

SPECIAL SECTION
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Radio- Electronics

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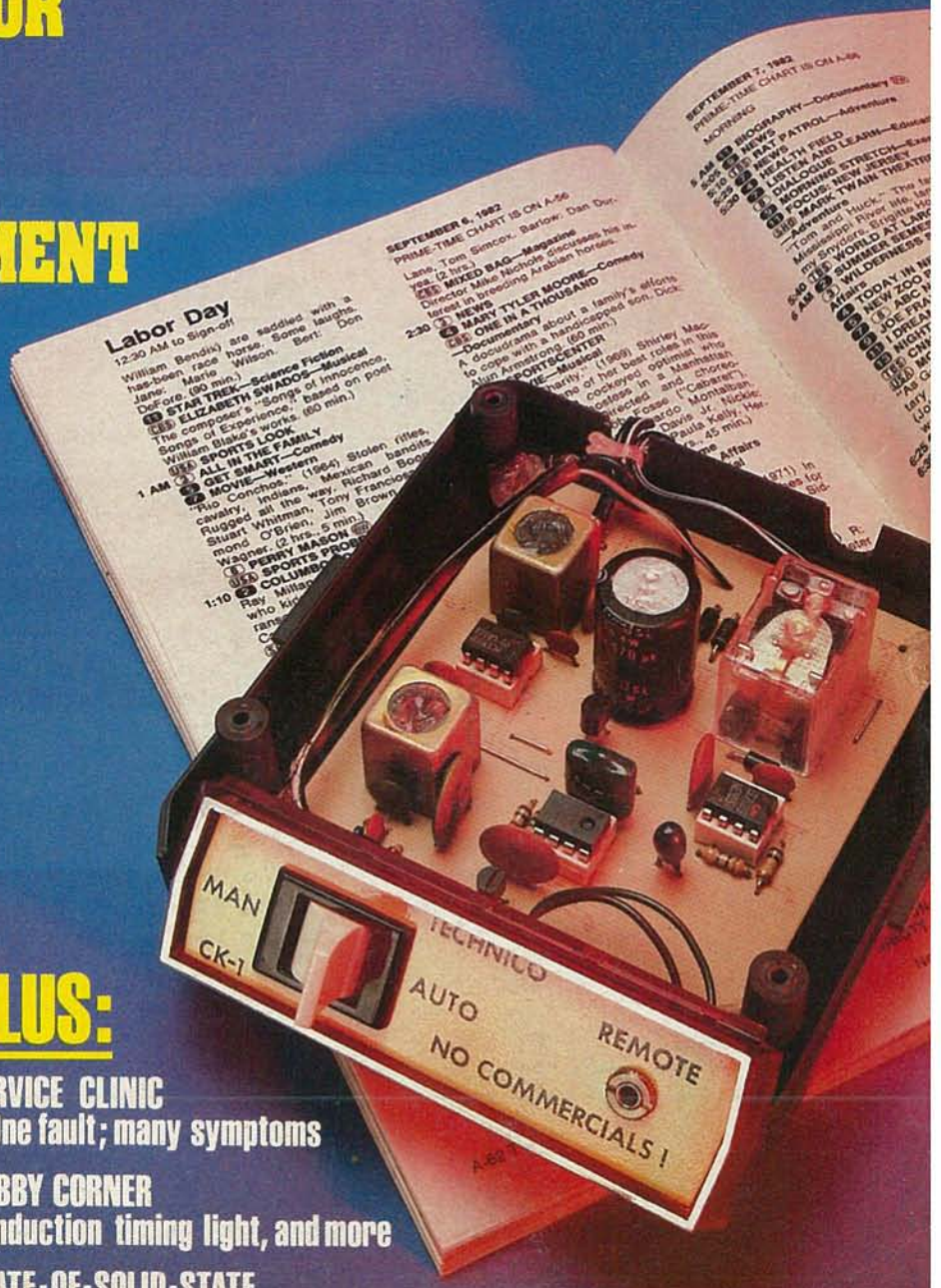
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COMMERCIAL EDITOR
for your VCR

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IN THE HOME**

- ★ The Entertainment Center
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- ★ New Video Components
- ★ Video Signal Processors
- ★ How to Connect Video Components

How to
ET PC BOARDS

Back-to-school series
**MULTI-STAGE
AMPLIFIERS**



Labor Day

12:30 AM to Sign-off

William Bendix) are saddled with a

has been race horse. Some laughs. Don

Jane. Marie Wilson

Defore. (90 min.)

STAR TREK—Science Fiction—Musical

ELIZABETH SWADLO—Musical

Songs of Experience. (80 min.)

SPORTS LOOK

William Bledsoe's works. (80 min.)

ALL IN THE FAMILY—Comedy

GET SMART—Comedy

MOVIE—Western. (1964) Stolen titles,

"Rio Conchos." Indians, Mexican bandits,

cavalry, Indians, Richard Boone

Rugged all the way. Richard Boone

Stuart Whitman, Jim

mona, O'Brien, Jim

Wagner. (2 hrs., 5 min.)

PERRY MASON

SPORTS PROGRAM

COLUMBO

Ray Milland

who kills

range

Case

1:10

SEPTEMBER 6-1982

PRIME-TIME CHART IS ON A-56

Lane, Tom Simcox. Barlow, Dan Du-

Yes. (2 hrs.)

