manufactur so that if you have to ins a drive but don't have spec sheet, you can loca the important paramete in the included tables.

1992 ELECTRONIC TEST / CESSORIES CATALOG; fr( ITT Pomona, 1500 Ea Ninth Street, P.O. Box 27( Pomona, CA 91769-27( Phone: 714-469-2900; Fa 714-629-3317; free.

Specially featured in tl 140-page catalog are expanded line of oscil scope-probe kits and tv

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new lines of test probes and clips for miniature and heavy-duty applications. Also featured are accessories designed to make testing SMT devices and highdensity leaded components easier and more reliable. New products include new IC clip kits, coax/BNC universal adapter kits, digital multimeter test-lead kits, cable and patch accessories, and jumper kits. ITT Pomona's popular selection of jumpers and cables, boxes,

plugs and jacks, connectors, adapters, singlepoint test clips, and static control devices are also described. An easy-to-use index is provided to help readers quickly locate specific products.

LENK'S LASER HANDBOOK: Featuring CD, CDV, and CD-ROM Technology; by John D. Lenk. McGraw-Hill, Inc., Professional Book Group, 11 West 19th Street, New York, NY 10011; Tel. 1-800-2-MCGRAW; \$39.95.

This new addition to McGraw-Hill's Consumer Electronics Series is a practical reference book filled with the information needed to troubleshoot today's laser-based products. Aimed at service technicians and field-service engineers who work with laser-based technology,



CIRCLE 35 ON FREE INFORMATION CARD

the book focuses on videodisc and compact-disc players, and CD-ROM units, using information that is applicable to all sorts of laser devices, including CD-I (compactdisc interactive) units. The book shows precisely how to repair laser-based equipment and how to pinpoint trouble in a component or module. Step-by-step and circuit-by-circuit examples are used to explain not only how laser equipment

works, but also how to service it.

Opening with a discussion of the basics of CD players, the book goes on to explain various techniques for encoding and decoding as well as the basic principles of optical readout. Proper operation and installation are described, as are the test equipment and tools required for servicing and maintenance.

Typical circuits for compact-disc and videodisc players are explored, accompanied by schematics, block diagrams, and a discussion of the theory of operation. The book also discusses approaches to troubleshooting and adjusting CD and laser disc products based on the type of failure or symptoms of trouble. **R-E** 







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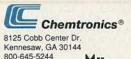
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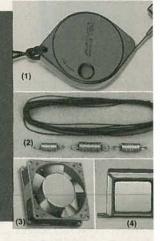
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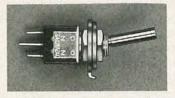


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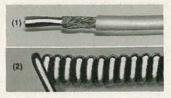
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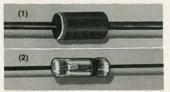


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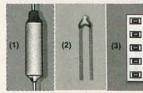
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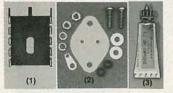
 #270-1320. 228° C., #270-1322

 (2) Thermistor. Resistance is protional to temperature. #271-110

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 pieces. 15 popular values! Rated 1%

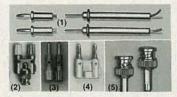
 5%. #271-313
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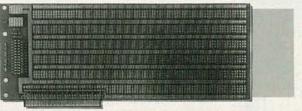
Connector. Ideal for use with F

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# BUYER'S GUIDE TO DMM'S

Today's bench DMM gives you more performance for your instrument dollar than ever before.

#### **STAN PRENTISS**

THE BENCHTOP DIGITAL MULTIMETER (DMM) IS THE workhorse test instrument in today's labs and sevice shops while the handheld DMM is the favored vertatile tester in the field. Bench/portable DMM's have seized the high ground where emphasis is on accuracy, resolution and automated testing. Benchtop DMM's are in use for electronic equipment testing, service, and calibration in both shop and field. They are also filling important roles in design departments, scientific labs, and as components in industrial data acquisition and automated test systems.

To be classed as a DMM, a meter is usually expected to be able to make the five basic electrical measurements: DC and AC volts, DC and AC current, and resistance. However, the term DMM now covers a wide range of instruments from the low-cost handheld, battery-powered units costing less than \$100 to the AC line-powered bench/portable units whose base list prices can range from \$200 to nearly \$2700.

Benchtop DMM's are distinguished from handhelds by their rectangular cases and front-face displays and controls. Their long depth dimensions give them stability on a bench or other flat surface. Most are equipped with tilt bails to raise their front faces for easier user reading. Despite those differences, there is considerable overlap in features between high-end handheld and low-end benchtop DMMs.

In fact, some bench-type meters are little more than repackaged handhelds, and they cannot be distinguished by looking only at specifications. All handheld DMM's are portable, but benchtop DMM's are also portable, and those that are battery powered can be used conveniently in the field.

Many bench DMM's have liquid-crystal displays, typical of today's handhelds, but most of the highend models have vacuum-fluorescent or LED displays that afford better viewing in subdued light. These displays can be used because the benchtop units have less restricted power budgets than the handhelds.

#### How they work

Figure 1 is a simplified functional block diagram of a DMM. In practice, both AC and DC voltage measurements are made by one circuit, and both AC and DC currents are measured by another. Typically there is a separate resistance measuring circuit.

The latest DMM's, both handheld and bench, include either a microcontroller (microcomputer-on-a-chip) or microprocessor for various control and selfcheck functions. A microprocessing function block has been omitted from the simplified diagram, but it would typically be located between the analog-to-digital (A/D) converter and the display.

The microprocessor or microcontroller provides control signals for the true root-meansquare (RMS) voltage and current converter if the DMM has one. Bench meters are more likely to have this feature than handhelds; thus it is not exclusive to either one. However, provision for interfacing with systems is an exclusive bench DMM feature.

Bench DMM manufacturers have used a number of different schemes to convert analog input signals into a digital readout. Among them are voltage-tofrequency, successive approximation, reciprocating remainder, and dual-slope integration conversion. However, the most popular scheme in use today is some form of dual-slope integration. Figure 2 is a simplified block diagram of a dual-slope integrating DMM. Instead of converting voltage to frequency as is done in other methods, it is converted to time. The timing sequence for this technique is shown in Fig. 3. Switch S1 connects the unknown input voltage to the integrator consisting of input resistor R1 and an operational amplifier with capacitor C1 in its feedback loop.

Switch S1 remains in that position for the integration period. During this time, C1 is charged at a rate determined by R1. At the end of the integrat period, C1 has a charge that proportional to the input v age. The op-amp causes i voltage across C1 to build 1 early so the charging rate is g erned by the current throu resistor R1.

In the second phase of du slope operation, control lo switches S1 to connect a re ence voltage to R1 and the inj of the integrator. The referen polarity is always opposite that of the unknown input v age. Two references are used one negative and the oth

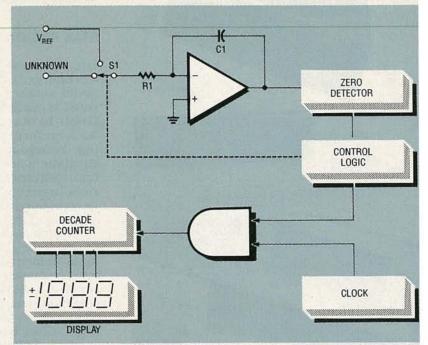
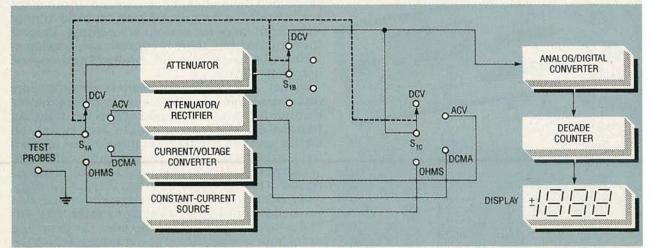


FIG. 2—SIMPLIFIED BLOCK DIAGRAM of a dual-slope integrating DMM. Switch connects the unknown voltage to the integrator for the first half of the cycle, and reference voltage for the second half. The time to discharge capacitor C1 is conver to a digital readout.



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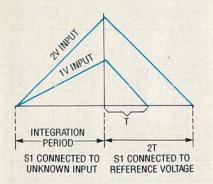


FIG. 3—DUAL-SLOPE INTEGRATION measuring sequence. The slope angle depends on the magnitude of the unknown input voltage and the time to discharge to zero is a function of that value.

positive. At the beginning of the second phase, the output from the clock is gated. Capacitor C1 is then discharged linearly by the reference voltage, and the clock is stopped when C1 is discharged through zero.

If the unknown input voltage is doubled, C1 charges up twice as fast. Since a constant reference voltage is applied to discharge the capacitor, the discharge rate will be constant. This means that the time to discharge C1 is doubled if the unknown input is doubled, as shown in Fig. 3. The accuracy of the dual-slope technique depends primarily on its reference voltages. This method is favored because errors introduced during charging are canceled during discharge. Some benchtop DMM A/D converters can now take 1000 readings per second. Hewlett-Packard's HP 34401A DMM can, for example, makes up to 50 range/function changes per second.

Figure 4 is a simplified diagram of a typical DMM circuit for measuring both AC and DC voltage. Many different proprietary circuits are used for true RMS AC and DC voltage conversion. As can be seen, the output of the true RMS converter goes to the A/D converter.

Both AC and DC current can be measured with the circuit shown in simplified form in Fig. 5. Again, the range switch block represents either manual or automatic range functions. Fuse F1 can represent two fuses in series, one rated for low current values, and the other rated for

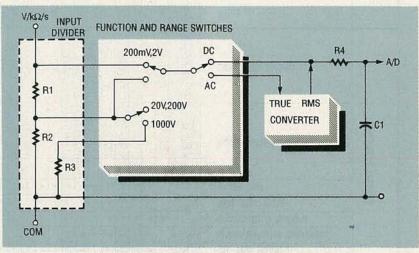


FIG. 4—SIMPLIFIED DIAGRAM of a typical voltage-measurement circuit.

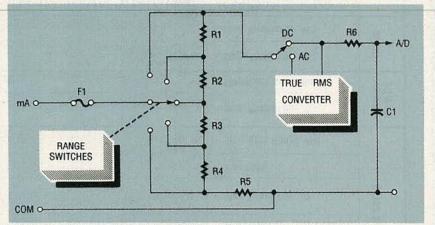


FIG. 5—SIMPLIFIED DIAGRAM of a typical current-measurement circuit.

high current values. Those protect the meter against accidental damage and the user from shock. The true RMS converter functions in both the voltage and current measurement circuits and, as in the voltage circuit, the output goes to the A/D converter.

Several different resistance measuring circuits can be found in today's crop of bench DMM's. The most popular are the constant-current source and voltage-ratio techniques. The constant-current source supplies current to the unknown resistance value. The DMM then measures the voltage drop across the unknown. In the voltage ratio-method, shown in simplified form in Fig. 6, there is a voltage source and an internal divider consisting of a reference resistor  $(R_{REF})$  and the unknown value  $(R_X)$ . The value of V<sub>RX</sub> is obtained from the lower op-amp.

Figure 7 illustrates a conventional 2-wire ohms converter found in most bench DMM's. It is based on a DC constant-current source. Figure 8 is a simplified schematic of a four-wire ohms converter that will eliminate measurement errors introduced by lead resistance in precision DMM's.

Alternate or optional battery power enhances the bench/portable meter's portability and improves its isolation from the AC line for certain critical measurements. It also ensures the safety of the user when making certain kinds of high-voltage and current measurements. Most low-cost, battery-powered bench DMM's use multiple "C" or "D" cells, or 9-volt transistor batteries, typically disposable alkaline. However, the battery power sources of the higher performance meters are typically multiple rechargeable nickelcadmium (Ni-Cd) cells or sealed 34

#### TABLE 1—BENCHTOP DIGITAL MULTIMETER CHARACTERISTICS

			BASIC FEATURE	s		OC VOLTS	1		AC	AC & DC AMPS OHMS					SPECIAL FEATURES											
		er of Digits	ay Counts		Basic Accuracy, %	Max. DC Voltage w/o Probe	Max. Resolution, µV	Basic Accuracy, %	Max. AC Voltage w/o Probe	Resolution, µV	RMS	Frequency Range, kHz	Resolution, µA	Amps w/o Probe	Resolution, $m\Omega$	Resistance, $M\Omega$	Auto/Manual Ranging	Frequency Measurement	dB, dBm Readout	Offset/Relative Reference	Continuity & Audible	Min./Max. Hold	Diode Test	System Interface	Other (See Notes)	
Manufacturer	Model No.	Number	Display		Basic	Max. w/o P	Max.	Basic	Max. w/o P	Max.	True RMS	Frequ	Max.	Max	Max.	Max.	Auto/	Frequ	dB, d	Offse	Conti	Min./	Diod	Syste	Othe	Price
Analogic	DP100	5½	200,000	LCD	0.0035	450	0.1	0.35	450	1.0	•	50	1.0	2.0	1.0	20		•	-	•	-	_		232(ST)	(B/AC)(DM)(TM)	595
B&K	2831A	31/2	2000	LED	0.1	1200	100	0.5	1000	100	-	40	0.1	20	100	20	-	-	-	-	•	1	•		-	295
	2832	3½	2000	LCD	0.5	1000	100	1.0	1000	100	- 1	500	0.1	20	100	20	-	-	-	-	•	1	•	-	(B/AC)(CA)	195
-	2833	41/2	20,000	LCD	0.05	1000	10	0.5	1000	10	•	50	0.01	20		20	-	•	•	_	•	-	•	a finite a finite da	(B/AC)(DH)	356
Beckman	350	31/2	2000	LCD	0.1	1500	100	0.6	1000	100	-	10	0.1	10		20	1	-	-			-	•			259
Deckindi	360B	3%	2000	LCD	0.1	1500	100	0.6	100	100	•	40	0.1	10		20	-	-	-	-	•	-	•		(TM)	329
Daetron	MM300A	41/2	20,000	LCD	0.2	500	10	2.0	500	1000	-	0.06	1000	2		20	•	•	-	•	•	•	-	232(OP)	(BG)(M)(PH)	450
Fluke	37	31/2	3200	LCD	0.1	1000	100	0.5	1000	100	-	30	0.1	10		32	•	-	-	•	•	•	•		(BG)(CO)(DH)	269
TIUNC	45	5	100,000	FL	0.02	1000	1.0	0.2	750	1.0	•	100	0.1	10		300	•	•	•	•	•		• 2	232(ST),488(OP)	(DH)	635
	8010A	3½	2000	LCD	0.1	1000	100	0.5	750	100		50	0.1	10		20		-	-	-	-		•		(CO)	379
A CONTRACTOR OF CONTRACTOR		31/2	2000	LCD	0.1	1000	100	0.5	750	100	•	50	0.1	2	1	20	-						•		(CO)	389
	8012B			LCD	0.03	1000	100	0.5	750	10		50	0.01	2	10	20			•				•		(CO)	479
	8050A	41/2	20,000	the second second second	0.005	1000	1.0	0.16	1000	1.0	(OP)	100	10	2	10	10		No.		10/5	Sector	(minit)		488(OP)	(00)	875
	8840	5½	200,000	FL				and the second second second second	1000	0.1	(OP)	100	1.0	2	0.1	20							-	488(OP)		1095
	8842A	5%	200,000	FL	0.003	1000	0.1	0.08		10	(UP)	50	0.1	10	10	20	1		1000		1000	•	-	400(UF)	(DH)	229
Goldstar	DM7241	4½	20,000	LCD	0.05	1000	10	0.5	750	10	-	50	0.1	10	10	20	-	-	-	_	-	-	-		(DD)	229
Hewlett-			1 000 000		0.0005	1000	0.1	0.00	750	0.1		200	0.01	2	0.1	120								000 A00(CT)		995
Packard	34401A	6½	1,200,000	FL	0.0035	1000	0.1	0.06	750 300	0.1		300 300	0.01	3		30	•			10000	-		ACCESSION OF	232, 488(ST)	(M)(DM)(PH) (B/AC)	1095
	3468A/B	5½	300,000	LCD	0.018	300	1.0	0.26	300	1.0		300	1.0	3		30				1000		1		488(ST)	(D/AU)	1295
	3478A	5%	300,000	LCD	0.006	300 1000	1.0	0.20	750	10		100	0.01	10		200	•			Series -		•		400(01) 488(0P)	(B/AC)(OP)	495
Keithley	175A	4½	20,000	LCD	0.03		10		300			100	0.001	3		300				•		•		488(ST)	(D/AU)(UP)	1495
Surger and	196	6½	3,029,999	LED	0.003	300	0.01	0.15	750	1.0		100	0.001	10		220		1000	•		THE OWNER		-	488(0P)	(B/AC)(OP)	659
in the second	197A	5%	219,999	LCD	0.01	1000	1.0	0.35								300	•	-	-			-	-		(B/AC)(UP)	
State of the state	199	5½	302,999	LED	0.006	300	1.0	0.15	300	1.0	1.22	100	0.01	3		1000	•	•	•	•	1	•	-	488(ST)	(05)(714)	1045
	20001		21,000,000	FL	0.0007	1100	0.01	0.03	750	0.1	•	2000	.00001		0.001		•		•		-		-	488(ST)	(CF)(TM)	2695
Kenwood	DL712	3½	2000	LCD	0.1	1100	100	0.75	850	1000	1000	0.5	0.1	10		20	1.00	-	-	-	-			000 100/00	(B/AC)	249
Leader	856	41/2	20,000	LED	0.05	1000	10	1.0	750	100	•	100	10	3		30	•	•	•	•	•	-		232 or 488(OP)	(BG)	800
Protek	HC-797	4%	32,000	LCD	0.04	1000	10	0.4	750	10	•	30	0.1	10		30	•	•	-	•	•	•	•	232(ST)	(BG)(CF)(M)	350
Simpson	460-6	41/2	20,000	LCD	0.07	1000	10	1.5	750	10	•	100	0.1	10	10	20	-	-	•	-	•	-	•		(BG)(B/AC)	500
	461-2R	3½	2,000	LED	0.10	1000	100	0.5	750	100	•	50	0.1	2	100	20	-	-	-	-	-	-	•	Real Person	(B)	300
Second Second	464-4	3%	2.000	LED	0.10	1000	100	1.0	750	100		100	0.1	10	100	20	-				•		•			360

	1000	111	01 000		000	1000	10	40	1000	10	•	1001	0.01	0		• 56	•	•	• • • •	-	(MT/(NV/M)	600
Iriplett	4800	4%	000,62	LUU	0.04	1000	2	0.0	0001	2		20-	10.0	1		+					//	
Ynkanawa	7551	5%	200.000	LED	0.005	1000	1.0	0.2	700	1.0	L	0.2	0.01	2	1.0 2	• 00	•	•	1	- 232 or 488	(W)	895
	7552	5%	200.000		0.005	1000	1.0	0.2	200	1.0	1	1	100	20	1.0 2	200 •	•	•		- 232 or 488	(W)	995
	7561	6%	2.000.000	LED	0.003	1000	1.0	1	1	1	1	1	1	2	0.1 2	• 00	•	•		- 232 or 488	(M)	1195
	7562	6%	2,000,000		0.003	1000	1.0	0.15	700	1.0	I.	I	0.01	2	0.1 2	• 00	•	•	1	- 232 or 488	(W)	1295
	10000													-								

RS-232 compatible IEEE-488 (GP1B) compatible

Temperature measurement

(PH) 232 232 488

Conductance measurement

Crest factor Conductance Display hold

M) DH)

Battery operation Battery & AC line powered Capacitance measurement

(BG) (B) (CA) (CA)

graph

Bar

NOTES:

Memory

Optional

(OP)

Standard

(LS)

Peak hold

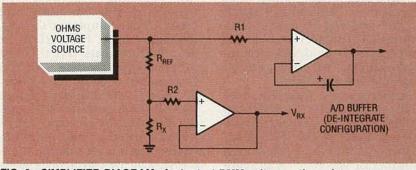


FIG. 6—SIMPLIFIED DIAGRAM of a typical DMM voltage ratio resistance-measurement circuit.

lead-acid batteries. Manufacturers offer power options: 117 or 220 volts, 50 to 60-Hz line, and/or batteries.

Some bench DMM's have been designed for dual use as portable or rack-mounted instruments. Rack mounting is used in systems applications. The purchaser can specify the case style desired.

#### Available bench DMM's

If you examine Table 1, Digital Multimeter Characteristics, included here you will see 3-1/2digit bench/portable DMM's list priced below \$200, and higher performance instruments priced for more than \$2600—a price spread of better than an order of magnitude. Those prices include only the bare necessities such as basic test leads, power cord, and manuals. Most DMM accessories such as probes and carrying cases are extra expense items.

The basic electrical measurement functions in bench DMM's are usually supplemented with special features, many of which are can also found in the handhelds. The most common examples are diode test and audible continuity. As stated earlier, bench models are more likely to include true RMS AC voltage and current measurement than handhelds, and system compatibility is an exclusive feature with the bench models. Many bench DMM's can also display such math functions as min/ max hold, and provide readouts in dB and dBm.

With their focus on accurate measurement, many DMM's have only a few of the special features found in handhelds.

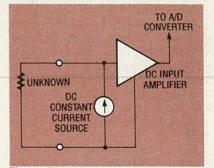


Fig. 7—SIMPLE 2-WIRE ohms converter.

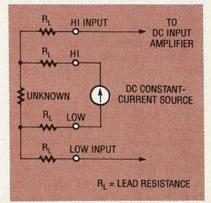


FIG. 8—SIMPLIFIED 4-WIRE ohms converter for eliminating lead resistance errors. Current is supplied by a separate current source.

The line separating the features of handhelds from those of benchtop DMM's is fuzzy, at least in the \$200 to \$300 price price range. Features such as peak hold, sleep mode, and capacitance measurement are more likely to be found on the handhelds.

But it's worth keeping in mind that, generally speaking, you get what you pay for in bench/portable DMMs. Prices have been raised on many of the models that have been on the market for ten years or more. However, there is general agreement that you can get more for your DMM dollar today than ever before. This is due, in part, to the ongoing transfer of advanced A/D converter IC's and microprocessor devices to instrumentation, and intense worldwide competition.

Experts agree that the most critical parameter in benchtop DMM's (and the one you should look at first) is DC voltage accuracy. It can be stated in a number or ways: ± (% of reading + number of least significant digits),  $\pm$  (% of reading + % of range),  $\pm$  (% of reading + number of counts), and even parts per million (ppm) of reading as  $\pm$  (ppm of reading + ppm of range). Regardless of the form used, it will be given at a specified ambient temperature, usually between 18°C and 28°C, with relative humidity up to 90%.

Accuracy ratings might be given for periods of one year, six months, 90 days, or even 24 hours. To cut through all this complexity, most manufacturers give short forms of their accuracy specification known as basic DC-voltage accuracy. That is what we used in our characteristics table.

After you have checked out basic DC-voltage accuracy, you should study the specifications and literature to be sure that the meter you are considering is a quality product built ruggedly enough to meet your needs and includes all of the accepted user safety and instrument protection features for that class of product. Next you might want to examine the basic AC-voltage accuracy rating.

A 5-1/2- or 4-1/2-digit DMM can be expected to offer higher accuracy and resolution than a 3-1/2-digit DMM, and this should be kept in mind during your search. The 1/2 in the specification of the DMM display refers to the use of the digit 1 in the most significant digit (MSD) position. The full scale on a 4-1/2 digit display, for example, is 9999, but the additional digit 1 permits the display to show a value that is 100% higher (19999) in what is known as the 100% overrange condition.

There is a close but not abso-

lute correlation between the number of display digits and counts. For example a 3-1/2-digit DMM might have 2000 or even 3200 counts, a 4-1/2-digit meter typically has 20,000 counts, but it could have more. A 5-digit DMM will have 100,000 counts, but a 5-1/2-digit meter would be expected to have a count of at least 200,000. Rather than try to define such confusing designations as a 4-3/4 -digit DMM, it is a lot easier to look for the display count that we have listed in our selection table.

As you move upscale in bench DMM's, prices rise accordingly. Therefore, it becomes even more important to pay attention to the reputation of the manufacturer, and inquire about any guarantees or warranties being offered. Not surprisingly, even some reputable manufacturers who are willing to offer 3-year warranties on their handheld and low-cost bench DMM's are reluctant to offer more than a one-year warranty on their most sophisticated bench models. However, one-year warranties can usually be extended for a fee. We suggest that you carefully read and compare specifications, literature, and any available evaluations of products before you buy any bench DMM.

In searching for the bench DMM best suited to your needs, you should consider the value of such features as autoranging vs. manual ranging. Autoranging automatically determines the proper range and polarity for the DMM to display a measurement with the best resolution. Manual ranging allows you to override the autoranging function and make manual se-



ANALOGIC'S DP 100.



**BECKMAN INDUSTRIAL'S MODEL 3** 

lections. While some DMM have both, many high-performance DMM's do not incluautoranging. This is acceptate because those instruments were those instruments were the ranges of u known signals are general known.

#### **Measurement** capabilities

In addition to basic DC-vo age accuracy, you will want know the maximum DC volta that can be measured without plug-in probe, and maximu DC-voltage resolution. Bas DC-voltage accuracy will ran from 0.1% in 3-1/2-digit (200 count) meters to 0.003 % 5-1/2-digit (200,000 coun meters. A rating of 1000 vol maximum DC voltage is con monplace in modern bene DMM's; it will be obtained at t high end of as many as fi ranges, typically 200 mV, 2 V, 1 V, 200 V, and 1000V. Maximu DC-voltage resolution shou typically be a value of 100 mici volts or less.

The AC-voltage specification also include basic accurate maximum resolution, and ma imum (RMS) value without to use of a probe. Frequency ran in hertz over which AC me surements are valid is anoth variable included under *A* volts. Basic AC-voltage accura is typically a fraction of the DMM's DC-voltage accuracy; some cases it is as much as a order of magnitude less.

Maximum AC resolution, al measured in microvolts, ty ically matches the values for I volts. Maximum voltages f bench DMM's are typically t tween 750 and 1000 volts / RMS. Expect five AC volta ranges comparable to the I voltage ranges. The ability measure true RMS voltage is popular feature for bence

36

DMM's; most bench DMM's either offer this feature or make provision for it as an option.

Maximum ratings for AC and DC current in amperes that can be measured without a probe are also important for the bench DMM user. Don't be surprised to find that handhelds have higher maximum current ratings than high-performance bench meters to protect against the unknown conditions encountered in making current measurements in the field. They could be only 2 amperes. However, you can expect the bench meter to have better current resolution-0.1 microampere or better.

Maximum resistance values that can be read on a bench DMM are typically 20 megohms or better, but they could be as



#### FLUKE'S DUAL-DISPLAY MODEL 45

high as 300 megohms. However, maximum resistance resolution could be 1 milliohm or less in the high-end models. The diode tests and conductance measurement functions are less popular in highperformance DMM's.

#### **Special features**

Beyond the capability for measuring the five basic electrical parameters and doing diode tests, many DMM functions are considered to be special. In general, with the exception of system compatibility, none are exclusive to bench DMM's.

Because of their normally controlled working environments, accuracy and resolution outrank versatility in the selection of a bench DMM. It has been found that specialized temperature, frequency, and even capacitance measuring instruments are preferred over



**GOLDSTAR'S MODEL DM-7241** 

those functions in DMMs used in labs or shops. Manufacturers will, however, include these functions if they find a demand for them. But if you don't need them you could be paying a lot for a feature with lower performance than is obtainable in a specialized instrument.

By contrast, the handhelds follow the Swiss Army knife philosophy of stuffing as many "tools" as is practical in a single package. Battery conservation and protection against personal shock and destruction of the DMM are, as you might expect, more important characteristics in handhelds.

True RMS voltage and current: This feature provides accurate measurement of nonsinusoidal waveforms such as square waves, pulses, or the outputs of silicon-controlled rectifiers. This function, either standard or optional, is widely found in bench DMM's, but less often found in handhelds. You can be sure that any DMM with the true RMS measurement feature will cost more than a comparable meter without that feature.

System compatibility: Some bench DMM's offer an EIA RS-232C and/or IEEE-488 interface as either standard or optional features. The RS-232C interface allows data to be interfaced to any serial printer or computer. The data can be filed, manipulated, printed, or transmitted by modem. Internal DMM circuits format the measurement data transmission. Host computer software permits remote operation of all instrument functions.

IEEE-488 (also known as the general purpose information bus or GPIB), is a parallel inter-

face bus that consists of eight bidirectional data lines and eight signal grounds (three wires for handshakes between equipment, and five wires for management). Some DMM's permit both RS-232C and GPIB interfaces. Although RS-232C interfaces are becoming more popular because of the proliferation of personal computers in laboratories, the IEEE-488



HEWLETT-PACKARD'S MODEL 34401A



bus is still widely used in industrial data acquisition systems.

Decibel (dB) and dBM readout: DMM's with this function measure and display the dB gain or loss of amplifiers, filters or attenuators. The dBm readout is referenced to 1 milliwatt and 600 ohms.

Offset/relative reference: This function stores the input in memory as a zero "reference". Any subsequent input is automatically compared to the reference in memory, and the display shows the difference  $(\pm)$  between these two values. It is also a handy feature for nulling out test lead resistance or measuring the dB gain for the stages of an audio amplifier. Relative reference works in all functions and ranges, and is a feature found on high-end handhelds and low-end benchtops.

Frequency or period: Some DMM's can also measure frequency or its reciprocal period. Bench-type DMM's are likely to have frequency responses in the megahertz range, especially if they have six or more readout digits. Some DMM's now have AC voltage input ranges as high as 15 MHz and AC current input to 1 MHz. These can be translated into period. You will want to know accuracy, sensitivity, maximum input, and trigger level for these measurements.

Temperature measurement: Some bench DMM's have the capability for measuring temperatures with one or more standard thermocouples or a resistance temperature detector (RTD). Here again, the specified accuracy and resolution will be important in your decision if you want this feature in your multimeter.

Continuity beeper and diode test: Continuity, diode and transistor checks can be made quickly on DMM's with this feature without looking at the display. A continuous tone indicates continuity, while a beep signals a forward-biased diode or transistor. The beeper is easily heard in a noisy industrial environment.

Analogic Corp. 8 Centenial Drive Peabody, MA 01960 (508) 977-3000 CIRCLE 301 ON FREE INFORMATION CARD

Beckman Industrial Corp. Instrumentation Products Div. 3883 Ruffin Road San Diego., CA 92123-1898 (619) 495-3200 CIRCLE 302 ON FREE INFORMATION CARD

B&K Precision 6470 W. Cortland St. Chicago, IL 60635 (312) 889-1448 CIRCLE 303 ON FREE INFORMATION CARD

Daetron 935 The Queensway, Box 641 Toronto, Ont. Canada M8Z-5Y9 (416) 676-1600 CIRCLE 304 ON FREE INFORMATION CARD

John Fluke Mfg. Co. Box 9090, MS 250E Everett, WA 98206 (800) 443-5853 (800-44-FLUKE) CIRCLE 305 ON FREE INFORMATION CARD

Min/max hold: DMM's with this feature store the highest and lowest readings, permitting you to monitor a signal for seconds, or even days. Collected average values during the period are calculated and displayed, and when the recording period ends, you can examine the readings at your convenience. Overloads or manual stop/starts won't erase memory until you give the command. Min/max recording is possible when measuring AC and DC voltage or current, and resis-



#### **B&K PRECISION'S MODEL 2833**

tance. This feature is also found on some handheld meters.

Analog Bar Graph: The analog bar graph is a segmented analog needle simulator. It performs the same role as a VOM needle, while eliminating the mechanical/inertial distortion

#### BENCHTOP DMM SOURCES

Goldstar Precision 13013 E. 166 St. Cerritpos, CA 90701 (213) 404-0101 CIRCLE 306 ON FREE INFORMATION CARD

Hewlett-Packard Company 19310 Pruneridge Ave. Cupertino, CA 95014 (800) 752-0900 CIRCLE 307 ON FREE INFORMATION CARD

Keithley Instruments, Inc. 28775 Aurora Road Cleveland, OH (216) 44139 (216) 248-0400 CIRCLE 308 ON FREE INFORMATION CARD

Kenwood USA Corp. 2201 E. Dominguez St. Long Beach, CA 90810 (213) 639-4200 CIRCLE 309 ON FREE INFORMATION CARD

Leader Instruments Corp. 380 Oser Ave. Hauppauge, NY 11788 (516) 231-6900 (800) 645-5104 CIRCLE 310 ON FREE INFORMATION CARD inherent in a needle's mov ment. This feature is not wid available in bench DMM's.

Sleep mode: This feature a tomatically shuts off power you forget or if no measur ments are taken for a specifi length of time, say 60 minut This conserves battery life. It found in some battery-power bench meters.

Peak hold: This feature useful for recording transier as low as 1 millisecond, of pecially from intermitte power lines or connection This mode can also be used measure the plus and min peak values of sine waves up about 450 Hz. It permits ea measurement of both peak li voltage and line current power supplies and electric equipment. However, it is rar found on bench DMM's.

Data hold: This feature known by different proprieta names such as "touch hold" "probe hold." It allows you keep your eyes on the prot and on the circuit under te The DMM's microcomputer ( termines when the input sign is steady, alerts you with a be and then captures and hol the measurement on the disp until you are ready to view it

Prema Precision Electronics Inc. 4650 Arrow Highway Building E-5 Montclair CA 91763 (714) 621-7292 CIRCLE 311 ON FREE INFORMATION CARE

Protek Inc. P.O. Box 59 Norwood, NJ 07648 (201) 767-7242 CIRCLE 312 ON FREE INFORMATION CARE

Simpson Electric Co. 853 Dundee Ave. Elgin, IL 60120 (708)697-2260 CIRCLE 313 ON FREE INFORMATION CARE

Triplett Corp. One Triplett Drive Bluffton, OH 45817 (419) 358-5015 CIRCLE 314 ON FREE INFORMATION CARD

Yokogawa Corp.-America 2 Dart Road Newnan, GA 30265 (404) 253-7000 CIRCLE 315 ON FREE INFORMATION CARE

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automatically updates with each new measurement.

#### **Reading spec sheets**

Manufacturers' specification sheets can be quite confusing because they do not follow a uniform format. They can be especially intimidating for the first time DMM buyer trying to decipher the baffling terminology. The wise buyer should understand such terms as CMR, NMR, and RMS, average, and peak AC readout.

Definitions of some of the commonly used terms in DMM specificastions sheets will be helpful to you. Remember that the manufacturers are trying to attract your attention to their instrument's more glamorous features which may or may not be something you need. So brochures do not necessarily rank the qualities that are of most importance to you in any logical order. The burden for interpreting these specifications falls on you, the prospective buyer and user. But this also holds for a lot of other purchases such as appliances, automobiles or homeentertainment products.

Noise is an especially severe problem in DMMs due to their high accuracy, resolution, and sensitivity. DMM's used in systems are more susceptible to noise than those used just in isolated measurement applications. Long signal leads and coupling between adjacent signal leads compound the problem in systems.

Noise can be defined by its origin relative to signal input lines on the DMM. Normal-mode noise enters with the signal and is superimposed on it. Common-mode noise is common to both the high and low signal inputs. Common mode noise becomes normal-mode noise when it flows into the DMM's signal inputs.

Normal-mode noise originates from power-line pickup, electromagnetic fields or even from within the device being measured. Noise can be sinusoidal, spikes or white noise. There are two techniques used to reduce normal-mode noise: integration and filtering. Inte-



LEADER'S MODEL 856



SIMPSON'S MODEL 460-6



TRIPLETT'S MODEL 4800

gration stretches the measurement out over a fixed period of time during which amplitude variations are averaged out. If the integration period includes a sufficient number of periodic noise cycles, the noise will be averaged out.

Filtering slows down the conversion process or measurement speed so it is used judiciously. Integration does a better job of rejecting line-related noise, and filtering is better for broadband noise.

Common-mode rejection (CMR) is a measure of the change in output voltage when both inputs are changed by equal amounts of AC and/or DC voltage. It is the measure of an instrument's ability to cancel undesirable signals entering the measurement circuit between the input and ground. A reasonable value for CMR in a bench DMM is 100 volts DC or peak AC from any earth input.

Normal-mode rejection (NMR) is a logarithmic measure of at-

tenuation of normal-mode noise components at specified frequencies in dB. For an amplifier used in instrumentation, the normal-mode signal is the actual difference signal being measured. This signal often has noise associated with it.

DC voltage uncertainty = [(ppm of reading) × (measured value) + (ppm of range) × (range used)/1,000,000.

*Percent uncertainty* = (ppm uncertainty)/10,000.

AC voltage uncertainty = [% of reading) × measured value) + (% of range) × (range used)]/100.

*Crest factor* (CF) expresses the waveform's peak value ratio to that of its RMS value. Crest factors in true RMS meters actually specify their dynamic ranges.

*Self-test*: DMM's with a microprocessor-controlled self-test feature usually run through this routine when they are turned on and the results show up on the display as segments that appear as rapidly as four times per second.

Self-check capability is extremely important for the assurance of accuracy and continued calibration to laboratory or international standards. Nevertheless, all DMM's, even those with this feature should be recalibrated at least once a year for maximum accuracy and reliability.

Imput impedance is the combined AC and DC resistance at the input of the DMM. An input impedance of 10 megohms or better virtually eliminates measurement errors caused by loading in most circuits.

The outstanding characteristics of bench-type DMM's have been highlighted primarily on a basis of published specifications and unusual features. In an evaluation section that follows we have selected a crosssection of available benchtop DMM's and discussed some of their outstanding features. All of the bench multimeters discussed in this article are capable of making the five basic electrical measurements: AC and DC voltage, AC and DC current and current.

#### Selected reviews

Analogic offers the DP100, with a 5-1/2-digit LCD display capable of measuring voltage current and resistance. It is also a frequency counter capable of measuring up to 25 MHz as well as an RTD temperature meter. The portable DP100 is powered from both built-in rechargeable batteries and the AC line. It offers  $\pm 0.003\%$  basic DC-voltage accuracy, 0.1-microvolt sensitivity and the ability to measure true RMS AC voltage and current, and 2- or 4-wire resistance. The DPM includes RS-232C compatibility or optional IEEE-488.2 converter.

B&K Precision offers three bench DMM's all capable of making the five basic measurements-AC and DC voltage and current, and resistance. The model 2831A is a 3-1/2-digit meter with a LED display that also offers continuity checking and diode test. It has a 0.1% basic DC-voltage accuracy rating. The model 2833 with a 4-1/2-digit LCD display, features true RMS voltage and current measurement. It has basic 0.05% DC-voltage accuracy. Other features include audible continuity check, diode test, data hold, dBm readout, and frequency measuring.

The B&K model 2832, with a 3-1/2 digit LCD display, measures capacitance. It's basic DCvoltage accuracy is stated at 0.5%. Other features include audible continuity check and diode test. This DMM can be powered from the AC line or six "C" cells.

Beckman Industrial Corp. offers two 3-1/2-digit bench/ portable DMMs with LCD displays, the models 350 and 360 B. Both are packaged in similar cases with a front-panel rotary function switch, and both have basic 0.1% basic DC-voltage accuracy and 22-megohm input impedance. In addition to the five measurement functions, the meters provide audible continuity checking and diode testing. The model 360 offers true RMS AC voltage and current measurement as well as temperature measurement with a type K thermocouple. Both meters are powered by six "D" cells and are built for field use.

John Fluke Mfg. Co. offers a family of seven bench/portable DMM's with 3-1/2 to 5-1/2-digit displays. Some also offer diode test and others measure conductance. There is, however, considerable variation in the special features offered. The 3-1/2-digit model 37 has a basic DC-voltage accuracy of 0.1% and the 5-digit model 45 has a basic DC-voltage accuracy of 0.02%. The model 45 has a dual vacuum-fluorescent display and it can measure frequency and display dB and dBm. It also has the min/max hold feature.

Fluke is also offering two 5-1/2-digit models in the 8800 series, the 8840A and 8842A. The 8840A offers 0.005% basic DC-voltage accuracy, while the 8842A has 0.003%. Both feature optional true RMS AC voltage and the ability to measure frequency to 100 kHz.

Goldstar Electronics offers the DM-7241 with a 4-1/2-digit LCD display. It has a basic DCvoltage accuracy of 0.05% and a resistance range to 20 megohms.

Hewlett-Packard offers four benchtop DMM's. The HP34401A has a 6-1/2-digit vacuum-fluorescent display and features true RMS AC voltage and current measurement. Resistance is measured in ohms with 2- and 4- wire circuits. In addition, the model 34401A measures frequency, period, and continuity, and can do diode tests and DC:DC ratios. DCvoltage accuracy is given as 0.0035% while AC-voltage accuracy is 0.06%. The meter's bandwidth is 3 Hz to 300 kHz. Its math functions include null, min/max average, dB and dBm readout, and limit test. Both the IEEE-488.2 and RS-232C interfaces are standard.

Keithley Instruments offers five bench/portable DMMs. The model 2001 has a 7-l/2- digit display and a rated basic DC-voltage accuracy of 0.0007%. Basic AC-voltage accuracy is given as 0.03%, and bandwidth is 1 Hz to 2 MHz. The 2001 can measure resistance values from 1 micro-ohm to 1 gigohm. It takes 2000 readings per second wi 4-1/2-digit resolution, 300-50 with 5-1/2-digit resolution, an between 45 and 200 with 6-1/ digit resolution.

The 2001's standard me surement functions include *A* crest factor, frequency from 11 to 15 MHz, true RMS, and pea and average AC. It also mak DC in-circuit current measur ments, and offers simultaneor displays. Options include a te channel scanner. GPI (IEEE-488.2) output is sta dard.

Kenwood USA Corp. offers tl model DL-712 that has bo manual and autorangin Other standard features of tl DMM include diode test, da hold, and continuity checkin

Leader Instruments' mod 856 has a 4-1/2-digit LED di play which includes a b graph. Equipped with aut ranging, it measures true RM and frequency. The 856 al does diode test and continui checks, displays dB and dBi and makes data comparisor Its basic DC-voltage accuracy 0.05%.

Simpson Electric offers fo 4-1/2- and 3-1/2-digit bencht DMM's. They all feature tri RMS AC measurement. Tl 4-1/2-digit 460-6 offers Ni-( battery/AC-line operation ai 0.07% DC volts accuracy. T 467-2, a 3-1/2-digit DMM, h peak hold.

Triplett Corp. offers the moc 4800, a 4-1/2-digit bencht DMM with true RMS readou Its features include data mem ry, peak hold, dBm readout, a toranging, temperature me surement, and diode test.

Yokogawa offers the 5-1/2-d it models 7551/7552 and t 6-1/2-digit models 7561/756 They are rated for basic D voltage accuracies of 0.005 and 0.003%, respectively. The DMM's can be calibrated wi external signals. The 7552 a 7562 offer RMS AC measur ment, while the 7551 offers 1 mean measurement. The 755 however, offers both frequer measurement and a 20-ampe range. The RS-232C IEEE-488 interfaces are sta dard on all these models.

# **CONTROLLER**

#### EDWARD J. KEEFE JR.

MOST SMALL BANDS USUALLY HAVE such a hard time paying for travel and instruments that their shows must forgo any type of sophisticated lighting. The lights provided by the clubs and bars where they play do little to showcase a band's talent. With the MIDI (Musical Instrument Digital Interface) light controller presented here, that can all end. A simple microprocessor with a handful of components can transform their act into a full fledged "concert." Everything that is needed to synchronize lights and music already comes out of the MIDI port of MIDI keyboards. This circuit will make use of that information and enhance the show.

MIDI is a communications protocol originally created for interfacing synthesizers and other electronic devices. It has evolved into a communications standard that is used in all phases of audio and video proOur MIDI light controller can turn an ordinary musical performance into a concert!

duction. MIDI allows devices to talk to each other with different types of control and data values. The values can be either CHANNEL, SYSTEM. REAL-TIME. OT SYSTEM EXCLUSIVE messages. MIDI communication is achieved through multi-byte messages, each consisting of one status byte followed by one or two data bytes. Real-time and exclusive messages are exceptions to that rule.

Two types of data bytes are sent over the MIDI cable, STATUS and DATA. Status bytes are eightbit binary numbers in which the most-significant bit is set to "1." A status byte sets the function of the data bytes that follow it, and a new status byte is required for each new action. The MIDI specification also outlines RUNNING STATUS. That defines the action for all data bytes that follow a status byte, until a new status byte is sent. That way more information can flow down the cable. Data bytes are eight-bit binary numbers in which the most-significant bit is set to "0."

The MIDI light controller presented here reacts to NOTE ON, NOTE OFF, START. STOP. and CONTINUE status bytes. The data bytes are used to determine which light to control and how

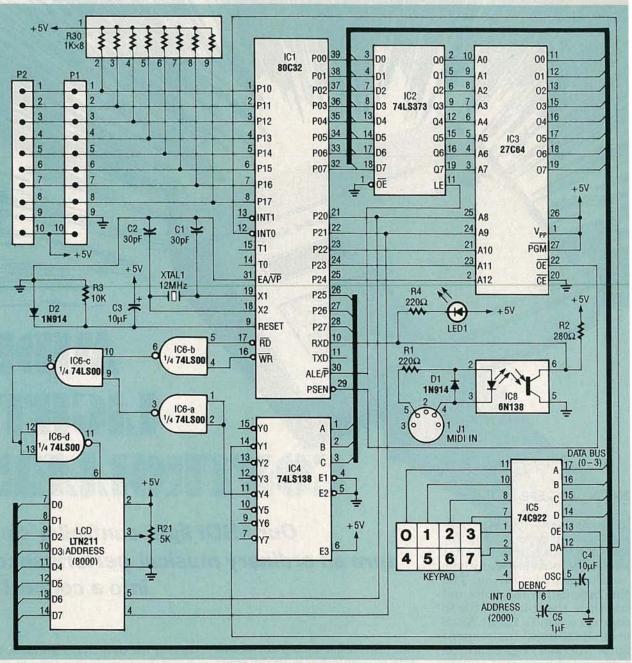


FIG. 1—SCHEMATIC OF THE MAIN PROCESSOR BOARD. The MIDI signal is fed in through an optocoupler (IC8) to the microcontroller (IC1).

to respond to different conditions on the MIDI. It can operate on any of the 16 MIDI channels, and will respond to note information from octaves 1–8. The MIDI light controller is userconfigured. You have control over whether the lights will latch, toggle, or stay off in reaction to note information on the MIDI by setting the OUTPUT TYPE keyboard selection.