FIXED BAG—Magazine

Director Mike Nichols discusses his in-

terest in breeding Arabian horses.

NEWS

MARY TYLER MOORE—Comedy

Documentary

ONE IN A THOUSAND

Documentary

A documentary about a family's efforts

to cope with a handicapped son, Dick

Armstrong. (60 min.)

SPORT SCENTER

Documentary

Shirley Mac

to her best roles in the

rocky optimist, and

costars in a Manhattan

("Casper").

Montalban

Directed by

Davis Sr. Node

Paula Kelly He-

(45 min.)

SEPTEMBER 7-1982

PRIME-TIME CHART IS ON A-56

NEWS

BIography—Documentary

NEWS PATROL—Education

HEALTH AND LEARN—Education

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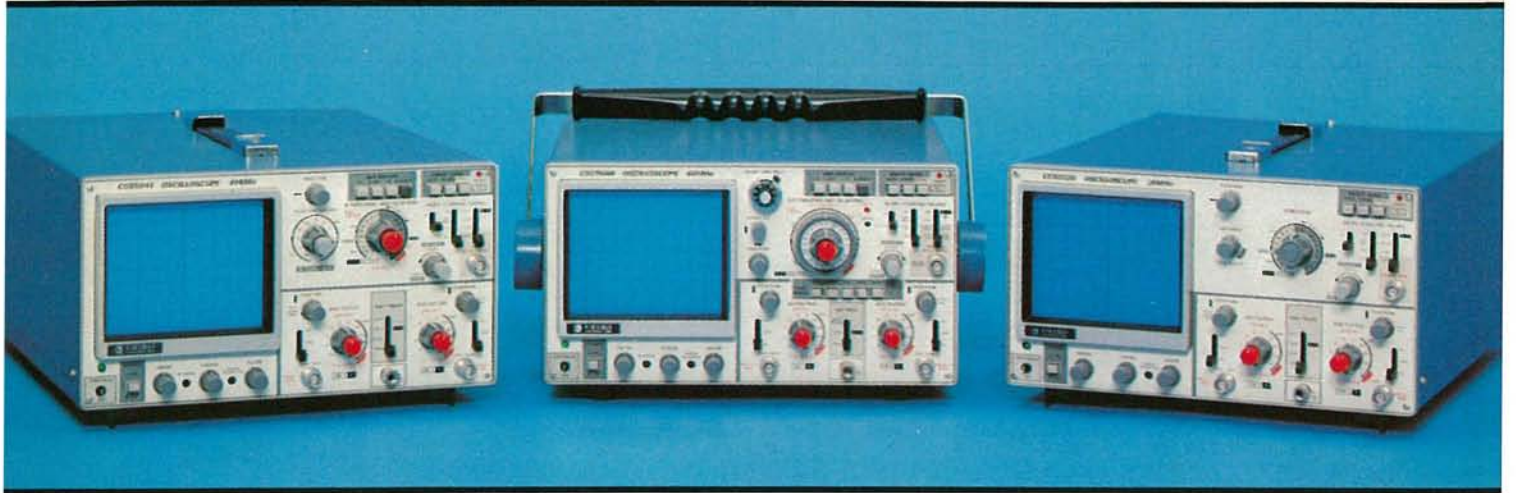
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Product Description	Part #	CE quant. 100 price per disc (\$)
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8" SSSD Shugart Compatible, 32 Hard Sector	3015	2.09
8" SSSD CPT 8000 Compatible, Soft Sector	3045	2.99
8" SSDD IBM Compatible (128 B/S, 26 Sectors)	3090	2.74
8" DSDD Soft Sector (Unformatted)	3102	3.34
8" DSDD Soft Sector (128 B/S, 26 Sectors)	3115	3.34
8" DSDD Soft Sector (256 B/S, 26 Sectors)	3103	3.34
8" DSDD Soft Sector (512 B/S, 15 Sectors)	3114	3.34
8" DSDD Soft Sector (1024 B/S, 8 Sectors)	3104	3.34
5 1/4" SSDD Soft Sector w/Hub Ring	3481	2.34
5 1/4" SSDD 10 Hard Sector w/Hub Ring	3483	2.34
5 1/4" SSDD 16 Hard Sector w/Hub Ring	3485	2.34
5 1/4" DSDD Soft Sector w/Hub Ring	3491	3.09
5 1/4" DSDD 10 Hard Sector w/Hub Ring	3493	3.09
5 1/4" DSDD 16 Hard Sector w/Hub Ring	3495	3.09
5 1/4" SSQD Soft Sector w/Hub Ring (96 TPI)	3504	2.99
5 1/4" DSQD Soft Sector w/Hub Ring (96 TPI)	3501	3.99

SSSD = Single Sided Single Density; SSDD = Single Sided Double Density; DSDD = Double Sided Double Density; SSQD = Single Sided Quad Density; DSQD = Double Sided Quad Density; TPI = Tracks per inch.