When a STOP command is received (STOP, START, and CONTINUE commands synchronize all elements of a MIDI system), the lights can either all go off, all go on, or light selectively, depending on the user's stop DEFINE parameters. That is useful for creating a "scene" between songs and during breaks on stage. The CONTROL OCTAVE determines which range of notes will be used to control the lights. The operating channel is set by selecting CHANNEL DEFINE. The message CHANGE CHANNEL # will be displayed on line 1. Line 2 will toggle between "1" (channels 1–8) and "2" (channel 9–16). You must select "1" or "2 for the desired range of char nels (1–8 or 9–16) and select th desired channel.

The lights are controlled by note information on the MID They can respond to actual notes of a song or notes that an placed in the sequence specifically for the light controller. The use existing notes, you would select the channel and octave from the sequence, and enter them on the keyboard. If the lights are to be controlled by separate track, you would enter

the note information for each light, keeping all notes in one octave. Timing and synchronization is provided by a sequencer. (A sequencer is any instrument that can store and read back MIDI data.)

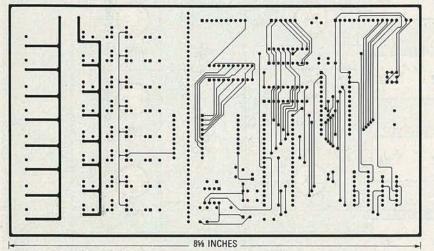
The light controller will also work directly from a keyboard without the aid of a sequencer by connecting a keyboard's MIDI output to the jack on the light controller. As you play the keyboard, the appropriate lights will illuminate.

#### The circuit

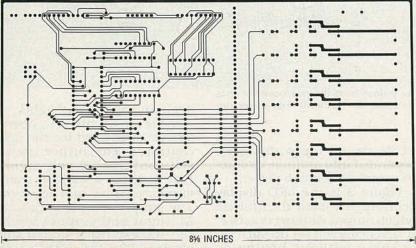
Figure 1 shows a schematic of the main processor board. The MIDI signal is fed in through IC8, an HP 6N138 optocoupler. Any optocoupler can be used as long as it has a rise time of less than 2 microseconds, and can turn on with less than 5 mA. The MIDI signal has the following operating specifications:

31.25-kHz baud rate, asynchronous, 1 start bit, 8 data bits, and 1 stop bit, with a period of 320 microseconds per serial byte. The signal is well suited to the serial portion of the Intel 80C32 microcontroller (IC1) that has built-in transmit and receive serial ports, 128 bytes of internal RAM, two interrupt lines (each with programmable priority), and external memory addressing capability up to 64K. The CMOS version of the 8032 was used because of its higher speed and lower power consumption.

The output of optocoupler IC8 enters the serial receive port (pin 10) of microcontroller IC1. Software in the 27C64 EPROM (IC3) controls port setup and baud-rate selection. At power up, a small reset circuit (R3, D2, and C3) initializes the microcontroller. An oscillator is made up of a 12-MHz crystal (XTAL1)



COMPONENT SIDE FOIL PATTERN for the light controller.



SOLDER-SIDE FOIL PATTERN for the light controller.

and two 30 pF capacitors (C1 and C2).

The address and data buses of the microcontroller are multiplexed. To remove the low-order address information, a 74LS373 8-bit latch is used. The 74LS373 is strobed with the ALE (ADDRESS LATCH ENABLE) signal from the microcontroller, and address data is removed. The 27C64 EPROM (IC3) is used to store program data. The EPROM is enabled by the microcontroller's PSEN (PROGRAM STORE ENABLE) line.

The keyboard and LCD circuits both require an address signal to interact with the data bus. A 74LS138 (IC4) is used to generate a signal when certain addresses are reached (1000h, 2000h, 4000h, 6000h...). The keyboard is mapped at external address 2000h. The keys are scanned by the 74C922 keyboard controller chip, IC5. All keyboard action, including debouncing, is handled by IC5. The scan rate is controlled by C4, and debouncing by C5.

The LCD is mapped at external address 8000h. An Optrex LTN211 two-line LCD module is used to display current channel and mode information. Also, user-definable selections are displayed there. Function data is written to the module at address 8000h, and display data is written at 8200h. Address selection and read/write functions are established by IC6, a 74LS00. Contrast of the LCD is altered by R21. Output from the microcontroller is on port 1, pins 1-8. A resistor network, R30, is used to pull-up output lines. The lines drive the Triac and LED sections.

Figure 2 shows the highpower output section of the light controller. Eight identical drivers are used to control the output channels. An output from IC1 is passed through a 1N914 diode and a 120-ohm resistor that drives an MOC3010 Triac-driver optocoupler. Output from the optocoupler is sent to a high-power Triac. The circuit is designed with 6-amp Triacs, which are fused at 5 amps for added protection. The 120volt AC input to the Triacs must

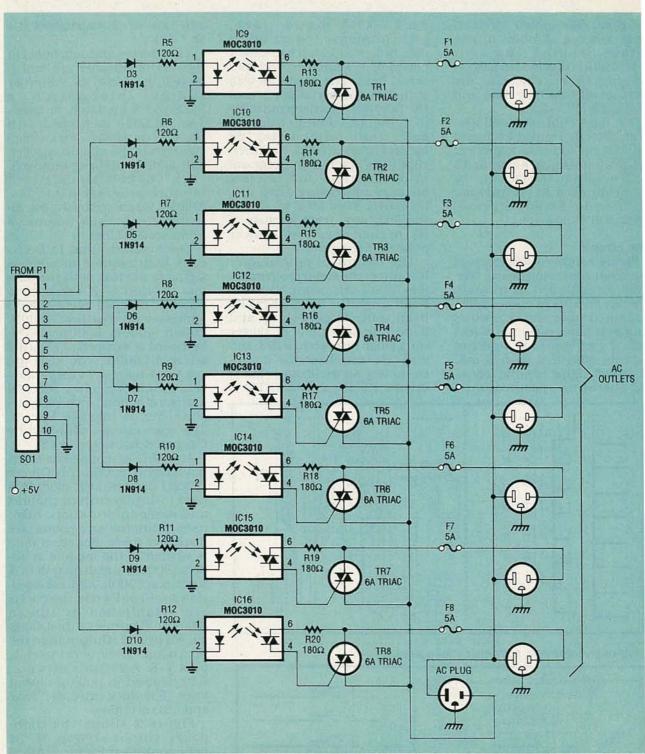


FIG. 2—HIGH-POWER OUTPUT SECTION. An output from the microcontroller is passed through a 1N914 diode, a 120-ohm resistor, an MOC3010 Triac-driver optocoupler, and finally to a high-power Triac.

be connected to a circuit that can handle the current. Power consumed by the light controller at 120 volts AC with all eight channels operating at full power is 4400 watts—that's almost 40 amps! Most household wiring is 15 or 20 amps per circuit breaker, so the AC input should be separated into multiple circuits to control that much power.

Figure 3 is the LED display section. It's used to indicate which output channel is active. It also helps you set up MIDI sequences without having to hook up any external lights.

A clean 5-volt DC powe source is required for the ligh controller. The author used self-contained 5-volt suppl that includes a built-in trans former, rectifier, and regulato The supply accepts a 120-vo AC input and outputs 5-volt DC. You can use a similar sup ply if you like, although they an

more expensive. Otherwise any 5-volt supply will do.

#### Software

Software for the light controller is interrupt-driven. Upon reception of MIDI data, an interrupt is generated. The software jumps to the interrupt routine for the serial port. First the interrupt is cleared, then the byte is placed in the receive buffer. As more data comes in, it is buffered. The keyboard also generates an interrupt any time

- All resistors are 1/4-watt, 5%, unless otherwise noted. R1, R4-220 ohms R2-280 ohms B3-10.000 ohms R5, R6, R8, R10, R12, R15, R17, R19, R22-R29-120 ohms R7, R9, R11, R13, R14, R16, R18, R20-180 ohms R21-5000 ohms, potentiometer R30-1K×8 SIP resistor Capacitors C1, C2-30 pF, mica C3-10 µF, 16 volts, electrolytic C4-10 µF, tantalum C5-1 µF, tantalum Semiconductors IC1-80C32 microcontroller IC2-74LS373 8-bit latch IC3-27C64 EPROM IC4-74LS138 3-to-8 demultiplexer IC5-74C922 16-key keypad encoder IC6-74LS00 quad NAND gate IC7-74LS244 octal buffer/line driver IC8-6N138 optocoupler IC9-IC16-MOC3010 optoisolator D1-D10-1N914 diode TR1-TR8-any 6-amp Triac in a T0-220 case LED1-LED9-Chassis-mount lightemitting diode, any color Other components XTAL1-12-MHz crystal F1-F8-5-amp fast-blo fuse P1, P2-10-pin header strip SO1, SO2-10-pin header socket J1-5-pin DIN socket Miscellaneous: inverter (for use with LCD backplane, see text), 5-volt power supply (see text), project case, grounded AC linecord, 8 grounded AC outlets, wire, solder, etc. Note: The following items are available from Audio Visual Imagery, P.O. BOX 332, Randolph, MA 02368: PC board-\$35.00 Programmed 27C64 EPROM— \$10.00
  - A diskette with the sample files will be included with any order. Please add \$1.50 S&H to any order. MA residents must add 5% sales tax.

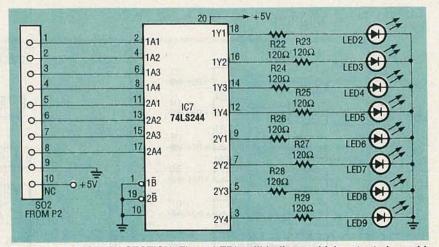


FIG. 3—LED DISPLAY SECTION. These LED's will indicate which output channel is active without having to set up the full-power lights.

a key is pressed. As in the serial section, first the interrupt is cleared.

The main monitor loop of the program simply tests to see if data is in the receive buffer or the keyboard flag is high. If data is in the buffer, the byte is checked for a high MSB indicating a status byte. If it is a status byte, the status register is updated. The status will then remain until a new status byte is received. The program then continues the main monitor loop. If the byte is data, it is examined to determine the action to be taken. Depending on the current user-set mode of the controller, a light will be turned on, a light will pulse, or nothing will happen. If the data turns out to be bad, it is flushed.

After examining the data and acting upon it, the main monitor loop is resumed. When the key flag is high, the data in the temporary buffer is examined. The action desired is stored in the appropriate register. The main monitor loop is then resumed. All user-defined actions are entered on the keyboard. As options are entered, the LCD displays the choices available. The option being defined is displayed on the top line, and the selections are cycled through. Press the key for the desired action. If more information is needed for the chosen item, it will be presented on the LCD. The source code for the EPROM and some sample files to run on a sequencer are available on the RE-BBS as a selfunarchiving zip file called MLC1.EXE. A programmed EPROM is available from the source mentioned in the Parts List.

#### Construction

The PC board for the light controller can be made using the foil patterns we've provided, or it can be purchased from the source mentioned in the Parts List. A parts-placement diagram is shown in Fig. 4. Notice the row of pads to the left of the P1 header that divides the board down the center; the board can be cut there if you want to remotely locate the power section of the board and then run lowvoltage wiring between the two sections.

If you are going to cut the board in two, do it before stuffing it, as it will be easier. Install the components according to Fig. 4, and check your work as you go. The pads marked "INV" on the board are for a voltage inverter (see parts list), if used. It that generates the proper voltages for the backlight on the LCD if the LCD you use has one. If it does have a backlight, and you wish to use it, install the inverter and connect the LCD's backlight terminals to the pads marked "BL" on the board.

Any kind of case will do for this project, as long as everything fits inside—keep in mind the power supply you will use and be sure to leave room for it. If you use a metal enclosure be

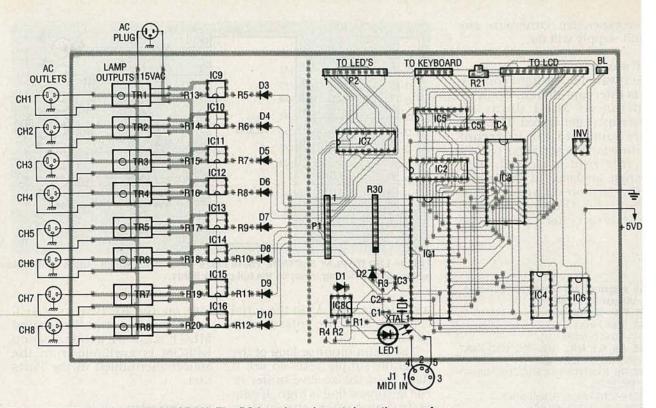


FIG. 4—PARTS-PLACEMENT DIAGRAM. The PC board can be cut down the row of pads to the left of the P1 header if you want to run low-voltage wiring between the two sections.

sure to ground it to the AC safety ground as shown in Fig. 2, and also properly ground all eight AC outlets.

As a reminder, the light controller can output about 5 amps per channel, and any wire common to all eight outputs must carry about 40 amps. The controller's internal AC wiring must therefore be chosen accordingly. Also, if your house wiring is rated 15 or 20 amps per circuit breaker, the AC input to the controller should be separated into at least two circuits. Figure 5 shows the current version of the PC board.

#### Testing

To test the light controller, simply adjust R21 to the middle of its range, and apply a clean 5volt DC power source to the circuit. A message will appear on the LCD. Adjust R21 for the best-looking display. Plug a MIDI cable into the MIDI IN jack (J1) on the light controller. The other end should be placed in the OUTPUT jack of a sequencer or the THRU jack of a synthesizer. (If you're not using a sequencer, connect it to the OUTPUT jack of the keyboard.) At power up, the light controller is on MIDI channel 1. Make sure your MIDI equipment is sending notes out on channel 1. As notes are sent, LED1 will light, indicating that data is being sent over the MIDI. As notes are received and examined, the appropriate LED(s) will light (as long as the notes are in the CONTROL OCTAVE of the light controller).

If the circuit does not work, check the supply voltage. Make sure XTAL1 is oscillating. Check for a 1-microsecond reset pulse on pin 9 of IC1 at power up. If the reset circuit does not produce a pulse, the circuit will not start up. At power up, the eight LED's should light and then extinguish. If that does not happen and the clock and reset circuits are working, check the wiring of the LED's and the Triac section. If nothing happens still, the problem is with the microcontroller. Check the wiring of the ADDRESS and DATA bus. The circuit will function without the keyboard and LCD sections. If the circuit seems to be working (responding to MIDI notes, turning on LED's) but the keyboard and/or LCD do n respond, check the wiring IC4 and IC6. Check for READ ar WRITE (RD and WR) signals fro IC1, pins 17 and 18 respective

Along with source code ar HEX data for the EPROM, sar ple songs are provided in the z file on the RE-BBS. (The samp files are also included with an order from the source me

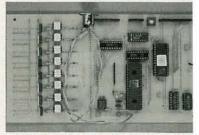


FIG. 5—THE CURRENT VERSION of t PC board, shown here with all parts stalled.

tioned in the Parts List.) Thr samples are provided in bo \*.WRK and \*.MID formats th will help you get started. TI songs were set up using Roland MT-32, Cakewalk 3. and our MIDI Light Controlle The MIDI Light Controller au knowledge of your sequence operation will provide you—au your audience—with a gre looking show.



# SOLID STATE RELAY

Why buy expensive, commercially made solid-state relays when you can make your own solid-state relays at a fraction of the price?

SOLID-STATE RELAYS MAKE IT A cinch to connect digital logic to the nasty world of 115 volts (or more). These handy little devices make it possible for a battery-operated supply to turn on 100-watt light bulbs, 10-horse power motors, a lawn sprinkling system, or almost anything else you can imagine.

Solid-state relays, or SSR's, usually consist of an optoisolator and a Triac, which is used as an AC switch. There are many SSR variations such as those using reed relays in place of the optoisolator, and those using SCR's instead of Triacs, but most consist of less than a dozen parts.

Solid-state relays can provide isolation from 2 to 7.5 kilovolts and can drive tens of amps. They're usually offered in a plastic-filled cube with a heat sink on the bottom and screws for attaching the four wires nothing could be simpler.

There are three common complaints with SSR's: First is the cost—twenty dollars apiece is about average. Second is the fact that they can't be repaired. When SSR's go bad, the fix could usually be a two-dollar Triac, but you can't get inside the plastic potting compound to repair it. Third is the fact that most SSR's made in large quantities are usually rated at about 10 amps. Who wants to pay for a ten-amp relay when all he needs

#### **RODNEY A. KREUTER**

is a two-amp relay? Besides, the larger the Triac, the larger the leakage currents.

There is hope for readers of **Radio-Electronics**, however. Using all new parts, the solidstate relay presented in this article can be built for less than eight dollars. An added bonus is that it can be repaired if anything goes wrong.

#### Operation

The basic operation of a solidstate relay is much like a switch that is controlled by an input voltage or current. That is illustrated in Fig. 1. Keep in mind that this switch can only be used for AC voltages because it will "latch up" on DC. (A Triac will turn off only when the current drops to zero.)

Our SSR circuit is shown in Fig. 2. Diode D1 provides protection in case you connect the input backward. Resistor R1 limits the input current. If you would like to use an SSR that requires a large input voltage (to increase the noise immunity) you can make R1 a large value. If you make R1 470 ohms, the relay will need about 12 volts to turn on. The power rating of R1 is a function of the maximum input voltage. For inputs up to 10 volts, a 1-watt resistor is needed.

The voltage across Q1's collector-emitter junction is almost constant (with a minimum input of three volts) at 1.75 volts (typical LED voltage) plus 0.7 volts (typical  $V_{BE}$ ), or 2.45 volts. The voltage across R1 will therefore be the input voltage minus about 2.5 volts.

The minimum input voltage needed to turn on the SSR is a function of the minimum LED current (the LED inside IC1) and R1. The minimum for the MOC3010 is 15 milliamps. That works out to an input voltage of about 4 volts using the components shown. You can reduce the minimum voltage needed by decreasing R1 or by using an optoisolator that requires less LED current. Since the LED needs about 1.75 volts across it before it begins to emit light, operation below three volts isn't practical. The maximum current through the LED is set by resistor R2.

When the voltage across R2 reaches about 0.65 volts, Q1 begins to conduct, shunting current from the LED. The result is that, although the current through R1 rises as the input voltage rises, the current through the LED stops increasing at about 15 milliamps. The minimum LED current, therefore, is not the minimum current you can pass through the LED; rather it is the minimum LED current that will operate the Triac.

Probably the most misunderstood aspect of solid-state re-

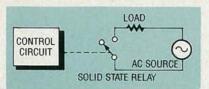


FIG. 1—A SOLID-STATE RELAY is much like a switch that is controlled by an input voltage or current.

you can use a transistor to provide a current sink as shown in Fig. 5.

#### Zero-voltage switching

Some of the newer SSR's provide zero-voltage or zero-crossing switching. In normal opera-

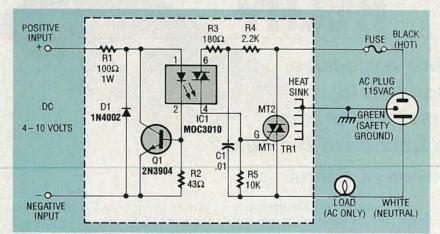


FIG. 2—SSR CIRCUIT. Diode D1 protects the unit if you connect the input backward, and R1 limits the input current.

lays is that the input requirement is really current, not voltage. That means that the driving circuit must be able to supply the current necessary to operate the LED in the SSR. In the example shown, the current is about 15 milliamps. The current can come from a current source as in Fig. 3 or a current sink as in Fig. 4. Most circuits can sink more current than they can source. For example, TTL can source only one milliamp or so, but it can sink 10 to 15 milliamps. If you must use a logic family that can't source or sink much current, such as the output of most computer ports,

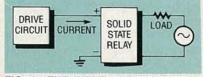


FIG. 3—THE INPUT REQUIREMENT of an SSR is current. Here the driving circuit is shown as a current source.

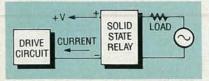


FIG. 4—THE DRIVING CIRCUIT is shown here as a current sink.

tion the trigger side of the relay is totally asynchronous to the AC side. That means that a trigger could occur during any part of the AC sine wave. If the trig-

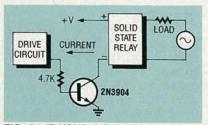


FIG. 5—IF YOUR CIRCUIT can't source or sink much current, use a transistor as a current sink.

ger occurs near a peak (90 270 degrees), a large curre will flow into the load almost i stantly. That creates a lot of F (radio frequency interference and also is very hard on the fi ment of ordinary light bulbs. order to prevent that, zen crossing SSR's accept the tr ger at any time but delay tun ing on the AC load until the net time the AC voltage pass through zero volts.

#### Safety

Building an SSR requir putting 110 volts on a print circuit board. From an electric point of view, that can be p fectly safe. However, it's prot bly a good idea to cover printed circuit runs on the 11 volt side of the PC board wi silicone sealer. Also, try to u only isolated Triacs (where t case is electrically isolated fro the Triac), and ground the he sink to the AC safety wi (green, or earth ground).

#### **Choosing a Triac**

There are three basic requi ments when choosing the or put Triac. First is to make su that it will handle the volta required. The minimum for 115-volt AC line requires a 20 volt Triac. A 220-volt line quires a 400-volt Triac. F member that those are t minimum so, for a few cer more, it pays to use the ne highest voltage rating.

The next requirement is corrent. A 6-amp Triac will handle amps only if it is properly he

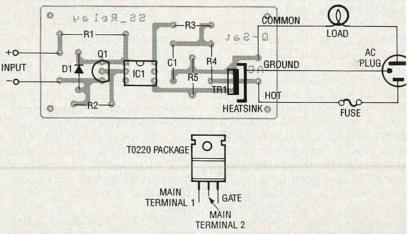


FIG. 6—PARTS-PLACEMENT DIAGRAM. You must heatsink the Triac and connect the heatsink to earth ground, even if you're using an isolated Triac.

#### PARTS LIST

All resistors are ¼-watt, 5%, unless otherwise noted.

R1-100 ohms, 1-watt (see text)

- R2-43 ohms (see text)
- R3-180 ohms
- R4-2200 ohms
- R5-10,000 ohms
- Capacitors

C1—0.01 µF, 500 volts, ceramic Semiconductors

IC1—Motorola MOC3010 Triac-output optoisolator (or MOC3011 for zero-crossing switching, see text) D1—1N4002 diode (or any 1-amp,

100-volt or greater diode)

Q1-2N3904 NPN transistor

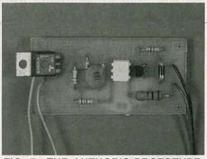


FIG. 7—THE AUTHOR'S PROTOTYPE. The board is shown here without a heatsink on the Triac, although one should definitely be used.

sinked. As a word of warning, motors draw a lot more current on start-up than they do during normal operation, sometimes as much as ten times more.

The third requirement is the gate current. The Motorola MOC3010 optoisolator will provide about 100 milliamps of drive current for the output Triac. That should be adequate TR1—Teccor Q4006L4 Triac (or equivalent, see text)

Note: The following items are available from Q-Sat, PO Box 110, Boalsburg, PA 16827:

• A printed circuit board for two solid-state relays, can be cut in two or left together (part number SSR-PCB)—\$8.00 postpaid

• A complete kit minus heat sinks for two solid-state relays (sorry, sold in multiples of two only, part number SSR-KIT)— \$18.00 postpaid

G-Sat 21/2 SS\_Relay

SSR FOIL PATTERN shown actual size.

for any Triac you can find in a TO-220 package.

Although not a strict requirement, an isolated Triac is a good safety precaution. Isolated Triacs provide electrical isolation from the electrical connections

#### TABLE 1-TYPICAL TRIAC-OUTPUT OPTOISOLATORS

Motorola Type	Blocking Voltage	LED Current (mA)	Maximum R2 (ohms)	Zero Crossing Switching	AC Line Voltage	
MOC3009	250	30	22	No	115	
MOC3010	250	15	43	No	115	
MOC3011	250	10	68	No	115	
MOC3012	250	5	130	No	115	
MOC3020	400	30	22	No	220	
MOC3021	400	15	43	No	220	
MOC3022	400	10	68	No	220	
MOC3023	400	5	130	No	220	
MOC3030	250	30	22	Yes	115	
MOC3031	250	15	43	Yes	115	
MOC3032	250	10	68	Yes	115	
MOC3040	400	30	22	Yes	220	
MOC3041	400	15	43	Yes	220	

to the case. Early Triacs were not normally isolated. That meant that you had to use mica washers and thermal grease. Thermal grease is still a good idea, but the mica washer isn't required for isolated Triacs. If you don't know whether or not your Triac is isolated, simply measure the resistance from each lead to the case. An isolated Triac will measure open on all three leads.

**Warning:** The Radio Shack 400-volt, 6-amp Triac (part number 276-1000) will work well in this circuit, but it is *not* isolated. You *must* use a TO-220 mica washer and thermal grease if you plan to use that device.

#### Construction

For a simple SSR, an optoisolator such as the Motorola MOC3010 will be sufficient. For a zero-crossing SSR, an MOC3031 will do. Many companies make optoisolators. Make sure yours has a Triac output and that the pinouts are compatible with your design. Table 1 shows some typical Triac-output optoisolator specifications.

Although the SSR can certainly be built without a printed circuit board, using the foil pattern we've provided will make building it a simple task. You can also buy a pre-made PC board from the source mentioned in the parts list. Figure 6 shows the parts-placement diagram. The only precaution, other than the one about working with 110 volts, is to heatsink the Triac. If you leave the leads on the Triac long, it should be a simple matter to find some heat sink to attach to the Triac. Just remember to connect the heatsink to an earth ground. Even if you're using an isolated Triac, the earth ground is still necessary. Otherwise you should buy your solid-state relays from a reliable company-don't build them yourself. Figure 7 shows the author's completed prototype.

Remember that SSR's can switch only an AC line. Trying to switch a DC line will result in a relay that closes but never opens. **R-E** 

#### ANTHONY J. CARISTI

IF YOU ENJOY HIKING OR DRIVING through mountainous country—but without knowing how far above sea level you are—this project will be of interest. It's a simple, easy-to-build, compact electronic altimeter that provides altitude readings from zero to 1999 feet with a resolution of 1 foot. The device has a 3<sup>1</sup>/<sub>2</sub>-digit LCD readout and is powered by a common 9-volt battery.

The altimeter's small size and light weight allows it to be easily carried wherever you travel. The modest power drain of the circuit ensures many hours of operation on a single 9-volt battery. If desired, the instrument can also be powered by the 12-volt electrical system of any vehicle.

The electronic altimeter is a pneumatically operated device that responds to absolute atmospheric air pressure. It operates on the same principle as the aneroid altimeters found in every aircraft.

Absolute atmospheric pressure is a mathematically definable parameter that varies inversely with altitude. At sea level the pressure is 14.7 pounds per square inch, and decreases as altitude increases.

The heart of the altimeter is a high-quality solid-state absolute-pressure sensor that is sensitive enough to detect changes in altitude as small as one foot. The surrounding circuitry amplifies the analog voltage output of the sensor, and converts it to digital form to drive the display.

#### The sensor

The pressure sensor was developed by Motorola Semiconductor Products, Inc., Phoenix, AZ. It's designed to respond to absolute atmospheric pressure, which is defined as pressure measured with respect to a perfect vacuum (zero pounds per square inch absolute, or 0 PSIA). Over the altimeter's range of interest, 0 to 1999 feet, the absolute atmospheric pressure varies from 14.696 PSIA to 13.67 PSIA (which corresponds to 29.92 to 27.82 inches of mercury, respectively).

Inside the sensor is a monolithic silicon piezoresistor that is ion-implanted on a thin silicon diaphragm. The pressure sensor contains two chambers separated by the silicon diaphragm. One of the sensor's chambers is exposed to atmospheric pressure by means of an external port. The other chamber is evacuated to as perfect a vacuum as possible, and sealed. That way the diaphragm of the sensor is under constant stress in accordance with the difference between atmospheric pressure on one side, and an essentially "perfect" vacuum on the other. The mechanical stress placed on the diaphragm (and the piezoelectric resistor) causes the sensor to generate an output voltage that is proportional to the applied pressure as seen by the open port of the solid-state sensor.

to 1999 feet), the sensor's or put voltage will change or about 1.4 millivolts.

A differential amplifier, co posed of three sections of LM324N quad op-amp (IC provides voltage amplification of the sensor output. The gas of the amplifier can be adjust by potentiometer R3 to allow the the normal tolerance different between different sensors. T DC voltage level at pin 8 of IC1 set to 2.5 volts when the timeter is at zero altitude (s level). At 1999 feet, the amplif output falls to about 2.4 vol

The DC voltage from the d ferential amplifier (IC1) driv an ICL7106CPL analog-to-d ital (A/D) converter (IC3). Th chip, which is used in ma commercial DMM's, conver the differential voltage from Ito digital form and drives t 3½-digit liquid crystal displ (DSP1). An external referen voltage, generated by R14–R1



Add new dimensions to your next trip to the mountains with our pocket-sized electronic altimeter.

Refer to the schematic of the altimeter circuit shown in Fig. 1. The piezoresistor within the sensor (IC4) is connected between pins 1 and 3, and is driven by the 5-volt power supply in the circuit. The taps on the resistor, connected transversely across the element, are brought out to pins 2 and 4 of the sensor.

Under normal conditions in which atmospheric pressure causes a stress on the piezoelectric resistor, the sensor differential output voltage is a finite but very small value about 20 millivolts. Over the range of interest for climbers (0 provides the proper conversion factor between the analog inp voltage and the desired digit display in feet.

As described earlier, the ouput voltage of the amplifier set tion is set to 2.5 volts by I when the altimeter is at slevel. That voltage is fed to the negative analog input of IC (pin 30). Since the altimet must read zero at sea level, the positive input of IC3 (pin 3) must see a constant 2.5-volt That way the differential inp voltage between the positi and negative inputs will be zer and the display will read 000

desired. As the elevation of the altimeter increases, the voltage fed to pin 30 decreases while the voltage at pin 31 remains constant. As a result, the net difference in voltage is detected by IC3 and converted into an increasing digital display readout of altitude.

There is one other factor that must be taken into consideration in a pneumatically sensing altimeter of this type. Changing weather conditions cause changes in barometric pressure from the standard value of 14.7 PSI (or 29.92 inches of mercury at sea level). As with any altimeter that reacts to absolute air pressure, the effect of changing barometric pressure must be canceled out.

In this project the effect of barometric pressure is canceled out by means of BARO SET potentiometer R11 which can be adjusted for about 2.4 to 2.6 volts at its wiper. Therefore, R11

R

ALTITUDE

must be set so that the voltage at pin 14 of IC1-d causes the display to indicate the correct known altitude at any reference location. Once that is done, the altimeter is calibrated for the current barometric reading but changing weather conditions can change the readings significantly. You will probably have to recalibrate the altimeter before each use, or simply set it to zero and use it to measure a relative altitude.

Power for the altimeter comes from a 9-volt battery or from a vehicle's 12-volt electrical system. Either source is fed to an AN78L05 fixed 5-volt regulator (IC2). The altimeter's modest current drain of about 6<sup>1</sup>/<sub>2</sub> millamperes allows many hours of operation from a 9-volt battery. When 12 volts is used as the power source, a resistor and diode are used to isolate the altimeter circuit from any electrical transients.

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#### Construction

The altimeter is built using two single-sided PC boards one analog and one digital.

The two PC boards can be made using the foil patterns we've provided, or they can be purchased from the source mentioned in the Parts List. Point-to-point wiring can also be used. Parts-placement diagrams for the analog and digital boards are shown in Figs. 2 and 3, respectively.

You should use sockets for the two DIP IC's, but because of the limited space in the prototype's enclosure, only low-profile sockets would fit. Do not install the IC's in the sockets until instructed to do so. Note that the altimeter's accuracy will suffer if metal-film resistors, where specified, are not used. Also

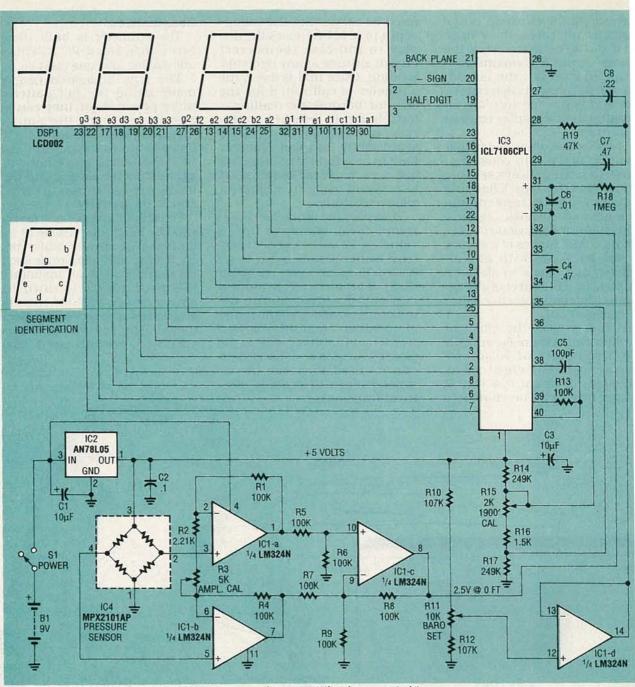
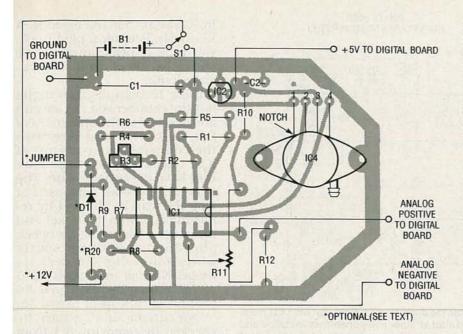


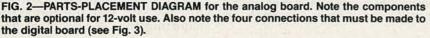
FIG. 1—THE PRESSURE SENSOR (IC4) has an analog output that is converted to digital form by IC3 and then displayed on the LCD.

note that R20 and D1 (shown on the analog board in Fig. 2) are to be installed only if you wish to power the unit from a vehicle's 12-volt source.

The pressure sensor should be handled with care. Its leads must be bent at right angles so that the sensor lies flat on the analog board. Use two longnose pliers when bending the leads—one to prevent stress on the lead where it enters the plastic body and the other to bend it. Before forming the leads, locate pin 1 of the sensor; it's identified by a notch cut into the lead. Then you'll be able to form the leads in the correct direction. Mounting hardware for the sensor is optional because the leads will hold it in place. No pneumatic connection to the sensor's pressure port is required, except calibration.

The LCD module can also be mounted in a socket if you like; you can make one for it by cutting a 40-pin DIP socket in ha lengthwise. To keep the a timeter as compact as possibl the LCD module is mounted of the copper side of the digit board, *after* all of the comp nents are installed on the corp ponent side. (You can install socket for the LCD now, b don't install the LCD just ye Mounting the LCD on the co per side allows the digital boa to be mounted on the cover the enclosure.





Obtain a clip for the 9-volt battery, or salvage a connector taken from an exhausted battery. Be sure to observe proper polarity on the battery connector. Power switch S1 and potentiometer R11 should be installed on the side of the enclosure, easily accessible to the user. If you are dealing with limited space (a small enclosure), use a miniature toggle or slide switch for S1. Once R11 is set. it's best if it can not be accidentally changed during the use of the altimeter. Therefore, R11 should be a screwdriver-adjusted part. If mounting R11 is a problem, the resistor can be epoxied in place next to the power switch. A multiturn potentiometer will make it easier to calibrate the unit.

A number of jumper wires, labelled JU1–JU20, are required on the component side of the digital board to complete the circuit. You must solder a jumper wire between each pair of pads—JU1 to JU1, JU2 to JU2, and so on. Use #24 or #26 insulated stranded wire. Examine both boards carefully for bad solder joints, shorts, and improperly installed components before continuing.

#### **Final assembly**

When both boards are com-

pleted, power, ground, and the two differential output wires from IC1-c and IC1-d must be connected between the analog and digital boards, as indicated in Figs. 2 and 3. Use insulated stranded wire for those connections and be sure to allow sufficient length to permit proper mounting in the enclosure. In the prototype, the boards are placed one above the other.

#### 12-volt power

If you want the option of powering the altimeter from a vehicle's 12-volt source, R20 and D1 must be installed on the analog board. To allow operation from either the 9-volt battery or 12-volt source, a miniature jack can be installed in the enclosure to allow connection to the vehicle's electrical system and still allow portable use. Be sure to observe proper polarity when wiring the altimeter for 12-volt operation. You can make a power cable by putting a mini plug on one end and, most likely, an automotive cigarette lighter plug on the other. Just be sure to plug the cable into the altimeter first, to avoid having a live male connector completely exposed.

The altimeter can be wired for 12-volt power exclusively, or as a combination 9- and 12-volt unit. Follow the wiring diagrams in Figs. 4 and 5 for the selected power-source option. Give the project one final visual inspection before continuing with the checkout. Figure 6 shows the author's prototype.

#### **Electrical checkout**

Before proceeding, make sure the IC's are not in the sockets. You'll need an accurate DC voltmeter with an input resistance of at least 1 megohm to perform the checkout.

Use a fresh 9-volt alkaline battery or a well-regulated DC power supply to power the circuit. If the power supply has current-limiting capability, set the limit to 10 milliamps to protect the project in the event of a malfunction. (The normal current drawn by the circuit is about 6½ millamperes.) Set the supply for 9- or 12-volts output, as applicable.

First, check voltage regulator IC2. Apply power to the circuit and measure its output voltage. Anything between 4.75 and 5.25 volts is good. In case of trouble here, check the orientation of C1 and IC2, and the polarity of the power supply. Measure the terminal voltage of the battery or supply while it is powering the circuit to be sure it is delivering at least 7 volts to the regulator. Disconnect power and measure the resistance between the 5-volt line and ground to be sure there is no short circuit. As a last resort, try a new regulator.

Check the analog circuit next. Insert IC1 into its socket, and apply power to the circuit. Measure the voltage at pin 8 of IC1; adjust R3 for a reading of 2.5 volts. Measure the voltage at pin 14 of IC1 while rotating potentiometer R11 over its range. Make sure the range of adjustment is about 2.4 to 2.6 volts. If you don't see the correct voltages, check the wiring and components associated with IC1. Check the pressure sensor for correct orientation. Try changing IC1 as a last resort.

Disconnect power from the circuit and insert IC3 into its socket. Place the readout in its socket on the solder side of the

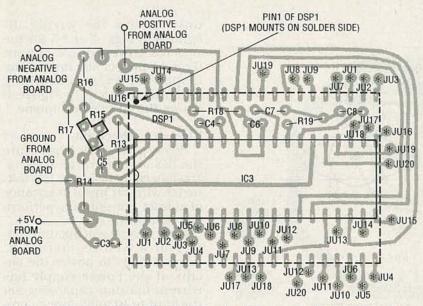


FIG. 3—PARTS-PLACEMENT DIAGRAM for the digital board. Note that the LCD module mounts on the solder side of the board after all other components are installed *and* the board is tested (see text for details). Also note the twenty pairs of pads labelled JU1–JU20. You must solder a jumper wire between each pair of pads; for example, JU1 to JU1, JU2 to JU2, and so on. by IC3 at pin 1 of the readout. normal indication at pin 1 is a volt peak-to-peak square wa with a period of about 140 n croseconds.

If the readout displays digi but the numbers do not ascer and descend with adjustment R11, the fault is most likely wi the components or wiring IC3. Check all parts to be su they have the correct value. T reference network composed R14-R17 can be checked by 1 moving IC3 from its socket ai measuring the voltage betwee pins 35 and 36 of the sock while adjusting R15 over i range. Normal indication is to 35 millivolts with pin ( positive with respect to pin 3

Operation of IC3 can l checked by removing IC1 fro its socket and temporari shorting pins 8 and 14 of I(

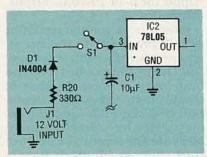


FIG. 4—WIRE THE ALTIMETER as shown here for 12-volt power exclusively.

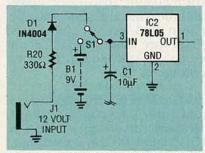


FIG. 5—THE ALTIMETER IS SHOWN wired for both 9- and 12-volt operation.

digital board in accordance with Figs. 3 and 7. If you didn't use a socket for the display, position it so that its terminals are flush with the component side of the board and solder.

With the LCD installed, set R15 to mid position, apply power, and adjust R11 over its range. The display should indicate some number that can be varied via R11 between approxi-



FIG. 6—HERE'S THE AUTHOR'S PROTOTYPE. Notice the four wires that connect t analog and digital boards together.

mately -1000 and +1000.

If the display is completely blank, check the orientation of the LCD module and IC3 to be sure that they have not been placed backwards in the circuit. If available, an oscilloscope can be used to verify the presence of the backplane signal generated socket. That causes the di ferential input voltage fed to th A/D chip to be zero, and the di play should read 000.

Once the display is operatir properly, operation of the a timeter can be verified befor final calibration. Set R15 to m position and adjust R11 for

Radio-Electronics, May 1992

lisplay of about 20 or 30 feet.

Physically move the altimeter nigher and lower to observe the change in altitude reading. Hold the project horizontally or vertically as you make this test; a change from one orientation to the other can cause the readout to vary 3 or 4 feet, due to gravitational force on the extremely sensitive solid-state pressure sensor.

You should be able to detect and resolve 1 or 2 feet of vertical displacement. It is normal for the display to fluctuate 1 or 2 digits. Additionally, it should be noted that the altimeter is a pressure-sensitive device and will respond to any variation in barometric pressure, in which a change in pressure of only 0.001 inch of mercury at sea level will

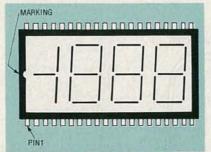


FIG. 7—THE LCD MODULE has a marking similar to that on an IC, indicating where pin 1 is. Note that the module mounts on the solder side of the digital board, so be careful where pin 1 goes.

cause a change of 1 foot in the altitude reading.

Even though the resolution of the instrument is 1 foot, the altimeter will not have either the accuracy or stability to indicate altitude that accurately.

#### PARTS LIST

#### All resistors are ¼-watt, 5%, unless otherwise noted.

- R1, R4-R9-100,000 ohms, 1% metal film
- R2—2210 ohms, 1% metal film R3—5000 ohms, cermet PC-mount
- potentiometer
- R10, R12—107,000 ohms, 1% metal film
- R11—10,000 ohms, PC-mount potentiometer, screwdriver adjust
- R13-100,000 ohms
- R14, R17-249,000, 1% metal film R15-2000 ohms, cermet PC-
- mount potentiometer
- R16-1500 ohms, 1% metal film
- R18-1 megohm
- R19-47,000 ohms
- R20—330 ohms (optional for 12volt operation)

#### Capacitors

- C1—10 μF, 25 volts, axial electrolytic
- C2-0.1  $\mu\text{F},$  50 volts, ceramic disc C3-10  $\mu\text{F},$  25 volts, radial elec-
- trolytic C4, C7-0.47 μF, 50 volts, metal
- film C5—100 pF, 50 volts, ceramic disc
- C6—0.01  $\mu$ F, 50 volts, ceramic disc C8—0.22  $\mu$ F, 50 volts, metal film

#### Semiconductors

- IC1-LM324N quad op-amp
- IC2-AN78L05 5-volt regulator
- IC3—ICL7106CPL A/D converter (Intersil)
- IC4—MPX2101AP semiconductor

- pressure sensor, 15 PSI absolute (Motorola)
- D1—1N4004 silicon diode (optional for 12-volt operation)
- DSP1—3½-digit LCD module (DigiKey LCD002)

#### Other components

- J1—Miniature jack (optional for 12volt operation—also requires matching plug)
- S1—SPDT minature slide or toggle switch
- B1—9-volt transistor battery, alkaline or heavy duty
- Miscellaneous: 9-volt battery clip, IC sockets, #24 gauge stranded hookup wire, enclosure, hardware, plastic tubing, tee fitting, clamp, soda or wine bottle, food coloring, solder, etc.
- Note: The following parts are available from A. Caristi, 69 White Pond Road, Waldwick, NJ 07463:

• Set of 2 etched and drilled PC boards (analog and digital)—\$19.95

- LM324N op-amp (IC1)—\$2.00
   AN78L05 regulator (IC2)—
- \$1.50
- ICL7106CPL A/D converter (IC3)—\$17.50
- MPX2101AP pressure sensor (IC4)—\$39.50
- Set of 13 metal-film resistors—\$4.95
- Please add \$2.75 postage/handling.

#### **Final calibration**

To perform the final circuit adjustment, you will need a tape measure, three lengths of <sup>1</sup>/4inch outside diameter clear plastic tubing, a tee for a threeway connection, and a clear plastic or glass bottle of clean drinking water. A soda or wine bottle is a good choice. Be sure that the inner diameter of the plastic tubing allows it to be stretched over the pressure port of the sensor for an air-tight seal. The suggested test setup is shown in Fig. 8.

In this procedure, the absolute pressure that represents 1900 feet of altitude will be simulated by drawing a partial vacuum to cause a column of water to rise in the straight, clear tube. This procedure uses basic laws of physics to set a desired pressure differential.

As noted earlier, absolute air pressure, under standard conditions, is 14.696 PSI at zero altitude (sea level). At 1900 feet, the pressure falls to 13.716 PSI. The pressure levels can also be specified in other units such as inches of mercury or inches of water. In this case, the desired pressure differential between zero altitude and 1900 feet, 0.98 PSI, is equivalent to 27.13 inches of water, rounded out to 27<sup>1</sup>/<sub>8</sub> inches. Thus, to simulate the change in pressure from zero to 1900 feet, a column of water 271/8 inches high can be used.

Set up the altimeter and apparatus as shown in Fig. 8. In this test, the water in the vertical tube is drawn up to the required height of 27<sup>1</sup>/<sub>8</sub> inches by drawing a vacuum at the open end of the tubing.

With no vacuum applied to the open end of the plastic tube and the pressure sensor connected as shown in Fig. 8, turn on the altimeter and allow a minute for the circuit to stabilize. Adjust R11 for an altitude reading of some small positive number, such as 50 feet or so. The number selected is not significant.