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Memorex Quantum 2400 feet Easy Load II Cart.	27JR	16.99
Memorex Quantum 1200 feet Wrightline Seal	27FW	12.50
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NCR Cartridge	98-26600-32	160.00
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Flag Free	72-26600-03	320.00
Mark XI 200 MB. Error Free	03-35041	720.00
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Mark XII 200 MB. NCR/CDC Flag Free	03-39001-01	515.00
Honeywell Flag Free	03-39000-01	515.00
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SEASON'S GREETINGS

*The editors and staff
of Radio-Electronics
join in sending
holiday greetings and
our best wishes for
a happy new year*

ON THE COVER

If you like to tape those vintage black-and-white movies that show up late at night on TV—but don't like to tape the commercials that accompany them—this automatic commercial editor is for you! It watches the movie along with your VCR and, when a commercial turns up, stops the tape until the movie begins again. The result is a tape of the movie, and nothing else. It's easy to build, and will...literally...provide you with hours of pleasure. Get started building your own editor. Plans begin on page 43.

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VIDEO ELECTRONICS

DAVID LACHENBRUCH
CONTRIBUTING EDITOR

STEREO TV HITS SNAG

The dream of stereophonic sound for television in 1983 vanished in a puff of hostile smoke when an industrywide committee studying the subject went back to the drawing board instead of recommending a system. Since 1979, a subcommittee of EIA's Broadcast TV Systems Committee has been examining and testing three proposed multichannel TV sound systems. Each of the systems—proposed by EIA of Japan, Telesonics Systems Inc., and Zenith Radio Co.—met the subcommittee's basic criteria: (1) Stereo sound quality comparable to FM stereo. (2) An additional "picture-related" sound channel for such uses as foreign-language translation. (3) Compatibility with existing monophonic-sound receivers.

The subcommittee completed its tests of the three systems on time, and its timetable called for it to select a system for recommendation to the FCC in mid-October. Everything was going according to plan when the subcommittee's steering committee met to tally members' votes for a system. However, it was confronted unexpectedly by a charge from Telesonics Systems that the tests were incomplete. Telesonics requested that the subcommittee reopen its record—and implicitly threatened legal action unless the subcommittee agreed. Accordingly, the subcommittee decided to reopen the entire subject to cover the Telesonics objection, as well as another problem. That new problem related to the discovery that most cable-TV systems would be unable to relay stereophonic sound under any of the systems, and, in fact, the presence of a stereo signal could disrupt the carriage of mono programs on cable by causing adjacent-channel interference.

The delay could be six months to a year, or longer—if the subcommittee ever completes its work at all. At press-time, the group was preparing to ask the FCC to hold off any action on the issue of multichannel TV sound until new tests could be completed. But the FCC staff had already prepared a proposal for issuance by the Commission to permit TV stations to choose any stereo sound system which would meet certain criteria—a so-called "marketplace" decision similar to the one that permitted any of a number of AM stereo systems, all mutually incompatible, to be broadcast simultaneously. (The AM stereo order resulted in such confusion that few stations bothered to add stereo, and set manufacturers held off making AM stereo receivers.) TV broadcasters and set makers are concerned that the FCC will issue its proposal for "marketplace" stereo TV standards (meaning no standard) before the subcommittee can complete its new tests. If such an order is proposed by the FCC, the subcommittee will be dissolved, in view of legal opinions that it would be unlawful for representatives of competing systems to meet while the FCC is considering the issue. No matter what happens, it now appears there will be a long delay, and if you want to hear stereo sound with your TV you'll have to tune to a PBS simulcast or go to Japan or Germany.

BANK-BY-TV

The first major home banking and information system has been inaugurated commercially by Chemical Bank in the New York City area after a 10-month test in 200 homes. Chemical's *Pronto* system uses any home computer, a telephone modem, automatic phone dialer, and special program cartridge; the Bank will supply the entire receiving system (except the TV set), based on an *Atari 400* computer, at around \$500, but other home computers can be used.

Pronto customers, who pay a monthly fee of about \$10, are able to call up bank statements on their TV screens, transfer funds among different Chemical Bank accounts, pay bills by electronic fund transfer, and send electronic mail messages to other *Pronto* subscribers. The system plugs into a modular telephone jack and is connected to the bank's Tandem computer. Future services to be offered on the *Pronto* system include financial services such as portfolio management, teleshopping, educational courses, home budgeting, and security monitoring. Chemical hopes that its system will become nationwide, and has already presented seminars to 250 other banks and institutions.

The bank said that more than 250 major companies already are accepting electronic bill-payment via *Pronto*—including American Express, Master Charge, Visa, major oil companies, most department stores, Sears, utilities, insurance companies, and major New York landlords. In paying bills, *Pronto* users can specify regular monthly payments, and even which day of the month they wish bills to be paid. Privacy is assured by a triple-security system including household and personal ID numbers, as well as an individual password. Bank officials predicted that 10 million American households will have interactive devices attached to their TV sets by the end of the decade.

R-E

"No one else gives you as many functions in a handheld DMM."

Now you can move up to Fluke."

We've got great news for people who've been holding out for a high quality, high performance DMM at a moderate price: Fluke's new nine-function model D 804 is now available at select electronics supply stores.

With a suggested U.S. price of only \$249 and features you won't find in any other handheld DMM, the D 804 is an exceptional value. Here's why.

Logic level and continuity testing: A real time-saver for troubleshooting passive circuits in pcb's, cables, relay panels and the like. The D 804 has a switch-selectable audible tone and visual symbols to indicate continuity or logic levels.

Direct temperature readings in °C: Used with any K-type

thermocouple, the D 804 delivers fully-compensated readings in °C from -20°C to +1265°C, for checking heating and refrigeration systems.