Now, gently draw vacuum at the open end of the tubing so that the column of water rises  $27\frac{1}{8}$  inches above the level of

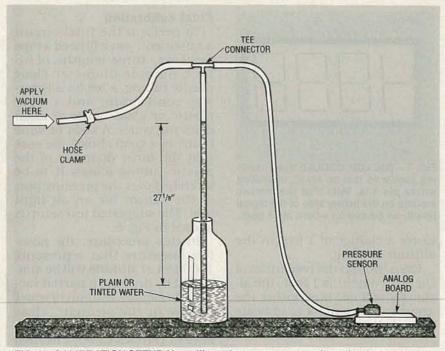


FIG. 8—CALIBRATION SETUP. You will need a tape measure, clear plastic tubing, a tee connector, and a clear bottle with water in it. Food coloring will make the water easier to see. See text for details.

water in the bottle. You can clamp the hose to maintain the required vacuum. Make sure you have a solid column of water with no air bubbles in it. Allow the display to reach a stable level and note the reading. Adjust

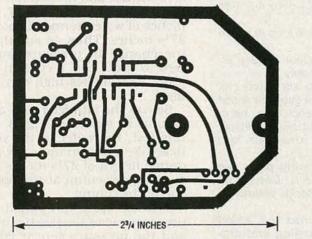
#### Using the altimeter

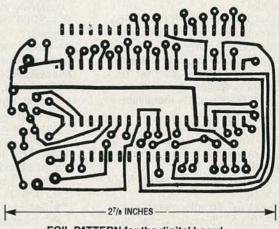
Since the altimeter is subject to the variations of barometric readings, the instrument must always be corrected for the current barometric pressure before starting out on an excursion. such as an airport where the titude is documented, setti the altimeter to reflect the c rect altitude, and then retu ing home immediately. T reading of the altimeter at yo home will then be your ref ence altitude. For best accura do this during a period of stea barometric pressure con tions.

For future use, before leavi home with the altimeter, sim turn it on and allow a minute so for the display to stabili Then set R11 so that the readi is equal to the reference altitu at your home.

Once set, do not readjust F since you will have no referer against which to adjust it. A later time, if you travel to a other location with a known titude, R11 can be readjusted reflect a more accurate altitu if the barometric reading h changed since the original s ting of R11.

When taking an altitude reaing, turn the unit on and all about a minute for the circuit stabilize. Remember, if there a sudden change in air pressusuch as might occur with changing weather, the readi





FOIL PATTERN for the analog board.

R15 so that the display indicates 1900 feet more than the original reading (50 feet) set previously.

Repeat the above test as necessary until you are satisfied that the change in altimeter reading is as close to 1900 feet as possible when the water column is drawn up 27<sup>1</sup>/<sub>8</sub> inches. This completes calibration. You do not have to know the actual barometric reading to do this. Simply turn the unit on, allow a minute for the circuit to stabilize, and set R11 to obtain a display that represents the known existing altitude at your location.

Learning the true altitude of your home can be done by taking the altimeter to a location

FOIL PATTERN for the digital board.

can fluctuate. Wait for the d play to settle down.

Be sure to turn off your b tery-powered altimeter wh not taking altitude reading That will conserve power a provide extremely long batter life, which should be about or 15 hours of altimeter oper ing time when using an alkali battery.

# BUILD THIS MICROPROCESSOR DEVELOPMENT SYSTEM

# Construction details for our inexpensive 1802 microprocessor development system.

LAST TIME WE DISCUSSED THE CIRcuitry for our 1802 development system and described how the software functions. This time we'll talk about construction and operation.

#### Construction

The complete unit uses three PC boards, corresponding to the sections of the circuit (main, keypad/display, EPROM). Foil patterns are provided if you want to make your own boards; boards and kits are also available commercially (see the parts list).

The main chassis measures  $8'' \times 4.6'' \times 1.5''$ . As shown in Fig. 7, S1 (reset) and S2 (EPROM power) mount on top of the chassis, as does a four-connector terminal block that brings several voltage sources out of the chassis for use by experimental circuits (developed on the breadboard). In addition, there is space for two 63-row solderless breadboards, and two 63-row power buses. Further, the rear edge of the case is slotted to allow the pins of P3 to protrude.

The power supply enters one

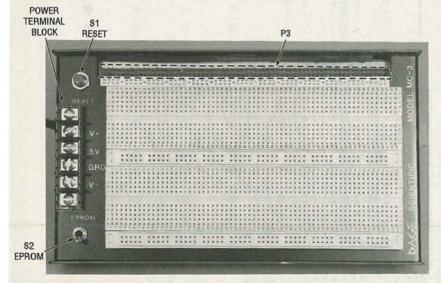


FIG. 7—MAIN CHASSIS ASSEMBLY. The terminal block on the left delivers power to breadboard circuits. Note that P3 consists of separate wire-wrap pins that protrude through a slot in the case.

side of the chassis through a grommet; the 6-wire telephone jack for the keypad/display unit fits in a slot on the other side.

#### Main board

Mount all parts on the main PC board, as shown in Fig. 8. Resistors R13-R24 must be 1/8watt units in order to mount on 0.3" centers. All other resistors mount on 0.4" centers. Sockets should be used in all IC positions, and are required for IC20 (the EPROM burner slot) and IC22 (the EPROM that contains the operating system). You can buy a pre-programmed EPROM (see the ordering information in the parts list for details) or burn your own using the hex dump shown in Listing 1.

The operating system requires the first output port, IC2. The other output ports can be installed during assembly, or as the need arises. In addition, you can eliminate IC3–IC13 if you don't need parallel inputs. The author recommends that you install at least two output ports (IC2 and IC3) and two input ports (IC8 and IC9).

You must install the operating-system EPROM at IC22 (0000h), and 8K of RAM at IC19 (E000h). You needn't install components at IC20 and IC21 unless you need additional memory.

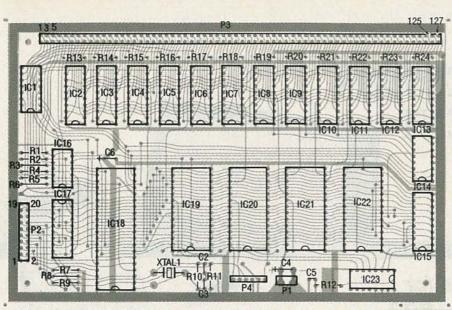


FIG. 8—MAIN PC BOARD. Note that R13–R24 must be ½-watt units to fit in the available space. Sockets are required for IC20 and IC22, and optional but recommended elsewhere.

The bus connector (P3) consists of 128 individual wirewrap pins, each measuring 0.075". The best way to install them is to insert them through the board and into a female header to hold them perpendicular while soldering. Figure 9 details the function of each pin.

Connect one wire from ground to the reset switch, and another to the pad marked *reset* on the main board. Figure 10 shows the completed main board.

#### Keypad/display assembly

Assemble the keypad/display unit as shown in Fig. 11. Mount the IC's without sockets, as there is not enough clearance to use them. However, mount each display using half a socket under the rear row of pins only. Doing so angles the display about 20 degrees for better viewing. The pull-down resistors for the key switches must be ½watt units to fit the 0.3-inch mounting centers.

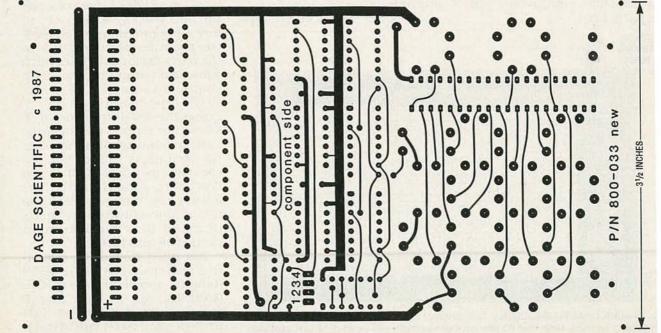
The six-conductor telephone cord connects directly to the foil

side of the board, as shown Fig. 12; secure the cord with nylon wire wrap. The other er of the cord has a modular plu that fits into J1 on the ma board. The color codes in phor cords and connectors seem vary, so we haven't provided sp cific details. It doesn't matt which color you use for which signal; just make sure th you're consistent at both enof the cable. Figure 13 shows th completed keypad/displa board.

#### **EPROM** board assembly

Assemble the EPROM boar as shown in Fig. 14. Mount six-pin female connector (J1) c the solder side of the board: will mate with P4 on the ma board, and serves to hold th EPROM board in place. Th completed EPROM board shown in Fig. 15. Whe mounted properly, the EPRO board rides about 1/2" above th EPROM that is being pro grammed (see Fig. 16). In from of this connector are two sold pads used to connect th EPROM programming voltag Connect the ground side (gno only if the programming voltag doesn't have a common grour. with the main board.

In case you want to insta RAM in IC20, remove th EPROM circuit; otherwise even

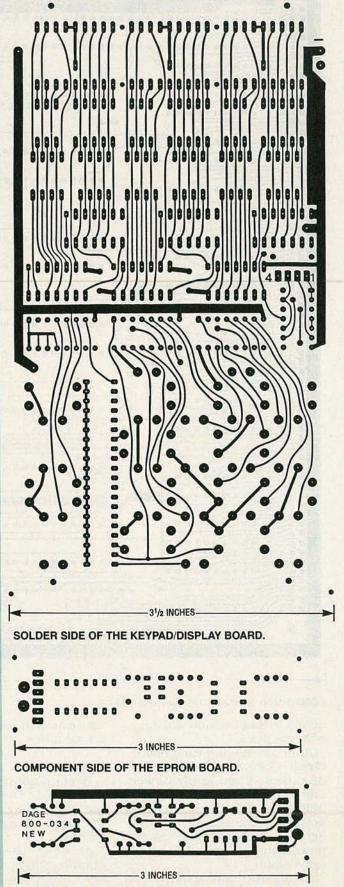


COMPONENT SIDE OF THE KEYPAD/DISPLAY BOARD.

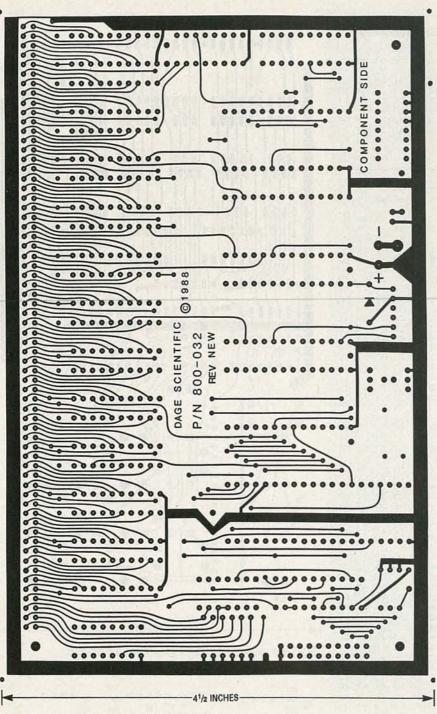
58

Radio-Electronics, May 1992

ISTIN	G1-	HEXC	DUMP	OFC	PE	RAT		S	ST	EM	(000	0-0	04	TF)
0000 0010 0020 0040 0050 0060 0070 0080 0090 0080 0080 0080 0080 008	71 00 F8 00 91 A8 062 22 30 22 E2 96 96 B3 21 71 D3 7E 7A 30 46 B9 BA 30 97 52 73 88 B7 42	F8         B4         B4	00 B3 F8 60 3A E3 FA F0 06 A7 FF FF 86 73 A3 E2 19 4D 7B 7A 99 86 A9 46 A9 46 A9 46 A9 46 A9 46 A9 48 59 60 60 9A 73 42 B8	F8 A4 62 C2 F8 FF 93 12 89 7B 89 7B 89 7B 89 7B 89 7B 89 7B 89 28 80 28	09 F8 00 00 FF B6 72 81 7A 46 27 73	A3 00 62 24 52 FF 83 A6 3D 7B D5 32 87 22 98	D3 B5 01 30 7B FF A6 F0 01 7A FF C3 3A 02 73	F8 F8 E2 52 61 FF 46 B6 0D 7B E9 C3 B7 87	FE 70 08 C0 08 FF B3 30 05 7A 00 08 04 22 73	B2 A5 69 04 22 FF 46 6F C1 7B CE 28 00 22 97	F8 F8 F8 F8 F8 F8 F8 F8 F7 C0 A3 FF A3 FF A3 7A 46 73 E22 73	FF 00 FF 87 0 FF 87 FF 87 0 FF 87 0 FF 7 FF 7	A2 B8 3A 00 3A 00	E2 F8 3D 52 45 D3 FF 7B 87 7B 3A 73 99 260
100 110 120 130 150 150 150 150 150 150 150 150 150 15	60 60 88 73 3A 18 3A 22 F8 10 01 52 B8 72 88 73 58 18 D4 02 58 18 BA 72 E6 72 22 F8	98 F8 D8 B7 2 62 A 78 2 A8 9 98 3 99 2 09 3 89 4 D4 2 B8 2 B9	73 97 73 F8 00 52 20 58 72 97 72 22 08 72 87 72 87 72 87 72 87 752 F8 76 F6 98 72 87 75 78 76 76 88 70 20 70 20 70 70 20 70 70 70 70 70 70 70 70 70 70 70 70 70	00 62 FF 3A 69 FB F0 FF F6 18	B8 22 3A 43 FF 7 BF 6 99 58 7 8 7 8 7 8 7 8 7 8 7 8 7 8 7 8 7 8	F8 F8 22 F8 FF FA D5 F8 BA F6	87 91 01 08 00 3A 0F F6 F6 D4 F6 FF 09 88 88 27	A8 52 69 AA 78 32 33 A8 02 F6 01	F8 62 F8 F8 47 6E F8 09	22 FF 00 08 F9 1A 05 9A BA 18	B7 D8 FA 52 AA 80 30 58 58 D4 09 09 89	A7 69 0F 2D8 A8 18 18 26 A3 00 73 00	97 27 FB 3A 22 69 60 FF 99 90 F6 58 99 52 72	73 97 FF 22 F8 FB 72 FF DF BA 9A F6 09 73 62 B9
2200 2210 2220 2230 2240 2250 2260 2270 2280 2290 2280 2290 2200 2290 2200 2290 2200	72 A9 F8 80 B8 F0 D4 00 B9 27 56 F0 D6 F1 A8 02 F8 02 73 90 F8 02 73 40 73 40 72 77 73 40 72 77 93 7	A8         A8         A8         A8         C         A8	B8         72           9A         FA           D5         F8           D5         F8           3A         3B           AO         300           D4         01           FE         FE           09         9A           24         02           F8         C8           E9         F8           240         73           C5         F8           73         94	A8 OF 00 FE 89 24 OC FE 58 59 A1 D1 52 40 01 73	F0 32 93 F1 60 52 09 87 30 73 73 85	1D A9 4A A9	D5 FF D4 F8 30 A3 FE F1 8A 3A B8 79 52 E2 10 95	88 01 04 28 04 33 52 F6 59 F8 73 F8 92 80 73	73 18 80 A7 88 01 88 08 33 19 FA FF B0 73 86	98 30 D4 89 F6 80 F8 28 94 38 82 80 82 90 73	52 16 01 FE 33 09 FF 58 F6 29 E8 73 F8 A0 73 96	08 D6 A9 5C	00 BA FF 99 F6 D4 F8 02 9A 59 73 24 A0 F8 83 78	B8 72 F0 7E 33 01 F7 22 F6 FF 49 FF F0 00 73 73
0300 0310 0320 0340 0350 0360 0370 0380 0380 0380 0320 0320 0320 0320 0350	97 7. 98 7. 97 7 AA 8 B7 3 01 8 F6 3 89 F6 3 89 F7 4 80 7 80 7 7 AA 8 80 F7 7 AA 8 80 F7 7 80 7 80 7 7 7 AA 8 80 F7 7 80 7 80 7 7 80 7 7 80 7 7 80 7 7 80 7 80	3       8C         3       F8         7       30         2       49         0       D4         3       66         E       A9         0       D4         F       1F         7       20         0       AD         0       A9         0       A5         E       C4	CO 00	73 F8 01 2A FF 09 FF 30 3A 40 40 F6	D4 27	73 73 8A 30 01 87 30 FF 40 40 40 22 9B	99 9D 3FE 41 34 0C 3A 66 F8 BB BB B7 B3 C0 75	100 Carlot 1.	8A 8E 59 76 76 8A 76 8A 8A 8A 8A 8A 8A 8A 8A 8A 8A 8A 8A 8A	73 73 90 2F 33 33 52 3B F8 40 40 40 FF B7	9A 9E BF 98 89 93 95 BE BA 86 B2 FF 17	73 73 0F 80 89 F6 F8 F1 89 A1 40 40 FF F8	88 87 17 33 04 49 57 D1 AE AA A6 A2 46 00	A7 D4 2F F8 40
0400 0410 0420 0430 0440 0450 0460 0480 0480 0480 0480 0480 0480 048	A9 2 32 6 66 1 FF F 27 1 00 D 09 7 A1 F 22 6 F8 0 3E D	0 03 6 F6 A 5E 6 9B A 0B B 1C 0 D5 7 27	97 37 8C A5 F6 F6 D4 01 75 B5 5C D4 29 30 46 B5 C4 C4 0C C0 40 B5 B2 A2 7B 77 F8 95 52 65	8D 8D 8D 19 19 4C 46 3A 02 57 12 6F 88 F8	87 B9 D4 FF 16 80 9C A7 77 24 F8 E2 67 E3 91 61	32 D4 02 F0 16 D4 B9 27 D5 FF 80 69 66 78 A8 22	13 01 09 D5 46 01 8C 17 D4 D4 B7 22 65 62 E3 D8	8C 80 9A 46 BC D6 A9 27 04 03 07 6A 64 01 62 3F	F4 F8 5E BB 46 FF D4 87 39 EE 7A 22 63 62 00 E8	AC FF 1E 46 AC F0 01 C4 04 00 F8 6B 62 00 62 FF	C7 BE 9D AB 99 EC 80 C4 88 00 C0 22 61 61 01 FF	1D F8 BA 16 3A 0B D4 3A 04 00 B7 6C 222 55 E2 FF	C4 F0 D4 E6 52 F3 01 77 A0 000 57 22 3D B8 FF	E2 69



SOLDER SIDE OF THE EPROM BOARD.



#### COMPONENT SIDE OF THE MAIN PC BOARD.

access to that location will incur a 50-ms delay.

After assembling each board, check all work, and correct any mistakes. Then apply power, and hold down the 0 key. If all is well, all segments and decimal points of the display should light up. If they do not, remove power and check all connections again.

Electronic construction is complete; now you can mount the boards in their proper chassis locations.

#### Operation

Boot up normally; the display should read "A-0000." The "A" indicates Address Select mode; the zeros indicate the current address.

Actually, the monitor program has four modes: Address Select, Memory Monitor, Run, and Debug.

#### PARTS LIST-MAIN BOARD

All resistors are ¼-watt, 5%, unles otherwise noted R1, R3-R8, R11, R12-1000 ohms R2-150,000 ohms R9-30.000 ohms R10-22 megohms R13-R24-51,000 ohms, 1/8 watt Capacitors C1-1 µF, 35 volts, tantalum C2, C3-20 pF, ceramic C4-10 µF, 25 volts, tantalum C5, C6-0.1 µF, mini ceramic Semiconductors IC1-74HC238 3-to-8 line decoder IC2-IC13-74HC373 octal D latch IC14-74HC138 3-to-8 line decoder IC15-74HC373 octal D latch IC16-74HC86 quad 2-input XOR gate IC17-74HC299 8-bit shift register IC18-1802 microprocessor IC19-6264 static RAM IC20-see text IC21-see text IC22-2764 EPROM (with operating system) IC23-4556 dual 1-of-4 decoder Other components XTAL1-2.010 MHz crystal P1-P4-wire-wrap pins, 0.025" square × 0.75"

J1-6-conductor telephone jack

#### PARTS LIST—KEYPAD/DISPLAY BOARD

All resistors are ¼-watt, 5%, unless otherwise noted R1–R20—51,000 ohms, ¼-watt R21–R68—330 ohms R69—100,000 ohms Semiconductors IC1–IC6—74HC164 8-bit shift register IC7—74HC00 quad 2-input NAND gate IC8–IC10—4021 8-bit shift register Other components DS1–DS3—dual 7-segment LED display, 0.5″, common anode S1–S20—SPST, normally open, pushbutton, PC mount

When the display shows "Athe monitor is in the Addres Select mode. Any time the ope ating system is in contro pressing F4 returns you to Ac dress Select mode.

To enter a new address, jus press the corresponding keys The digits you enter scroll fror right to left; if you make a mis take, simply enter new digit until you see correct addres displayed.

After entering the desired ac dress, you have three choices with corresponding keys

#### PARTS LIST-EPROM BOARD

All resistors are ¼-watt, 5%, unless otherwise noted

R1, R4-22 megohms R2-47.000 ohms R3-100,000 ohms Capacitors C1-0.001 µF, Mylar C2-100 pF, ceramic C3-0.001 µF, Mylar C4-0.02 µF, 5%, Mylar C5-0.1 µF, ceramic Semiconductors IC1-74HC02 quad 2-input NOR gate IC2-555 timer D1-1N4148 diode Q1, Q2-2N4124 NPN transistor Miscellaneous: Chassis & hardware. power supply, telephone cord & connectors, terminal block, toggle switch,

push button switch, solderless breadboarding connectors, PC boards. Note: The following items are available from Dage Scientific, 6124 Bald-

win St., Valley Springs, CA 95252 (209) 772-2076: • Kit including everything but power

supply (Model MC-2)-\$195

Surplus power supply (+12, +5, -5)—\$11

Operating system in EPROM—\$10
 Set of 3 PC boards and manual—
\$35

Please add \$5 shipping & handling per order. California residents add applicable sales tax.

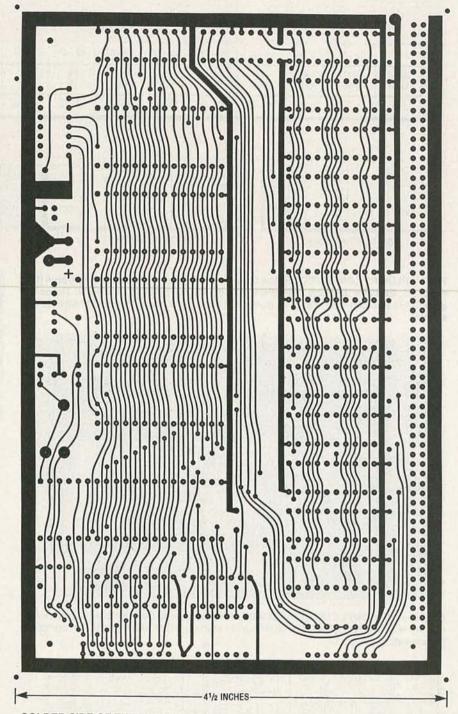
Monitor (F1), Run (F2), and Debug (F3).

Monitor mode allows you to examine and modify memory contents on a byte-by-byte basis. Run mode lets the CPU execute the program at the current address. Debug mode inserts a breakpoint at the current address.

To enter Monitor mode, type the desired address and press F1. That address and its contents will appear. For example, 0000.FF indicates a hexadecimal value of FF at address 0000h.

In this mode, the function keys take on new meanings. F1 stores the currently displayed value into the currently displayed address and moves on to the next address. F2 displays the next address. F3 displays the previous address.

To change the currently displayed value, use the hex keys to roll new digits into positions 5 and 6. If you make a mistake, simply enter new digits until



#### SOLDER SIDE OF THE MAIN PC BOARD.

the correct value appears. Memory contents will not be altered until you press F1. When you do press F1, the currently displayed value will be stored at the displayed address, and the next address will be displayed. If the value can not be stored into memory, the address counter will not increment. (It's possible to program values one byte at a time into an EPROM using that procedure, but there's a better way, as discussed below.) And remember: Press F4 at any time to return to Address Select mode.

After storing a program in memory, you can execute it using the Run command. Starting from Address Select mode, enter the desired starting address and press F2. The monitor program then transfers control to your program. If your program hangs, press the

1-7	9-17	19-2	27 29-	- 37	39-47	49-57	59-67	69-77	79-87	89-97	99-107	109-17	119-27
SER. FLAG	#2 0UT	#3 0U		4 UT	#5 OUT	#6 OUT	#7 OUT	#A IN	#B IN	#C IN	#D IN	#E IN	#F IN
2-8	10-18	20-	28 30-	- 38	40-48	50-58	60-68	70-78	80-88	90-98	100-08	110-18	120-2
					đ								
	1		SERIAL		ATA L	n D3	D5	D7	• 5	тен Тр	1 D3	D5	D7

FIG. 9—PLUG P3 CONNECTIONS. The 128 pins of P3 consist of one group of 8 pins (for serial I/O, EF flags, Q clock, and interrupt) and 12 groups of 10 pins each. Those 12 groups break down into six input ports and six output ports, each with pinouts as shown.

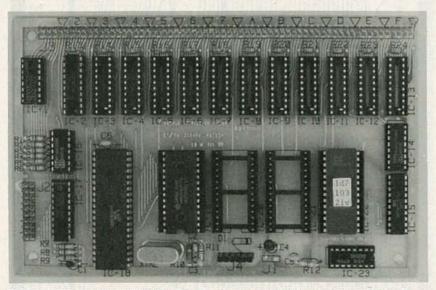


FIG. 10—THE COMPLETED MAIN BOARD. Sockets should be used in all IC positions, and are required for IC20 (the EPROM burner slot) and IC22 (the EPROM that contains the operating system).

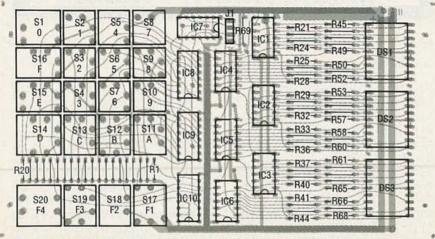


FIG. 11—KEYPAD/DISPLAY PC BOARD. Mount all parts as shown here. If you use our case, don't use IC sockets except under the rear row of display pins.

FIG. 12—PHONE CABLE connectio Solder the wires directly to the foil s of the board.

reset switch to regain monit control.

In case your program does work the first time, you can u Debug mode to track dov problems. Use Address Sele mode to select a likely addre for troubleshooting and pre F3. You'll return to Address S lect mode. Now enter the des ed starting address and pre F2. Later, when the CPU hits t breakpoint address, it will sta executing a special debug pi gram that allows you to view t CPU's internal registers, and verify that what you intended happen is indeed happening.

You can set only one bre point at a time; you cann breakpoint addresses in RO. When your program reaches t breakpoint, it will halt and d play the current address. Ye are now in the Debug mode.

In Debug mode, the displ appears the same as in Monit mode. However, as you press t hex keys the display will she the internal register number ( positions 1–4) and the value

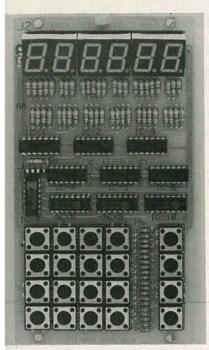


FIG. 13—COMPLETED KEYPAD/DISplay board. Mount each display using half a socket under the rear row of pins only, to provide better viewing.

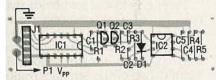


FIG. 14—EPROM PC BOARD. Mount all parts except J1 on the component side of the board; mount J1 on the foil side.



FIG. 15—THE COMPLETED EPROM board. A six-pin female connector (J1) on the solder side mates with P4 on the main board.

that register (in positions 5–6).

The debug program uses registers 0 and 1, which have been reserved for DMA and interrupt. Pressing hex key 0 displays the contents of the D register in positions 1 & 2, the X register in position 3, and the P register in position 4. Pressing hex key 1 displays Q in position 2 (set = 1, reset = 0) and DF in position 4.

While in debug mode, register contents can be altered by first selecting a register pair and then pressing function key F3. Change the value by rolling new digits in from right to left. When the correct value appears, press F1; otherwise press F3 to back out without changing the current register. Registers D, X, P, Q, and DF can also be modified by selecting hex keys 0 and 1 as described above. To exit debug mode and continue execution, press F2. Of course you can press F4 to return to Address Select mode.

The debug breakpoint alters program memory by replacing three bytes at the selected address. When the user program reaches the breakpoint address, the debug program takes over and restores the original three bytes to the proper locations. However, if the user program never reaches the breakpoint, those three bytes will never be restored. In that case you must restore them either by continuing execution at the breakpoint, or by reentering the bytes manually using Monitor mode. If you continue at the break-

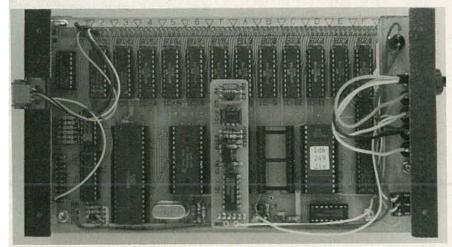


FIG. 16—THE EPROM BOARD mounts on the main board and rides about  $\frac{1}{2}$ " above the EPROM that is being programmed.

Address	Memory Contents
FF03	Start address (hi)
FF04	Start address (lo)
FF05	End address (hi)
FF06	End address (lo)
FF07	Destination address (h
FF08	Destination address (Id

point, the debug program will restore the three bytes and immediately jump into Debug mode. As usual, you can modify registers, continue execution, or return to Address Select mode.

#### **EPROM** programming

With the EPROM programming board connected to J4 and the proper programming voltage available, flip the EPROM switch to on, and you are ready to program the EPROM mounted at IC20. All that is required to program a location is to "write" to it. As mentioned earlier, you can do this byte at a time using the Monitor mode. However, due to the error-prone nature of that procedure, the author recommends a more automated procedure.

The preferred method is to enter your program in RAM and then transfer it to EPROM with the operating system's built-in "move" utility, which in fact will move a block of data anywhere in memory, not just to EPROM. Start the utility by running at 0488. Doing so transfers the move utility itself to RAM starting at FF00. Now enter the start, end, and destination addresses as shown in Table 1.

Double-check your values to ensure that they are correct, and then run at FF00. The display will show the remaining number of bytes to be transferred. It will be changing rapidly, but will at least give some idea about how things are progressing. In case data cannot be transferred correctly, the program will terminate and the display will show the address that didn't change.

That about wraps things up. Actually, now that the hardware's built, the real fun is just about to begin. **R-E** 

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Stop throwing away those used Polaro batteries—use them in your electronic experiments instead!

#### MARC SPIWAK

DID YOU EVER NOTICE HOW POLARoid cameras never need batteries? That's because every pack of film you buy contains a brand-new Polaroid Polapulse battery to power the camera's motor and flash. Some people might consider that a modern convenience: Every change of film also replaces the battery. The worry of dead batteries is over, once and for all-or is it? While these flat 6-volt cells certainly make instant-picture taking as convenient as possible, they can be harsh on our environment, or at least our landfills. To add insult to injury, the Polapulse battery that you throw out with your old film pack is still in pretty good shape.

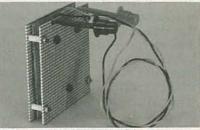


FIG. 1—THE BATTERY JIG lets you recycle "old" Polapulse batteries.

Most people, however, would have no use whatsoever for these used but still perfectly good batteries. The batteries, which can be accessed only by breaking open the film holder, don't fit into any standard battery holder. But readers of this magazine should certainly be able to find hundreds of uses for these "recycled" 6-volt cells.

#### Using the Polapulse.

To get at the battery, all you have to do is crack open the plastic film holder—after the film is finished, of course. Just be careful of the metal leafspring's sharp edges. You'll find the battery itself mounted on a piece of cardboard with its two contacts exposed on one side. Holding the battery with contacts nearest to the t positive is on the right. You remove the battery from cardboard mount if you'd lik work with a smaller over package.

The one problem with us the batteries is that it's hard make reliable electrical cont with the battery's positive a negative terminals. It's virtu impossible—as well as unsaf to solder to the terminals. Th where our handy battery comes in. It allows you to us Polapulse battery for whate the need may be, and pop i new one when it's dead.

The jig is fashioned from a pieces of perforated construction board, four screated and spacers, and some contact springs from a sacrificed AA-holder. Two of the springs maselectrical contact with the tatery, and a third spring helps hold the battery in place. Mathematical springs, and smithed the springs and smithed the springs and smithed the springs and smithed the springs the fixture.



FIG. 2—THE BATTERY JIG uses pieces of perforated construct board, screws, spacers, and spri from an AA-cell holder. Nylon scr hold the contact springs to the fixtur

You can, of course, use ligator clips or other connects that better suit your needs. A you may wish to modify the r chanical construction as we In any case, do try to get t most use possible from the batteries before they are throaway.

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Signed:

N OUR LAST INSTALLMENT WE ooked closely at theory and practical applications of the LM3914-series of LED bargraph driver IC's and concluded with 7-segment display systems. In this article, we'll examine 7-segment display driving techniques in detail, concentrating on decoder/driver devices and circuitry.

#### **Display** latching

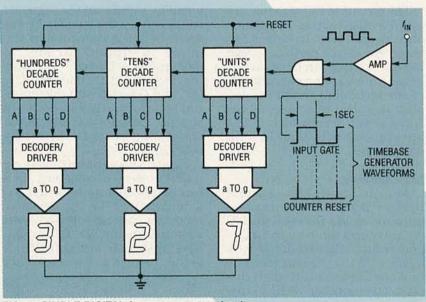
In the last article we introduced 7-segment displays and simple BCD-to-7-segment decoder/driver IC's that can be used to activate those displays. Figure 1 shows how three sets of those IC's can be used with a trio of decade counters to make a simple digital-readout "frequency" meter. In the figure, the amplified external frequency signal is fed to the input of the series-connected counters with one pin of a 2-input AND gate. The other input is derived from a timebase generator.

When the timebase input signal is low, the AND gate is off, and there is no input to the counters. When the timebase gate signal switches high, a brief RESET pulse is fed to all three counters, setting them all to zero. Simultaneously, the input gate turns on and remains on for one second. During that time the input-frequency pulses are summed by the counters. At the end of the one-second period, the gate turns off as the timebase gate signal goes low again. That ends the count and enables the display modules so that they give a steady reading of the total pulse count, and therefore the frequency. The whole process repeats itself again one second later when the timebase gate signal goes high.

The simple system illustrated has one major drawback: The display blurs during the counting period, becoming stable and readable only when each count is complete and the input gate is off. Figure 2 shows a circuit for frequency meters designed to overcome display blurring. In the circuit, a 4-bit data latch is wired between the output of each counter and the input of its decoder/driver IC.

# WORKING WITH LED DISPLAY DRIVERS

Learn all about 7-segment display decoder/drivers in our continuing coverage of optoelectronic IC's.



**BAY M. MARSTON** 

FIG. 1—SIMPLE DIGITAL frequency-meter circuit.

When the timebase gate signal goes high, a reset pulse is fed to all counters, setting them to zero. Simultaneously, the input gate is turned on, and the counters start to sum the input signal pulses. The count continues for one second while the 4-bit latches prevent the counter output from reaching the display drivers. As a result, the display remains stable during that interval. At the end of the period a brief LATCH-ENABLE pulse is fed to all latches. The instantaneous binarycoded decimal (BCD) outputs of each counter are then latched into memory, and also fed to the display via the decoder/driver IC's. That steadies the reading of the display to give a total pulse count, which corresponds to the input signal frequency. A few moments later, the sequence repeats itself with the counters resetting and then counting the input frequency pulses for one second, and so on.

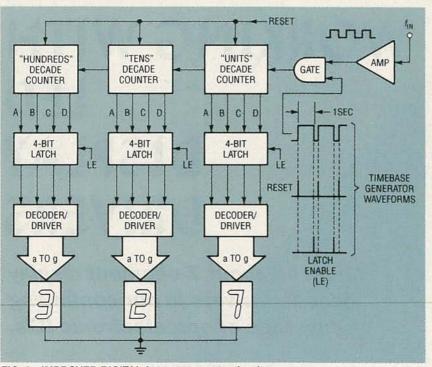


FIG. 2—IMPROVED DIGITAL frequency-meter circuit.

The circuit in Fig. 2 generates a stable display that is updated once per second. In 'actual circuits, the count periods shown in Figs. 1 and 2 can be set at any decade multiple or submultiple of one second, provided the display is suitably scaled. Many commercial decoder/driver IC's are available with built-in 4-bit data latches.

#### Multiplexing

You can see in Figs. 1 and 2 that at least 21 connections must be made between the IC circuitry and the 7-segment displays of a 3-digit readout. Similarly, at least 70 connections must be made for a 10-digit display. The number of IC-to-display connections can be significantly reduced with a technique known as *multiplexing*. Figures 3 and 4 illustrate multiplexing.

Figure 3 shows how each digit of a 3-digit common-cathode LED display is individually activated with only 10 external connections. In the display circuitry, all "a" segments are connected together, as are all other sets of segments ("b" to "g"). Thus only seven external "a" to "g" connections are made to the display, regardless of the number of digits used. However, no 7-segment display is illuminated by signals on the segment wires unless the display is enabled by tying its common terminal to ground. In Fig. 3, enabling is achieved by activating switching transistors Q1, Q2, and Q3 with suitable external signals. However, this scheme calls for an additional connecting wire for each display module.

Figure 3 also shows how three different sets of segment data can be selected with switch S1a, and each of the three display modules can be selected with switch S1-b and Q1, Q2, or Q3. S1-a and S1-b are ganged together to form a two-pole rotary switch which provides the mu tiplexing. These switch el ments represent a fast-actir electronic switch capable of cy ling the contacts through pos tions 1, 2, and 3 fast enough f flicker-free multiplexing.

The operating sequence of the circuit shown in Fig. 3 will 1 discussed here. Assume in tially that the contacts of S1 and S1-b are in position 1 : that S1-a selects segment da of display module 1  $(1_{a-e})$ , ar S1-b activates display module with Q1. Module 1 will the show an illuminated numb "3." Moments later the swite contacts move to position 2, s lecting segment data 2 (2and illuminating module 2 wit Q2. Module 2 will then be luminated to show the numb "2." Moments after that, th switch contacts move to pos tion 3, causing module 3 show the number "7." Thu only one digit is ever on at an one time.

In practical displays, the s quence is repeated fast enoug so that you do not see the di play segments being turned c and off. Your eye's persistence vision makes it look as if th three digits are all lit up to gether. The multiplexing frquency must be about 1 kHz.

Figure 4 shows a more rea istic arrangement for mult plexing a 3-digit frequence meter. The multiplexer (MUX) located between the outputs the three BCD data latches are the input of a display-drivir. BCD-to-7-segment decode

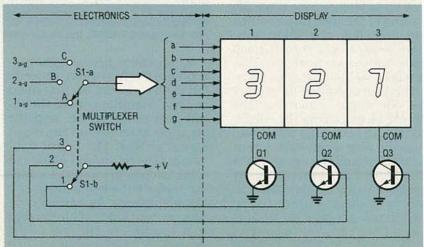


FIG. 3—MULTIPLEXING METHOD for a 3-digit common-cathode LED display.

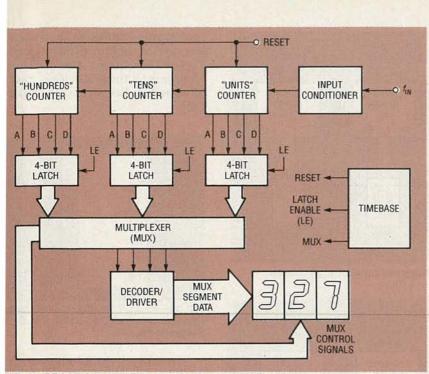


FIG. 4—PRACTICAL MULTIPLEXING of a 3-digit frequency meter.

driver IC. The scheme shown has two major advantages: Only one decoder/driver IC is needed (regardless of the number of readout digits), and its multiplexer includes only five ganged 3-way sequencing switches. One of those switches is for control data, and four are for BCDsegment data. That arrangement saves three ganged 3-way switches, as compared with the eight needed in Fig. 3.

Commercial large-scale integrated (LSI) IC's now available can perform all of the counting, latching, multiplexing, decoding, timing, and display-driving functions in Fig. 4. A device of this complexity is typically packaged in a dual-in-line package (DIP) with only 20 pins. Those pins provide for all necessary connections to the power supply, display modules, and inputs. Therefore, a complete 4digit counter can be built with a dedicated IC in a circuit such as that shown in Fig. 5. Another example of an LSI IC in a display circuit is the 3<sup>1</sup>/<sub>2</sub>-digit digital voltmeter (DVM) chip shown in Fig. 6.

#### **Ripple blanking**

Unless there is automatic suppression of the two unwanted leading zeros, the 4-digit circuit in Fig. 5 will give an

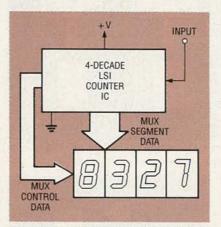


FIG. 5—4-DIGIT COUNTER CIRCUIT based on an LSI chip.

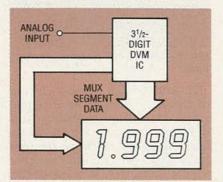


FIG. 6—3-1/2-DIGIT VOLTMETER based on an LSI chip.

actual reading of 0027 if it is used to measure a count of 27. Similarly, if the 3<sup>1</sup>/<sub>2</sub>-digit circuit of Fig. 6 measures 0.1 volt, it will display 0.100 volt unless the two unwanted trailing zeros are automatically suppressed.

In practical circuits, automatic blanking of leading or trailing zeros can be performed with a *ripple-blanking* technique, as shown in Figs. 7 and 8. Each decoder/driver IC (with a BCD input and 7-segment output) is provided with a RIPPLE-BLANKING INPUT (RBI) and scripple-blanking output (RBO) pins.

Assuming those pins are active-high, if the RBI terminal is held low (logic 0), the 7-segment outputs of the IC are enabled, but the RBO terminal is disabled (held low). If the RBI terminal is biased high (logic 1), the 7-segment outputs become disabled in the presence of a BCD 0000 input, and the RBO output goes high under the same condition. The RBO terminal, therefore, is normally low and goes high only if a BCD 0000 input is present at the same time the RBI terminal is high.

Figure 7 shows the rippleblanking technique for leadingzero suppression in a 4-digit display with a reading of 207. The RBI input of the thousands or most significant digit (MSD) decoder/driver is tied high, so the readout is automatically blanked in the presence of a zero when the RBO terminal is high. Consequently, the RBI pin of the hundreds IC is high, the readout shows 2, and the RBO terminal is low. The RBI input of the tens unit is low, so the readout shows 0 and its RBO output is low. The units readout shows the least significant digit (LSD), which does not require zero suppression. Its RBI pin is grounded and the readout shows 7. The display therefore gives a total reading of 207.

In the leading-zero suppression circuit, Fig.7, the rippleblanking feedback is applied backwards from the MSD to the LSD. Figure 8 shows how trailing-zero suppression is accomplished by reversing the direction of feedback from the LSD to the MSD. Therefore, when an input of 1.1 volts is fed to that circuit, the LSD is blanked, because its BCD input is 0000 and its RBI input is high. The RBO

F

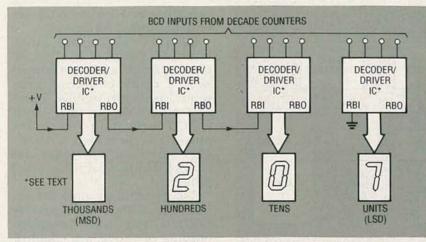


FIG. 7-RIPPLE-BLANKING for leading-zero suppression in a 4-digit counter.

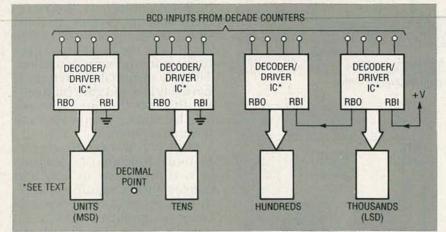


FIG. 8—RIPPLE-BLANKING for trailing-zero suppression of the last two digits of a 3-1/2-digit DVM readout.

terminal is high under that condition, so the hundredth's digit is also blanked in the presence of a 0000 BCD input.

Most decoder/driver IC's have RIPPLE-BLANKING INPUT and RIPPLE-BLANKING OUTPUT pins. Usually these pins are activelow. If a decoder/driver IC does not include integral rippleblanking logic, it can usually be obtained by adding external logic circuitry similar to that shown in Fig. 9. The RBO pin is connected to the BLANKING INPUT pin of the decoder/driver IC. Figure 9 shows an active-high circuit in which the output of the 4-input NOR gate goes high only with a 0000 BCD input. The RBO output goes high only if the 0.0 input is present when RBI is biased high.

#### Decoder/driver IC's

Many decoder/driver IC's are available commercially as both

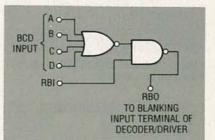


FIG. 9—ACTIVE-HIGH ripple-blanking logic.

TTL and CMOS devices. Some of those devices have integral ripple-blanking capability, and others have built-in data latches. A few of those devices have built-in decade-counter stages. Let's look at some of the most popular of those devices.

The 7447A and 7448 7-segment decoder/driver IC's are in the standard TTL family. They are also available in a low-power Schottky (LS) form designated as 74LS47 and 74LS48, respectively. All of those IC's have integral ripple-blanking facilitie but do not include data latch Figure 10 shows the pino common to those devices, ea of which is housed in a 16-p DIP.

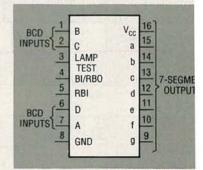


FIG. 10—PINOUT for BCD-to-7-segm decoder/drivers IC's 7447A, 74LS 7448, and 74LS48.

The 7447A/74LS47 has an tive-low output designed driving a common-anode L with external current-limiti resistors (Rx), as shown in F 11. The 7448/74LS48 has an tive-high output that drives common-cathode LED ir manner similar to that of 1 circuit in Fig. 11, but with 1 common terminal of the disp connected to ground. In cases, the Rx current-limiti resistors should be chosen limit the individual segme currents below the following: solute limits:

> 7447A = 40 mA 74LS47 = 24 mA 7448 and 74LS48 = 6 mA

Figure 12 shows how 7448/74LS48 can drive a liqu crystal display (LCD), using pair of 7486 or 74LS86 quad input xor gate IC's. An extern 50-Hz square wave applies t necessary phase signals to t display.