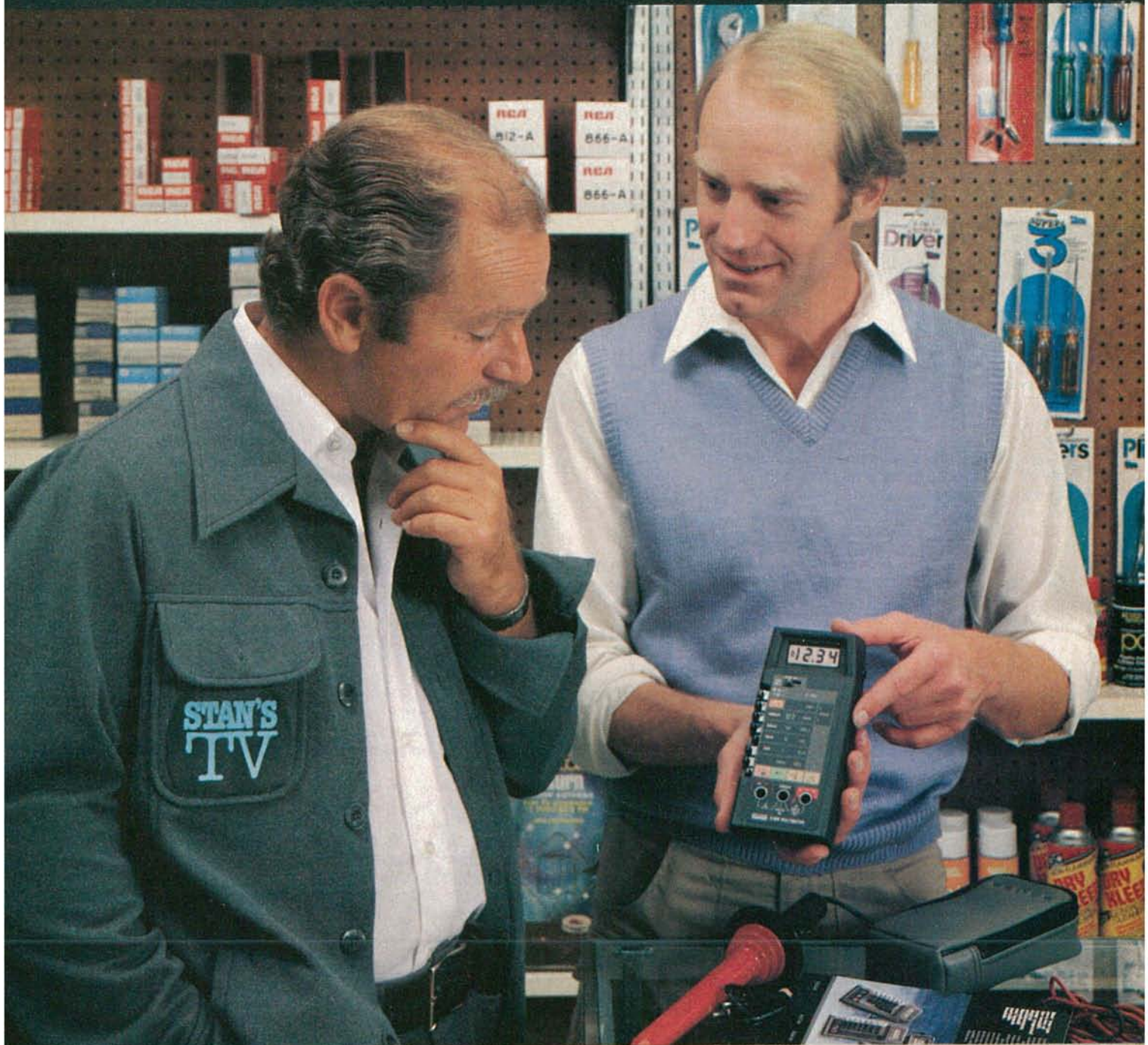
Peak hold feature captures transients: A short-term memory in the D 804 captures and holds the peak reading of a motor starting current.

And more: 0.1% basic dc accuracy, conductance, 26 measurement ranges, battery, safety-designed test leads and a one year parts and labor warranty. A full line of accessories is also available to extend the measurement capabilities of your DMM.

Ask your dealer about the powerful, versatile D 804 and the rest of Fluke's new Series D line of low-cost digital multimeters.



From the world leader in DMM's. Now we've designed one for you.



Suggested U.S. list price
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If your dealer doesn't carry Series D
Multimeters yet, call this number. We'll be
happy to tell you who does. 1-800-426-0361

FLUKE
®

WHAT'S NEWS

"Father of Television" is dead at age 93

Vladimir Kosma Zworykin died this past July 29, at the Princeton (NJ) Medical Center, just one day short of his 93rd birthday. Called "The Father of Television" by many, he always declined the honor, pointing out that hundreds had contributed to the art over many years. Invention, he said, is like building a ladder, and that as each engineer added a rung, "It enabled the others to climb a little higher and see the next problem a little better."

The development of modern television, however, is based on Zworykin's steady work from the 1920's onward, especially his invention of the iconoscope, the first practical camera tube.

Dr. Zworykin was born July 30, 1889, in Mourom, Russia. He attended the Petrograd Institute of Technology, receiving an Electrical Engineering degree in 1912. At the Institute he studied under Professor Boris Rosing, and was his lab assistant. Rosing believed as early as 1908 that the cathode-ray tube was the solution to practical television. Dr. Zworykin, who credits both his decision to become a scientist and his special interest in television to Boris Rosing's influence, followed that line of reasoning in his work on the iconoscope and kinescope.

Dr. Zworykin joined RCA in 1929, serving as Director of the RCA Electronic Research Laboratory in Camden, NJ until 1942, and afterward at Princeton, until his retirement in 1954.

Besides television Dr. Zworykin was interested in a wide variety of subjects, ranging from gunnery controls through multiplier tubes to electronically controlled automobiles. For a number of years after his retirement, he directed the Medical Electronics Center at the Rockefeller Institute in New York, working for the development of electronics methods in medicine and the life sciences.

Zworykin received his doctorate from the University of Pittsburgh in 1926. In 1966, President Johnson awarded him the United States' highest honor, the National Medal of Science, "for major contributions to the instruments of science and television, and for stimulation of the application of engineering to medicine." Altogether, he received 27 major awards and much other recognition from groups throughout the world.

Aircraft entertainment carried by infrared

A wireless-entertainment system using infrared light to carry the signal is being developed by the microelectronic systems division

of Hughes Aircraft Co. The new system is known as AIRE (Advanced Infra-Red Entertainment System). It uses a digital system transmitted throughout the plane by infrared rays. The signal is received and decoded by a headset worn by the passenger.

AIRE can transmit up to 16 channels. Each headset has a program-select and a volume control. A self-test feature assures that each unit remains in operating condition.

The use of infrared cuts the weight of the full system to less than half that of a conventional entertainment system, and the cost by one-third. Another great advantage of the infrared approach is that—unlike earlier wireless-entertainment systems—AIRE will not interfere with other aircraft equipment. It is expected that the new system will be on the market by the middle of 1983.

KEYFAX, SSS Systems plan teletext magazine

KEYFAX Electronic Publishing and Satellite Syndicated Systems have announced that KEYFAX National Teletext Magazine will be ready for delivery to cable households shortly. This is stated to be the first large-scale teletext consumer operation in the United States, and is expected by its founders to grow to 200,000 cable households within 30 months.

The new teletext magazine will be delivered via the vertical blanking interval (VBI) of the WTBS satellite service, which is available to more than 20 million cable households.

The service will be offered to subscribers at a suggested \$19.90 per month. That covers \$9.95 for the service itself and \$9.95 for rent of the necessary decoder. Operators will retain \$3.95 per month as revenue (\$4.95 after penetration to 1 percent of their system's basic subscribers).

KEYFAX has been on the VBI of WFLD-TV in Chicago since April 1981, on a test basis. It has also been on the VBI of the WTBS satellite service since May 1982, and was displayed (via satellite and then through two TV stations) at the Knoxville World's Fair.

N.J. State Police improve their radars

The State Police of New Jersey have recently publicized their acquisition of "beam-interrupter" switches for their radar units. Those switches, when fitted to the devices, permit cutting off the radar beam without turning off the equipment. When a car suspected of speeding is sighted, the beam can be turned on instantly. If he has a radar detector, the speeder is not warned until police actually start taking a reading.

Manufacturers of radar detectors state that such a device is only minimally effective: "Once the officer activates the unit to obtain a speed reading on a vehicle, every detector-equipped motorist within its range will receive a warning. ... The beam will alert all detector-equipped vehicles within several miles."

The State Police believe that, in spite of any claimed deficiencies, the beam interrupters are effective, and point out as evidence of that effectiveness that manufacturers have begun to install the beam interrupters on new radars as a standard equipment.

Video programs aid technical training

RCA's Corporate Engineering Education (CEE) unit at Cherry Hill, NJ, reports that a company-wide program of video courses has simplified the problem of keeping engineers, managers, and other technical personnel at RCA up to date on the very latest technological advances.

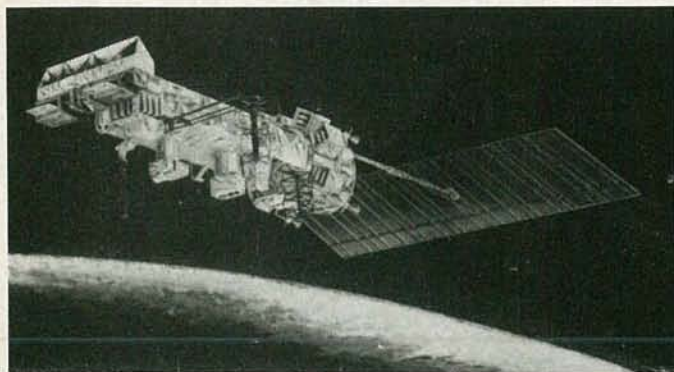
The company has recently begun to augment the video programs with live courses on subjects requiring extensive hands-on exercises and courses for small group projects.

Most video courses consist of 12 two-hour sessions, conducted as discussions, exercises, and hands-on activities supervised by an associate instructor.

In addition to the more than 70 video courses listed in CEE's catalog, CEE maintains a library of videotapes of technical seminars and talks; the library is maintained so that the tapes can be borrowed and viewed by RCA personnel at their convenience.