As shown in Fig. 10, each the 7447/7448 IC's has three i put control pins:LAMP TEST, RBO, and RBI. The LAMP TEST p drives all display segments when the pin goes to logic-le with the RBO pin on or at log high. When the BI/RBO pin pulled low, all outputs a blanked. The BI/RBO pin al functions as a RIPPLE-BLANKI OUTPUT pin. Figure 13 show how to connect the RIPPI BLANKING pins to give leadir

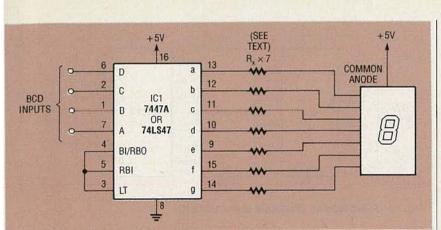


FIG. 11— DRIVING a 7-segment common-anode LED display with a 7447A-type decoder/driver.

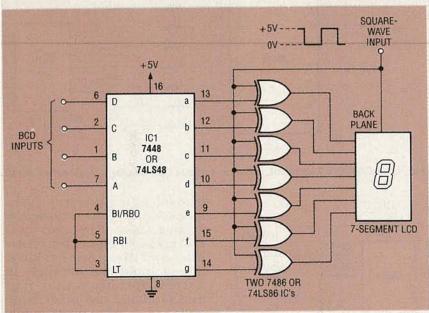


FIG. 12—DRIVING a 7-segment LCD with a 7448-type decoder drivers.

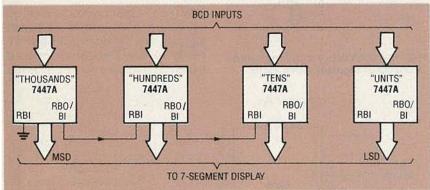


FIG. 13— SUPPRESSION OF FIRST THREE digits of a 4-digit display with 7447A-type decoder/drivers.

zero suppression on the first three digits of a 4-digit display.

The 4511B is a BCD-to-7-segment decoder/driver IC with an integral 4-bit data latch, but it lacks built-in ripple-blanking. That CMOS device features NPN bipolar transistor output stages capable of handling output currents up to 25 mA. It can drive most popular 7-segment displays. Figure 14 is a functional diagram of the IC. The 4511B will operate from any 5- to 18volt power supply.

The 4511B has three input-

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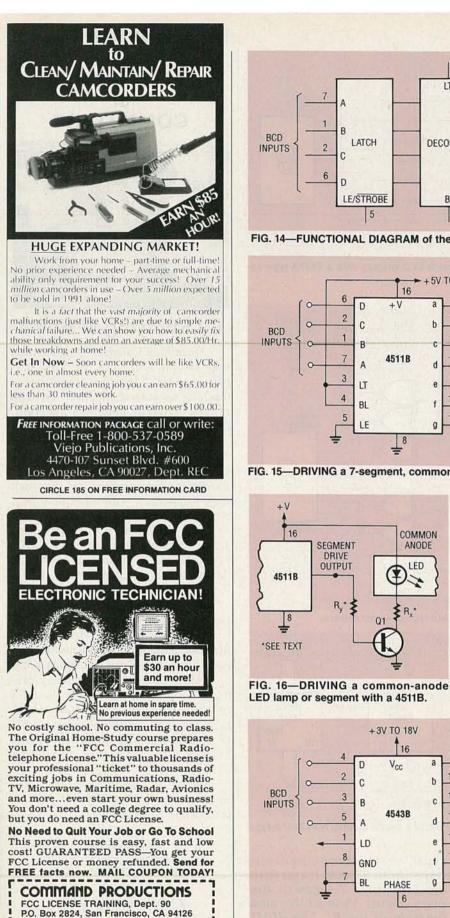
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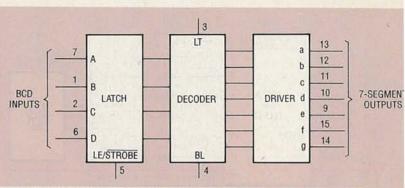
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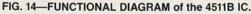
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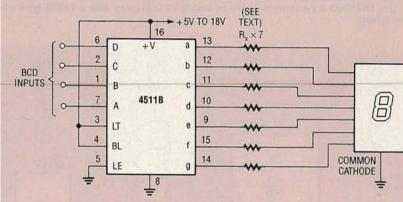
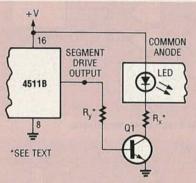


FIG. 15—DRIVING a 7-segment, common-cathode LED display module with a 451



control pins:LAMP TEST (LT) BLANKING (BL), and LATCH ENABL STROBE (LE/S). The LAMP TE: and BLANKING inputs are activ low, and the LATCH ENABL STROBE input is active-high. normal operation, LAMP TE: and BLANKING are made high and LATCH ENABLE/STROBE is he low.

When the LATCH ENABL STROBE pin is low, BCD inp signals are decoded and fed o rectly to the 7-segment outp pins. If LATCH ENABLE/STROI

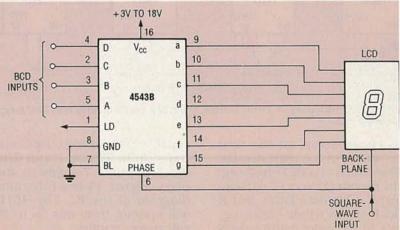


FIG. 17—DRIVING a 7-segment LCD with a 4543B.

joes high, the BCD input sigials present at the moment of ransition are latched into nemory and fed (in decoded orm) to the 7-segment outputs, while LATCH ENABLE/STROBE emains high. If the LAMP TEST nput is grounded, all output egments are activated, rejardless of the BCD inputs. If he BLANKING input is grounded while LAMP TEST is positive), all output segments are blanked.

Figure 15 shows the basic connections for driving a comnon-cathode LED. A currentimiting resistor ( $R_x$ ) must be wired in series with each disolay segment, and it must have a value chosen to hold the segnent current below 25 mA. Note that the segment outputs of the 4511B are not internally current-limited. Therefore, the device has no output-overload protection.

Figure 16 shows how to modiy the circuit in Fig. 15 to drive an LED common-anode display. In the example shown in Fig. 16, an NPN buffer transistor must be used between each outputdrive segment and the input segment of the display. Resistor  $R_X$  sets the operating segment current of the display in those examples, and  $R_Y$  sets the base current of the transistor.

The 4511B can also drive 7segment liquid-crystal displays (LCD) with an external squarewave PHASE signal and a set of XOR gates similar to those of Fig. 12. In practical circuits, however, it is better to use a 4543B IC for that specific application.

The 4543B is a 7-segment CMOS decoder/driver with an integral 4-bit data latch. It was designed for driving LCD's, but it can also drive most other 7-

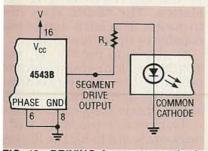


FIG. 18—DRIVING A common-cathode LED lamp or display segment with a 4543B.

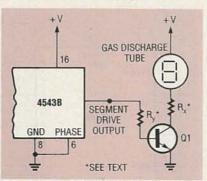


FIG. 19—DRIVING A gas-discharge tube or display segment with a 4543B.

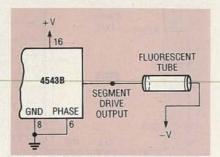


FIG. 20—DRIVING A fluorescent tube or display segment a 4543B.

segment displays.

The 4543B has three input control pins: LATCH DISABLE (LD), PHASE, and BLANKING (BL). In normal use, the LATCH DISABLE pin

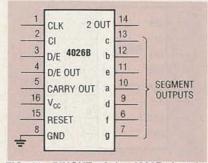


FIG. 21—PINOUT of the 4026B decade counter with 7-segment outputs.

is biased high and the BLANKING pin is tied low. The state of the PHASE pin depends on the display that is being driven. For driving LCD displays, a square wave (approximately 50 Hz, swinging fully between the ground and  $V_{\rm cc}$ ) must be applied to the PHASE pin. The PHASE pin must be grounded for driving common-cathode LED's and it must be tied to logic-high for driving any common-anode displays.

The display can be blanked at any time simply by driving the BLANKING pin to the logic-high state. When the LATCH DISABLE pin is in its normal high state, BCD inputs are decoded and fed

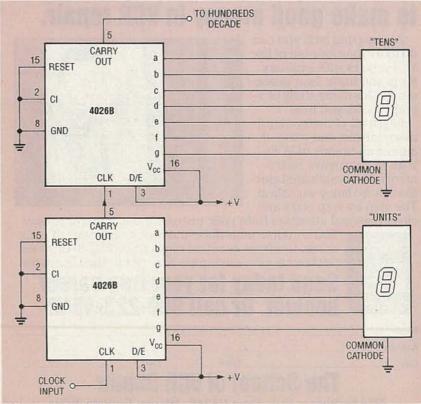


FIG. 22—CASCADING TWO 4026B decade counters.

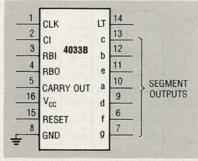


FIG. 23—PINOUT for the 4033B decade counter with 7-segment outputs and ripple blanking.

directly to the 7-segment output pins of the IC. When the LATCH DISABLE pin is pulled low, the BCD input signals present at the moment of transition are latched into memory and fed (in decoded form) to the 7-segment outputs while LATCH DISPLAY remains low.

Figure 17 shows a method for using the 4543B to drive an LCD, and Figs. 18–20 show how that circuit can be modified to drive other 7-segment displays.

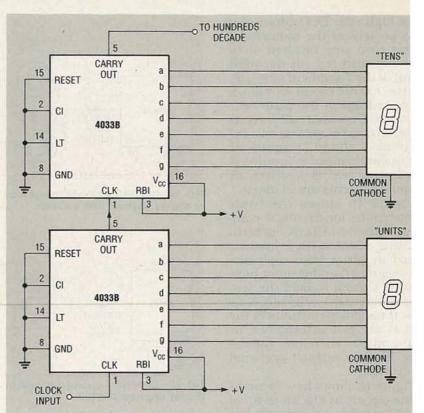


FIG. 24—CASCADING TWO 4033B decade counters without zero suppression.

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The value of  $R_x$  in Fig. 18 m be chosen to limit the output drive current to below 10 mA<sub>1</sub> segment (individual LED lar If higher drive currents a needed, use a buffer transis between the output of t 4543B and the input of the c play segment.

Figure 21 shows the pinou the 4026B, a complete deca counter with integral decoc driver circuitry. It can drive a segment common-cathode L display directly. The segme output currents are interna limited to about 5 mA with a volt supply or 10 mA with a volt supply. Therefore, the c play can be connected directly the output of the IC without ternal current-limiting sistors. The 4026B does 1 include a data latch and is 1 capable of ripple blanking.

As shown in the Fig. 21, 1 4026B has four input cont pins and three auxiliary out pins. The input pins are des nated: CLOCK (CLK), CLOCK INF IT (CI), RESET (RESET), an DISPLAY/ENABLE (D/E). The IC I a Schmitt trigger on its CLC (CLK) input line, and clock s continued on page

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# **AUDIO UPDATE**

### Let's Phase the Music: More comments on papers from the 91st AES convention

#### LARRY KLEIN

coustic phase, and its audibility, has been a subject of controversy ever since ermann von Helmholtz addressed e subject in the late 1800's. Yet. espite all we've learned about earing since Helmholtz, and the irrent availability of sophisticated w research tools, the controversy igers on. The argument is not, as he might think, simply an abstract, sometimes heated, debate nong philosophical psychocousticians. It has real-world conequences for the way audio signals hould be recorded an audio equipent should be designed.

I've been aware ever since I built v first Williamson amplifier in the arly 1950's that signal phase shift as a significant audio parameter-: least in power amplifiers. In its arly incarnations, the Williamson uffered from marginal instability ecause the 20-dB negative feedack loop sometimes shifted ositive at the frequency extremes. hat frequently resulted in pulsing oofers at one end of the spectrum nd RF oscillation at the other. But side from such obvious instability isturbances, I've never felt that linar phase shift posed a severe reat to fidelity.

In the early 1980's, many speaker nanufacturers decided that the onventional two or three drivers intalled on the single vertical front anel of a system resulted in acousc phase shifts caused by the diferent path lengths to the listener's ars of the woofer and tweeter sigals. That was usually "solved" by tepping a cabinet's front panel so hat the woofer was several inches prward of the tweeter. Aside from enerating new opportunities for dvertising copy writers (and some xpensive pot-bellied speakers), ne arrangement's audible advanages were elusive-at least to my ars. And, in any case, it seems to he that on a purely theoretical basis uch "time alignment" would only hold when a listener's ears were also aligned exactly on axis with the acoustic center of the two (or more) drivers.

I once saw/heard a speakerphase demonstration using a twoway system with a movable tweeter. A square wave was fed to the system and the speaker's output

#### INTERCHANNEL SPEAKER PHASE

Most of us first encountered "phase" as an audio phenomenon when we were told to make sure that the right and left speakers of our stereo system were connected in proper polarity. That meant that when presented with, say, a positive audio pulse from both amplifier channels, the cones of both woofers would move simultaneously in the same direction. If the speakers were wired incorrectly, the cones would move in opposite directions, and bass performance would suffer. In addition, the 180degree phase differences in the high frequencies would confuse the ear's localization system, disturbing the stereo sound-stage image and imparting a vague "phasiness" to the reproduced sound. In general, speaker polarity/phase, if correct in the original installation, remains so unless the connections are changed. R-E

picked up by microphone and displayed on an oscilloscope. As the tweeter installed on top of the system was slowly shuttled back and forth over a five-inch distance, the square wave would visibly distort and then restore itself as its highand low-frequency harmonic components shifted in and out of phase. I stood in front of the display for several minutes, really trying to hear the waveform distortion that was clearly evident on the scope screen. Following in the tradition of Herr von Helmholtz, I never did hear any effect. All of this is background for the latest phase controversy beautifully delineated in the following Audio Engineering Society (AES) preprint.

#### Observations on the Audibility of Acoustic Polarity [Greiner and Melton (3170 K-4)].

To understand what is meant by absolute polarity, it helps to use a simple musical example, such as the sound of a kickdrum. The impact on the drum head produces an air compression that moves outward and is picked up by a recording microphone. Ultimately, it is reproduced by a forward-moving speaker cone as an air compression. However, anyone who deals with electronics knows that the original electrical signal from the microphone has had its polarity flipped probably dozens of times by the recording and reproduction electronics before it reaches the speaker. In fact, there's a 50-50 chance that the initial kickdrum impact is being reproduced by a speaker cone that is pulling at the air rather than pushing it. In other words, what reaches our ears is an air rarefaction, although the original was a compression. It follows that our eardrums are also being pulled rather than pushed. Does it make a difference? That's what Greiner and Melton set out to explicate in their paper.

The audibility—or inaudibility—of absolute inversion is a recent element in the ongoing discussion. Unlike other phase shift phenomena discussed earlier, polarity inversion does *not* change the shape of a transient signal nor shift the phase relationship among its component elements. However, as the authors state, it does present to the ear a fundamentally different signal.

A major part of the authors' research effort involved extensive, carefully controlled listening tests to determine the types of signals

most sensitive to polarity inversion. It seems self evident that highly asymmetrical signals were most likely to be audibly changed by inversion. For example, note the trombone waveform shown in Fig. 1.

In general, it could be said that acoustic polarity inversion is clearly audible in some circumstances (particularly with test signals), although most of the time with real-world music it is not. It seems that even when waveforms have clear asymmetries as in Fig. 1—and not all of them do other effects inherent in the complex nature of most musical tones tend to mask the identifying characteristics of the inversion.

The authors conclude that "while polarity inversion is not easily heard with normal complex musical program material, as our large-scale listening tests showed, it is audible in many select and simplified musical settings. Thus it would seem sensible to keep track of polarity and to play the signal back with the correct polarity to assure the most accurate

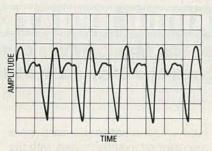


Fig. 1. Trombone waveform. In general, reed instruments show the greatest asymmetry.

reproduction of the original waveform."

No one can argue with the above as an ideal goal, but exactly how is it likely to be realized? Assuming that (1) we have a recording with all the instruments recorded in phase, that (2) our three-way speaker doesn't have its mid-range connected in inverted polarity to flatten the system's frequency response (a no longer common practice), and that (3) anyone in authority cared enough to manufacture discs and tapes with all their polarity ducks in



a row, how would the consu know when he had it right?

Next month I'll discuss one r AES paper and then call it quit: the 91st Convention.

#### PHASE INSENSITIVITY

It is well known that loudspeak cone travel is seldom equally line in both its outward and inward e cursion. In other words, the speak cone's push on the air may not I exactly equal to its pull, even whe driven by a low-distortion, high symmetrical sinewave. It seen probable to me that most huma ears are also asymmetrical in the responses to the compressions ar rarefactions of an impinging acou tic waveform and therefore genera harmonic distortions. Whether ar individual listener's degree of hea ing asymmetry correlates with the sensitivity to absolute phase is a open question. I noted with intere that Greiner and Melton suggest one point that a distorting sour system may increase the audibili of polarity inversion.

I know for a fact that I've alway been comparatively insensitive phase. For example, in the ear days of quadrophonics I was eva uating a four-channel synthesiz that attempted to generate artifici rear channels by tapping off th front channel, "wobbulating" it at 25-Hz rate, and feeding it to a pair rear speakers. The four-chann effect, for better or worse, was ban ly perceptible to my ears. A frier dropped in unexpectedly while was doing my listening and greete me with, "Hi Larry ... (three-secor pause) ... your rear speakers a out of phase." I sat alongside him c the couch between the wobbulate rear speakers and for the life of m couldn't hear what he had detecte instantly-and claimed was givin him a headache!

I also know that some listene are incredibly sensitive to very sma percentages of tape flutter, phase-shift phenomenon, but th numbers have to get pretty bad be fore I hear it. In case you are wo dering, my overall acuity ar sensitivity were (by test) pretty god in those days, so my phase inse sitivity was, in effect, an independe variable. I wonder what the norm distribution of phase sensitivity i and whether ignoring it may not ske the results of psychoacoustic stu ies of phase.

## **COMPUTER CONNECTIONS**

#### The personal digital assistant.

JEFF HOLTZMAN

adget lovers take heed: Personal Digital Assistants (PDA's) will be the fastestrowing segment of the computer ousiness within two years-maybe by the end of this year. What is a Personal Digital Assistant? It's a paperback-size computer that bridges he gap between calculator-like oranizers on the one hand, and hefty totebooks on the other. In so doing, he PDA hopes to provide both power and ease of use, but without he size, weight, and battery-life and weight constraints of a notebook PC.

Numerous technologies are comng together to make this category a eality. Low-voltage CPU's, system controllers, video controllers, and DRAM's from Intel, Cyrix, AMD, Nestern Digital, and others promse to drastically lower power consumption. The chips run at 3.3 volts. and could more than double battery ife, from 3.5 to 8.0 full hours. Emerging standards for credit-card sized memory modules seem to be taking nold. Hard-disk drives continue to shrink. Integral Peripherals has introduced 20- and 40-megabyte drives in 3-ounce packages that measure  $2'' \times 3''$ , and that consume 500 mW of power during operation and 15 mW in the "sleep" mode. IBM, Conner, Seagate, and others will ship similar drives later this year.

Look for two basic approaches to the PDA. One, promulgated by PC vendors, amounts to a miniature DOS machine with a QWERTY keyboard that allows touch typing. It has a full  $80 \times 25$  ( $640 \times 200$ , CGA-level graphics) non-backlit LCD screen. The other, promulgated by consumer electronics companies (e.g., Sony), consists of a proprietary device using a stylus input in place of a keyboard.

The DOS version may well conform to a spec developed by Phoenix Technologies, called the Companion PC (CPC). A CPC will be based on an 8086-level computeron-a-chip developed by Chips & Technologies (see the January 1992 issue of this column for more information). PCMCIA memory cards, Phoenix BIOS and power management software, and special battery technology developed by Duracell. As for software, Microsoft has delivered a ROMable version of DOS 5.0, and promises to do the same for Windows 3.1. Lotus might supply a suite of applications, including a task switcher, although Microsoft Works could very well provide steep competition.

Tantalizing views of the other category have been revealed by Sony. And there are persistent rumors that Apple and Sony are working together to put a Mac-style interface on this new breed. Sony has released one such device in Japan, but apparently has no plans to mar-



FIG. 1—DYCAM'S MODEL 1 digital camera stores 32 shots, each with a resolution of  $376 \times 240 \times 256$  gray levels. It's a harbinger of the future for PC-based digital imaging.

ket it here.

Electrical power usage remains a major issue with portable computers of all configurations. In the ROMable versions of DOS 5.0 and Windows 3.1. Microsoft has introduced an Advanced Power Management (APM) Application Programming Interface (API), co-developed with Intel and Phoenix, that provides the system with information about which peripherals are not being used, so it can remove and thereby conserve power. Several battery manufacturers have introduced flat. wide batteries that provide both a better fit with today's sleek designs, and more power. For example, Portable Energy Products has a battery that measures  $2.6'' \times 3.9'' \times 0.3''$ and produces two volts at three amps.

One interesting device is the "Commuter Computer," introduced by Memorex at the January Consumer Electronics Show (CES). The device weighs 1.3 pounds, lists for \$599, and includes a word processor, scheduler, notepad, the DR-DOS 5.0 operating system, carrying case, power adapter, and cables; an external 1.44 megabyte floppy drive is available for \$299.

#### **Multimedia update**

Here's an imaginative use of multimedia: The Virtual Press Conference. Redgate Communications of Vero Beach Florida plans to outfit more than 100 movers and shakers of the computer industry with 1.2meter satellite dishes, TV's, VCR's, fax machines, and dedicated phone lines, at a cost of about \$2 million. Redgate will then broadcast to them hour-long interactive press conferences, during which they can ask questions, receive printed materials by fax, and record the whole thing on VHS tape. The viewers will pay nothing for the service; Redgate in-



Radio-Electronics, May 1992 76



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stead sells the service to large computer companies for \$25,000 per hour. Lotus and DEC have expressed interest.

The Multimedia PC Marketing Council has upgraded the "baseline" spec for a multimedia PC to include a 386SX. The original version specified a 10-MHz 286, which simply didn't provide enough horsepower to run serious applications. In addition, the rapid infusion of 386 technology into general buying habits makes 286 compatibility less of an obstacle than it once was.

Apple has released a multimedia extension to its System 7.0 software. It's called QuickTime, and it provides developers with a standard software architecture for integrating time-variant data (sound, video, animation). In introducing the product, Apple CEO John Sculley said that QuickTime will be the major legacy from his tenure at Apple—more important even than the Macintosh.

The OuickTime architecture includes four components: system software, file formats, compression/decompression technology, and user-interface standards. Apple hopes it will become a cross-platform standard with support for Mac. DOS, Windows, UNIX, Silicon Graphics, DEC, and Cray systems. The video format is 160 × 120 pixels, running at 15 frames per second. That's obviously no competition for broadcast-quality video (BOV); Apple hopes that the standard will evolve to accommodate more powerful hardware that can support BQV.