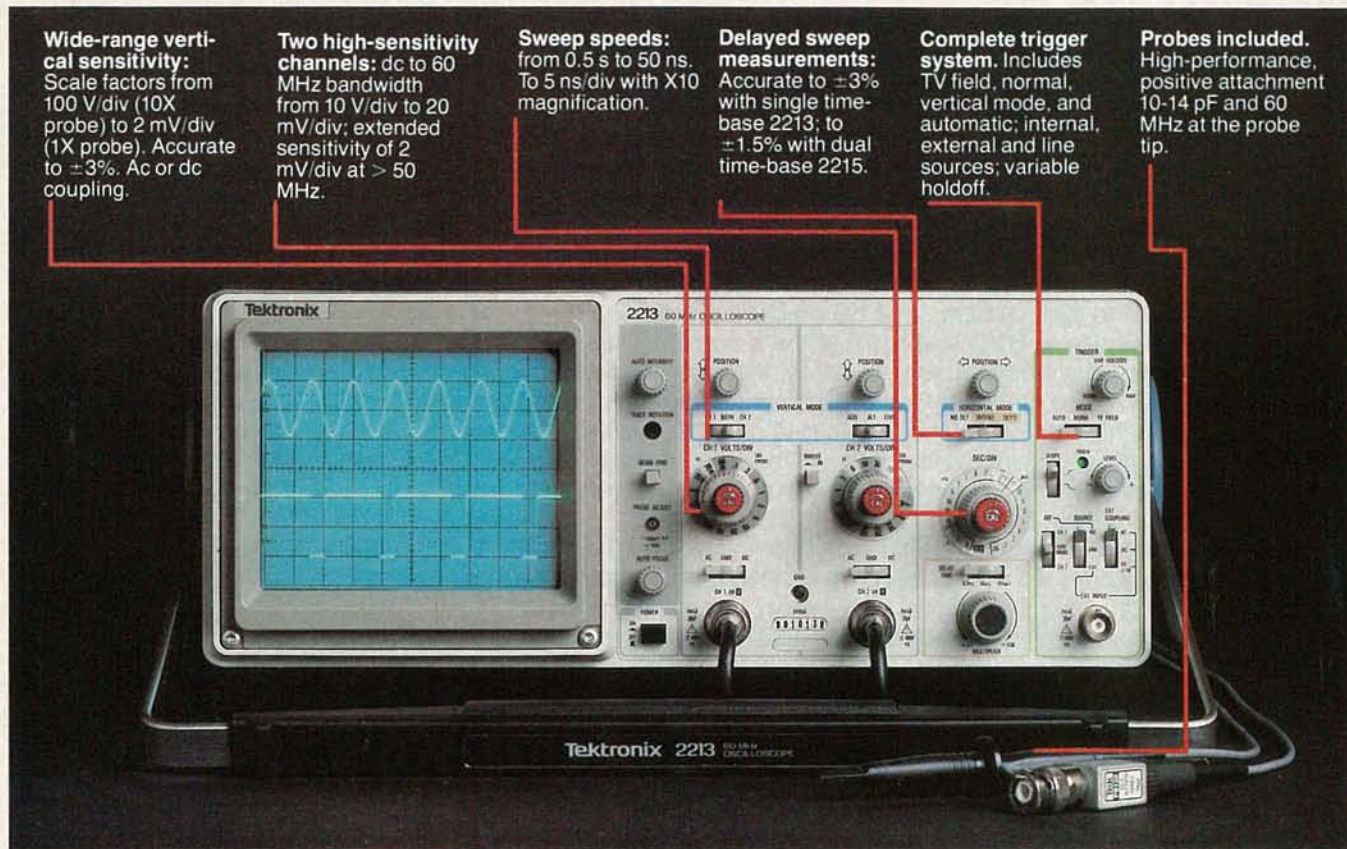
R-E

NEW METEOROLOGICAL SATELLITE



TIROS-N (ATN) METEOROLOGICAL SATELLITE is the first of a series of advanced TIROS-N/NOAA meteorological satellites that is being designed and built for the National Oceanic and Atmospheric Administration (NOAA). In addition to its usual weather-charting duties, the satellite will be used to relay distress calls from the emergency beacons of downed aircraft and ships in distress. Launch is scheduled for February 1983. NASA will act as program manager for the mission.

Now! A 60 MHz Tektronix scope built for your bench.



In 30 years of Tektronix oscilloscope leadership, no other scopes have recorded the immediate popular appeal of the Tek 2200 Series. The Tek 2213 and 2215 are unapproachable for the performance and reliability they offer at a surprisingly affordable price.

There's no compromise with Tektronix quality: The low cost is the result of a new design concept that cut mechanical parts by 65%. Cut cabling by 90%. Virtually eliminated board electrical connectors. And obviated the usual cooling fan.

Yet performance is written all over the front panels. There's the bandwidth for digital and analog circuits. The sensitivity for low signal measurements. The sweep speeds for fast logic families. And delayed sweep for fast, accurate timing measurements.

The cost: \$1200* for the 2213. \$1450* for the dual time base 2215. You can order, or obtain more information, through the Tektronix National Marketing Center, where technical personnel can answer your questions and expedite delivery. Your direct order includes

probes, operating manuals, 15-day return policy and full Tektronix warranty.

For quantity purchases, please contact your local Tektronix sales representative.

**ORDER TOLL FREE
1-800-426-2200**

Ask for Department J0147
In the state of Washington,
Call (206) 253-5353 collect.

*Price F.O.B. Beaverton, OR. Price subject to change.