Apple has announced a Quick-"player" that will allow PC-Time ' based Windows users to experience QuickTime movies. (Corel Systems, of CorelDraw fame, wrote the file-format translator.) In addition, Microsoft has announced support for QuickTime; add-ins for the Mac versions of Excel and Word will allow end-users to build simple movies within the respective applications. You could use this technology to combine a series of spreadsheets into an animated demonstration of how sales are projected to grow.

In response, IBM quickly demonstrated technology with twice the resolution and the same frame rate. Naturally it runs under OS/2 2.0. One demo showed how advanced features of OS/2 allowed several program processes to control separate halves of a duet in sync.

#### **Product watch**

The Dycam Digital Still Camera Model 1 is the first of a new breed of all-electronic snapshot takers. The Dycam is an easy-to-use, lightweight, hand-held camera that allows you to take as many as 32 pictures, then upload them to your PC or Mac for bit-map editing or desktop publishing. (Logitech markets a nearly identical version under its own name.)

Despite flawed installation instructions, getting the Dycam going is easy. Attach a base mount to the main unit and plug in a wall-mount charger. When the LED starts blinking, the battery is fully charged. Then it's just point, shoot, and upload the images to your PC via an RS-232 port.

The Dycam comes with control software for both PC and Mac; I tested the PC version, which actually comes with both Windows and DOS versions. I found the Windows version unstable, due to frequent UAE's (Unexplainable, unintentional, unwanted Application Errors); consequently I worked mostly with the DOS version.

The DOS-based software works pretty well. The main screen consists of a series of "thumbnails," miniature versions of each image. To view an enlarged version, use the arrow keys to highlight the desired image, press Enter, and the software downloads, decompresses, and displays the complete image; the whole process takes about 10 seconds per shot, depending on system speed.

Documentation is both clear and brief and it includes both Mac and PC instructions in one booklet; the software includes on-line help in the form of brief summaries of keystroke options.

The camera can operate in two modes, regular and tripod. Unfortunately, to switch modes you must download different software to the camera, a process that takes about 45 seconds. In regular mode, you just point and shoot; the camera determines shutter speed a whether to use the flash. Trip mode allows you to set the shut speed manually, from 1 to 640 m seconds. Shutter bugs would proably prefer to set shutter speed familiar inverse time units (1/t, e. 1/60, 1/30, etc.). There's no way attach a standard remote shutter lease, but you can take a pictu from the keyboard (Alt-T).

The software allows you to sa images to disk in a variety of f mats, including EPS, TIFF, PCX proprietary Dycam format, a others. The software provides sol ability to adjust image contrast a brightness, but for optimal resu you'll want to use a third-party to (e.g., PhotoShop). When the ca era is full, you will have to clear it c (Alt-C) to make room for more, a you cannot selectively delete c tures from memory. Although t camera maintains a time/da clock, you cannot include a tirr date stamp as part of the image.

When I first learned of t Dycam, I thought it would be ideal way to develop product a screen shots for the magazine. L fortunately, that will not be feasib The problem is not limitations w the software, but the optics. T lens system provides image qua comparable to a \$20 instant ca era-but the Dycam lists for almo \$1000. The CCD provides a tc resolution of 376 × 240 pixe which is less than found in a TV compensates by providing 256 l els of gray, which definitely hell because the eye responds better variation in tone and color than absolute resolution. Even the be shots, however, come out grai Shots of people generally fare b ter than inanimate objects.

Add-on lenses are available, cluding wide-angle, telephoto, a macro (close-up). Other access ries include an adapter for 12-v auto operation.

Despite its limitations, t Dycam opens a world of pc sibilities for users who don't ne publication quality. For example real estate agent could publi house shots to give prospective ents some idea of layout. Scho could create classroom newsl ters.

Dycam is to be commended for being first to market with an all-elecronic digital camera. An improved optic system, faster software, and nore reliable Windows operation vould make it a must-have item for professional desktop publishers. Geep your eye on this one. If you're nterested, contact Dycam, Inc., 1588 Topanga Canyon Blvd., Chatsworth, CA 91311 (818) 198-8008.

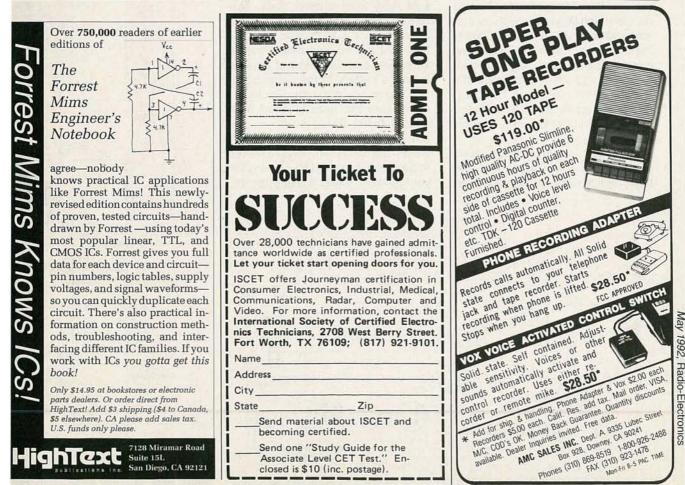
#### **Vews** bits

Slate Corporation is a technology eader in the fledgling Pen-computng market. Two officers of the comoany, **Dan Bricklin and Bob rankston**, designed the first preadsheet for personal computers, VisiCalc. The dynamic duo is at t again, with a pen-based preadsheet called At-Hand. Atland reads and writes 1-2-3 and Excel files, and runs under the Pen-Point operating system. It's schediled to ship in the second quarter.

After several false starts with portable and notebook PC's, the new IBM is trying a new tack. Big Blue just bought a 6% chunk of the French company, Groupe Bull, which owns Zenith Data Systems. The deal provides the rights for IBM to sell ZDS portables, and will give Groupe Bull access to IBM's RISC technology. The deal also provides for swapping microprocessor technology; one report indicates that GB will join IBM, Apple, and Motorola in developing the PowerPC, part of last summer's accord between IBM and Apple.

Worried about Japanese encroachment on American markets? Maybe it's time to start learning how to deal with them on their own terms—literally. A company called BayWare (415-949-3190) has introduced a Windows-based program that the company claims can teach the essence of Japanese in four to eight weeks, by studying only an hour per day. **Power Japanese**  was developed using my favorite multimedia authoring system, Tool-Book, and it includes an audio output device that connects to a standard parallel port. The device, which is housed in the connector shell, contains an 8-bit D/A converter and it can drive headphones or an audio amplifier. Software includes 2000 files of native Japanese speech, animated writing sequences, and a set of progressive lessons. The company plans to release versions in French, Spanish, German, and advanced Japanese.

Big-bucks consumer-oriented advertising is becoming the rule in the PC business. You've probably seen Intel's flashy "Vacancy" ads on cable. Now Microsoft is gearing up for a **\$30 million Windows promotion,** including \$8 million for TV spots alone. 1992 is going to be remembered in the PC industry as the year the battle for graphical operating environments became really down and dirty. **R-E** 



## DRAWING BOARD

### One step backward to automotive regulators and two steps forward to our oscilloscope.

#### ROBERT GROSSBLATT

ust when I thought we had driven in the final nail on the whole business of automotive charging systems, I got a letter with some stuff in it that I have to pass on. Even though we spent all our time talking about alternatorbased systems, there was a lot of mail from people who had generators. The voltage regulators we designed were specifically aimed at alternators but a regulator is still a regulator is still a regulator and, if you give it some thought, the one we designed can be used with a generator as well.

The most important factor in modifying the design is to realize that regulators for generator have to handle much larger amounts of current and because of that, our existing circuit can't drive the generator directly. The standard way to do deal with this is to use relays and that's exactly what was done by Craig Shippee, a reader from Bridgewater, Massachusetts who sent his design to me in the mail.

His circuit, shown in Fig. 1, does several interesting things. First, the basic design is the one we did for an alternator with a pulled-up field (one side of the alternator's field windings are hard wired to the *positive* side of the battery) but the generator is set up with a grounded field (one side of the field hard wired to *ground*). The other interesting part of the design is that it uses only one DPDT relay instead of the two or three relays found in standard generator regulators.

Craig's circuit is a classic example of what happens when need comes up against ingenuity and experimentation. He said that he couldn't give me the exact specifications on the relay he used because he salvaged an old 30-amp light relay from a dead truck. The relay opens when the regulator doesn't want it to charge the battery, and the 40-amp diode makes sure there's no backflow of current from the battery to the generator.

If you're going to use the circuit, make sure the relay is rated high enough to handle the maximum output current of the generator and tweak the 500-ohm potentiometer to get the cut-in point that's best for your system. Craig is using the cir-

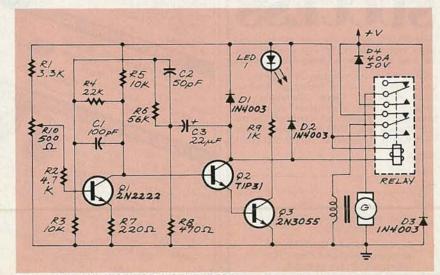


FIG. 1—A REGULATOR FOR A GENERATOR has to handle a lot of current, so a relay must be used. cuit on a tractor but there's no son why it can't be used with generator-based automotive e trical system.

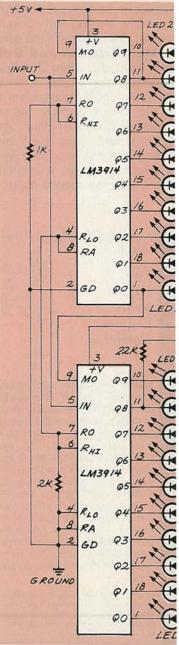


FIG. 2—TO EXPAND THE VERTICAL play to twenty elements we'll have to a second LM3914 to the circuit. 3914's were designed to be easily o caded.

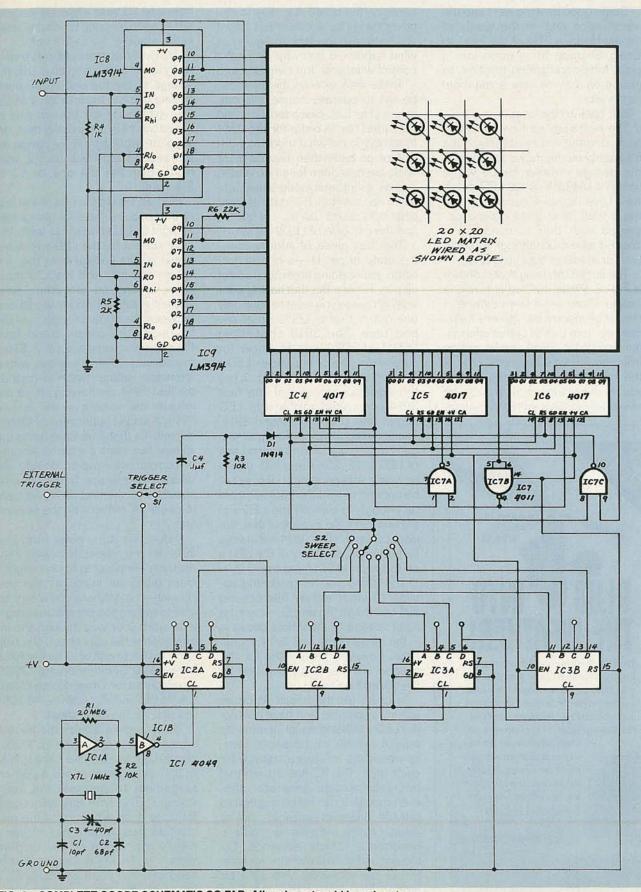


FIG. 3—COMPLETE SCOPE SCHEMATIC SO FAR. All we have to add is an input amp and a pad arrangement to prescale the input voltages.

May 1992, hadio-Electionics

Some good thinking went into the modification and, in the words of Jimmy Hatlo, a "Tip of the Hat" to Craig Shippee for a good job, a great letter, and taking the time to pass it on. Let me hear some more from you.

But back to the scope.

The next step we have to take is to expand the vertical display to the full twenty elements we specified in the design criteria. Since we're using an LM3914 as the LED driver, we'll have to add a second chip to the circuit. The 3914's were designed with that in mind, so it doesn't take much though to string two (or more) of them together.

The circuit for doing that is shown in Fig. 2. While the actual wiring is simple, there are a few points that should be made a bit clearer. I'll also mention here that all of that information (and a lot more) can be gotten from the 3914's data sheet. One of the most important things when you're designing circuits is to be absolutely familiar with the compo-



nents on the board. It's okay to discover things by accident but only if you then take the time to figure out what happened and why. You can't control what you don't understand.

There are four ways the 3914 can be set to operate: single-chip dot, single-chip bar, cascaded dot, and cascaded bar. In order for the 3914 to reliably know what mode to work in (dot or bar) when two or more chips are cascaded for an expanded display, its internal mode-select amplifier has to watch the state of three pins: MODE SELECT (pin 9), + V (pin 3), and the cathode of LED9 on pin 11.

That last piece of informationthe state of pin 11-is critical only when you're doing an expanded dot display. Remember that having a dot display means you want to have only one out of twenty LED's on at any one time. The 3914 controlling LED11-LED20 has to know for sure when one of the earlier LED's is being lit. The circuit in Fig. 2 has the MODE CONTROL pin of the first 3914 connected to the first LED output (pin 1) of the second 3914. That's because when the input signal is driving one of the second bank of LED's (12-20), there will always be some voltage at pin 1. It may not be enough to light the LED but it will be enough to turn off the LED's in the first 3914. As a result of that, the last LED on the first 3914 will always be turned off when any of the LED's on the second 3914 are being lit.

The same sort of reasoning applies to what keeps the second 3914 from lighting an LED when the input voltage to the whole circuit is in the range of the first 3914. In that case, however, the control for the second 3914 is being provided by the voltage on pin 11.

In order for the circuit to be as linear as possible and have each of the LED's indicate equal increments in input voltage, you have to be careful about the reference voltage for each chip. The IC has an internal reference-voltage generator and, even though it can be configured to provide different voltages, we're using it in the plain vanilla mode to generate a 1.2-volt reference. On the second IC in the chain, we have to set the reference voltage a bit higher. Remember that the first 3914 has to respond to input signals from 0 to 1.2 volts while the sec 3914 has to respond to input sign in the range of 1.2 to 2.4 volts.

That sounds more difficult tha really is. All we have to do is use voltage at the top end of the or parator chain in the first 3914 as low voltage for the bottom end the second 3914. By doing that, absolute working-voltage range the second 3914 will be 1.2 v higher than the first one, or 1.2 2.4 volts.

This isn't quite the end of the s ry because we have to give sc thought to LED current as well. drive current for the LED's is de mined by the voltage from the rail on one side and the IC's re ence voltage on the other si There's a ratio of about ten-to-c between the current for each of LED's and the current drawn fr the reference voltage at pin 7. Si the second 3914 is working wit reference voltage twice as high the first 3914, we have to adjust value of the resistor on the seco 3914. A simple application of Oh law tells us that if we have twice voltage but want the same amo of current, we have to double value of the resistor. Since we us about 1K on the first 3914, we have to use a 2K resistor on the seco one.

Before we turn away from 3914, let me repeat that while it relatively simple chip to use, ther more going on inside it than y imagine. The only way you'll ever able to get a good handle on usin is to work your way through the formation in the data sheet. The v to do that is to call or write the fo at National Semiconductor (29 Semiconductor Drive, Santa Cla CA 95052-8090, 408-721-50( and ask for the data sheet.

The complete schematic for scope so far is shown in Fig. 3. J about the only thing we have to a to it is an input amp and a pad rangement to prescale the in voltages. The time has also co for us to deal with some of the r chanical problems in the scope mainly the wiring of the disp While it's certainly not impossible hand wire four hundred LED's, it lot easier to use a commercia available multi-LED module.

Radio-Electronics, May 1992

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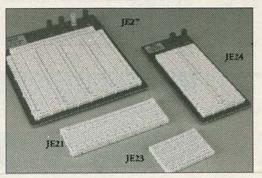
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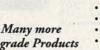
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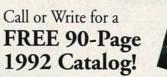
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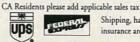
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 New B (Buried) - VSIS structure incorporating high resistance GaALAs crystal buried at both sides of the optical waveguide (V-

channel) preventing carriers in the active layer from diffusing from the direction parallel to the active layer, thus dramatically reducing wasted current not contributing to laser oscillation. This achieves a very low threshold current level of only 25mA. •Compact and lightweight, essential when giving laser-applied equipment-like CD players, VD players and laser printers - a small space saving footprint.

LT Package Dimensions

Wavelen		Wavelength	Optical Pwr Output (mW)		and the second se		
rder #	1-9	10+	(nm)	Typ.	Max.	Typical Applications	Weight
T020MC	13.90	12.90	780	3	5	Measuring instruments, Bar code readers	3 oz.
T022MC	13.90	12.90	780	3	5	CD players, CD-ROMs	3 oz.
T023MC	16.50	15.50	780	3	5	VD players, Analog devices	3 oz.
T021MC	55.00	50.00	780	10	15	High-speed laser printers, Medical apparatus	3 oz.
T010MC	49.00	40.00	810	3	5	Short-distance optical fiber communications, Measuring instruments	3 oz.

#### RACK MOUNT CABINETS

·Racks B, C, D, E and H feature extruded aluminum handles



op/Bottom, Side Panel and Chassis: 0.9mm Steel

nese industry standard 19" rack mount cabinets are made from .9mm steel, with the front nd rear panels being made from aluminum for easier, custom fitting. The front panel is 1/8" ick. These cabinets have been finished in gun metal grey and come flat packed, ready for our final construction. The enclosure has die punched venting. Handles are furnished on " and larger cabinets. These are sturdy, well made commercial quality cabinets and are ire to make your project look its best. (Please note "Panel" size is the front panel.)

rder #	1-9	10+	Panel	Cabinet (HxWxD)	Weight
3120A	36.95	34.95	19" x 1.6875"		7.0 lbs.
3120B	44.95	42.95	19" x 3.4375"		7.8 lbs.
3120C	49.95	47.95	19" x 3.4375"		9.0 lbs.
3120D	54.95	52.95	19" x 5.1875".		
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B630	5.45	4.95	6.5 x 1.4 x 0.4	0	0		630	0
B830	6.49	5.99	6.5 x 2.2 x 0.4	2		1	630	0
B1360			8.5 x 3.9 x 1.2					
B1660	17.45	16.95	8.5 x 5.1 x 1.2	4	400	2	1,260	3
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#### **DISPLAY DRIVERS**

#### continued from page 72

als do not have to be preaped. The counter is reset by iving the RESET pin high.

The CLOCK INHIBIT pin must be ounded to allow normal unting. When CLOCK INHIBIT is gh, the counters are inhibed. The display is blanked hen the display ENABLE pin is ounded.

The three auxiliary output ns of the 4026B are desigated DISPLAY/ENABLE OUT (D/E JT), CARRY OUT (CARRY OUT), 1d 2 OUT (2 OUT). The DISPLAY/ JABLE OUT signal is a slightly elayed copy of the DISPLAY/EN-BLE input signal. The CARRY JT signal is a symmetrical juare wave whose frequency is ne-tenth of the clock input frelency, and is used when casding 4026B counters. The 2 JT pin goes low only on a count two. Fig. 22 shows the circuit onnections for cascading ages.

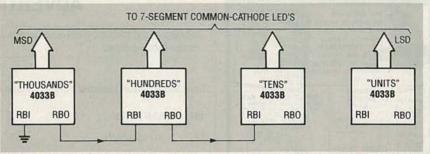


FIG. 25--MODIFICATION of Fig. 24 circuit to obtain automatic leading-zero suppression.

Figure 23 shows the pinout pattern of the 4033B, a modified version of the 4026B. The DISPLAY ENABLE and DISPLAY EN-ABLE OUT pins of the 4026B have been eliminated and replaced by RIPPLE-BLANKING INPUT (RBI) and RIPPLE-BLANKING OUTPUT (RBO).

Figure 24 shows the wiring scheme for two 4033B's in normal use. The RESET, CLOCK INHIB-IT, and LAMP TEST pins are all grounded, and the RIPPLE-BLANK-ING INPUT pin is positive. That configuration does not blank leading or trailing zeros.

If cascaded 4033B's are to give automatic leading-zero suppression, the circuit in Fig. 24 must be modified as shown in Fig. 25 to provide ripple-blanking. In Fig. 25, the RBI pin of the most significant digit (MSD) is grounded, and its RBO pin is connected to the RBI pin of the next least-significant stage. That pattern is repeated on all except the LSD, which does not require zero suppression. If trailing-zero suppression is required, the direction of rippleblanking feedback must be reversed. The RBI pin of the LSD is grounded and its RBO pin is wired to the RBI pin of the next least-significant stage. R-E

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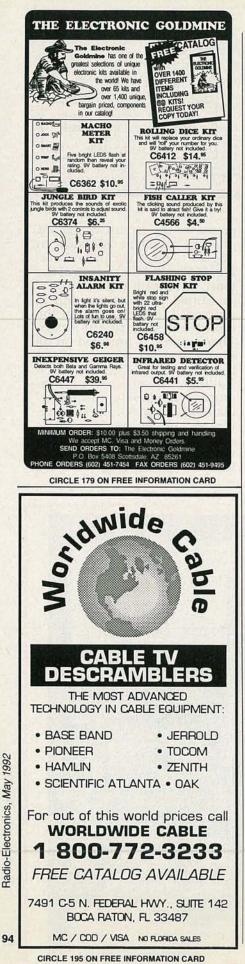
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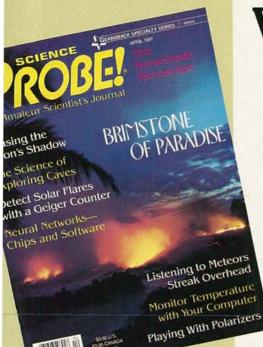
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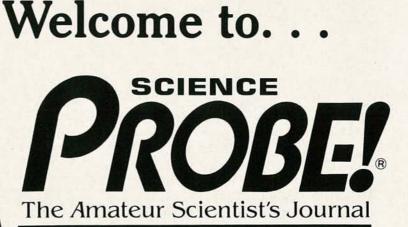
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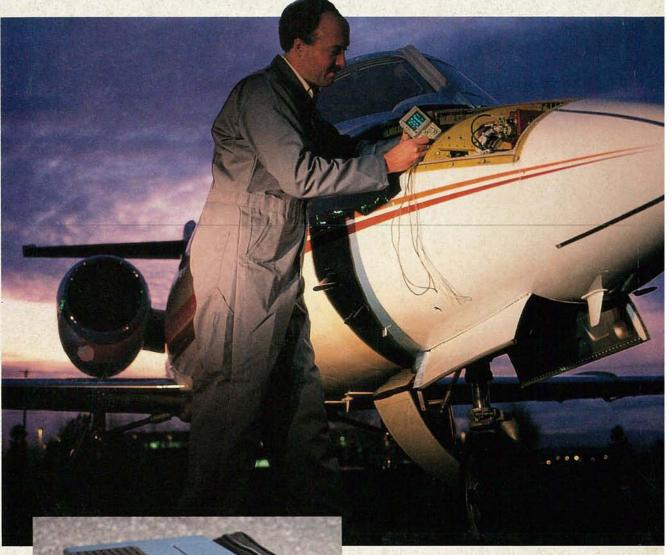
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