PUBLISHER'S LETTER

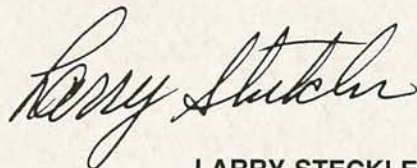
This December issue of **Radio-Electronics** magazine brings the year 1982 to a close. It has been an exciting one. We delivered four special sections during 1982—"Video Electronics" in January, "Your Own Computer" in April and again in October; plus a "Video Games" section in July. Along the way we were able to publish more than 150 articles covering a variety of subjects—audio, video, computers, radio, construction, how-to—to name just some of the varied areas we covered. We also logged a couple of exceptional construction stories—we showed how to build your own satellite TV receiver; and a first for home constructors: your own picture phone.

In 1983 we promise to bring you more of the same. The same quality, the same wide range of subject areas, the same first-rate electronics that kept you reading **Radio-Electronics** in record numbers all the way through 1982. We've enlarged our staff, we've added more pages to the magazine, and we've added a sister publication, **Special Projects**, that is specially designed to cater to the needs of the project builder.

However, there are some things we won't do. We will *not* become a computer magazine; we will *not* become a video-games magazine, and we will *not* dive into and follow blindly every one of the paths in electronics that is bound to develop during 1983. We *will* remain the broad-based, all-encompassing electronics magazine that we have been for the past 53 years. Each year we learn a little more, and that makes us better.

One bit of bad news, however—subscription prices are going up starting in January 1983. They are going up an average of \$2.00 per year. So if you want to save yourself a few dollars, use the subscription-order form in this issue—it's inside the back cover. Next month the prices on that card will be higher.

I'd like to thank you for your continued support and pledge that we will continue to do our very best to make **Radio-Electronics** the finest "electronics" magazine you read.



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A FREE 9100 Electronic Soldering System when you buy either desoldering system by December 15, 1982.



HOT VAC® MODEL 4091 PORTABLE DESOLDERING SYSTEM WITHOUT SHOP AIR. SUGGESTED RETAIL PRICE \$399.00.

Take this ruggedly constructed, light-weight unit anywhere—plugs into any 120V outlet. Solid-state control adjusts tip temperature from 500° to 1000°F and suppresses transient spikes.

High-flow vacuum is activated by a switch mounted in a handle that's biomechanically designed to reduce fatigue and increase productivity.

Super quiet and virtually maintenance-free, with Ungar's exclusive low-noise exhaust and unique vibration-absorbing pump mounting. All built into a galvanized steel case with a scratch-resistant, baked enamel finish.

Comes complete with: Long Life Tips #2005, #2006, #2007, #2008; Tip Cleaning Tool #2002; Kleen-Tip Sponge and Tray #400; and Pack of 10 Filters #2039.



GET A FREE MODEL 9100 ELECTRONIC SOLDERING SYSTEM. SUGGESTED RETAIL PRICE \$91.00

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Includes neon "on/off" light, right- or left-hand mount and flexible, heat-resistant cord.

Comes complete with: base unit with built-in controller, handle, 1/16" screwdriver, tip #9012, sponge and tip tray, sponge and iron holder.

HOT VAC® MODEL 2091 WORKS ON SHOP AIR FOR ECONOMICAL DESOLDERING. SUGGESTED RETAIL PRICE \$365.00.

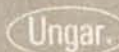
A solid-state control adjusts tip temperature from 500° to 1000°F. The load modulated heater assures instantaneous recovery and suppression of transient spikes.

The switch that activates the high-flow vacuum is conveniently built into a handle that is biomechanically designed to reduce fatigue and increase productivity.

Comes complete with: Long Life Tips #2005, #2006, #2007, #2008; Tip Cleaning Tool #2002; Free-Standing Holder #8800 with Kleen-Tip Sponge; and Tray #400. Pack of 10 Filters, #2039.

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Offer good until December 15, 1982.



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CIRCLE 9 ON FREE INFORMATION CARD

SATELLITE/TELETEXT NEWS

GARY ARLEN
CONTRIBUTING EDITOR

SATELLITE APPLICATIONS

The FCC has adopted a new procedure for accepting applications for satellite orbital slots, tailored for use with the pending proposal to push the slots closer together. The new plan, which probably will give preference to the 10 companies that have already filed applications, calls for proposals submitted by mid-May to be considered first in any processing procedure. After those projects have been acted upon, the FCC will consider new applications. The action reduces the likelihood that the FCC will conduct competitive hearings if two companies seek the same orbital slot.

MORE SATELLITE DATA SERVICES

There's more competition than ever for satellite space—and some of the major computer companies are now joining the fray. For example, Satellite Business Systems—the partnership of IBM, Comsat General, and Aetna insurance—is setting up a new medium- to high-speed Data Network Service, designed for customers with lower-volume data communications requirements. In a separate development, American Satellite Company and Tandem Computers Inc. are jointly developing the world's first computer network with full-integrated satellite transmission. The ASC-Tandem project, called Infosat, will offer continuous online transaction processing, distributed data processing, and information systems management.

AROUND THE SATELLITE CIRCUIT

Legal action on many fronts involving satellite reception: satellite master antenna TV operators in several cities are suing program suppliers such as Showtime, The Movie Channel, and others, for failing to deal with them to carry shows; in some cases the SMATV operators contend that pressure from local cable systems has prompted the retrenchment by programmers. Meanwhile, in a case that could set a precedent for satellite-TV reception, a federal court in Washington came down hard on a local dealer who sold unauthorized receivers designed to pick up HBO shows on multipoint distribution service microwave channels. The court awarded more than \$102,000 in damages to Marquee TV Network, the local MDS-HBO service, resulting from the "illegal conduct" of a small local company which advertised, sold, and installed MDS receivers; the award was the highest amount ever granted in a pay-TV piracy case, and represented reimbursement of funds which Marquee claimed it lost because of the unauthorized devices, which could pick up its HBO signal. The court also granted permanent injunctions against several other companies that sold similar reception equipment.

WORLDWIDE CNN

Cable News Network may become an international service, transmitted via Intelsat, if CNN's Ted Turner has his way. Turner is negotiating with Comsat for access to the Intelsat system so that CNN can be sent to South America, Africa, Europe, and Asia. CNN has already signed a deal with a Japanese cable TV company to carry CNN in Japan 17 hours per day as soon as an overseas transponder lease is completed. In return, CNN will begin picking up a daily round-up of Japanese business news, fed by satellite from Tokyo.

"DISH STRETCHER"

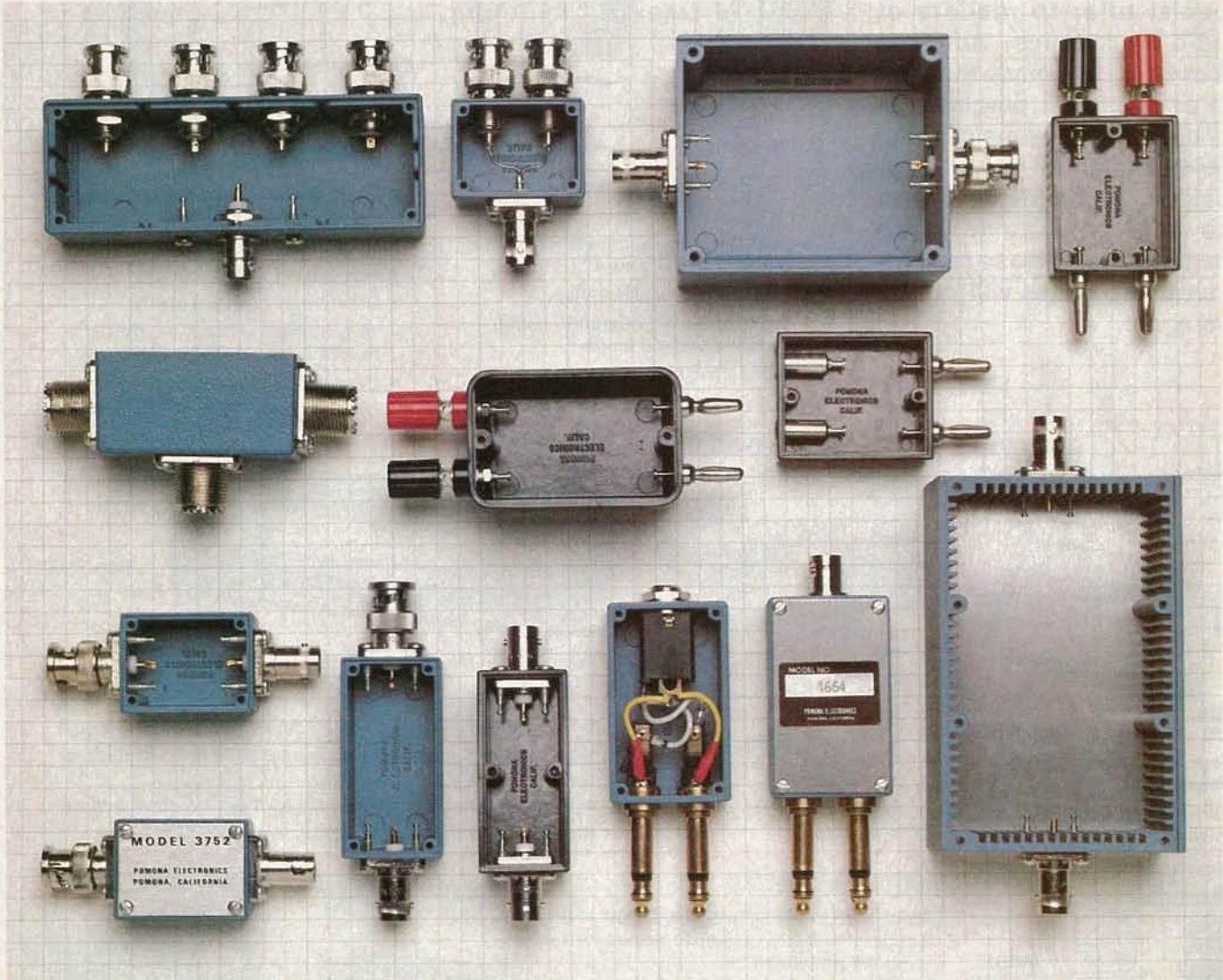
Imagine a little red wagon—the kind kids play with—equipped with a four-foot satellite dish, a 12-volt battery for power, 100° LNA, and a satellite receiver. Tulsa-based Automation Techniques Inc. called this portable set-up "Toysat"—and used the micro-miniature arrangement to emphasize the potential value of its new electronic "Dish Stretcher." That unit is intended to erase "sparklies" as well as the other kinds of interference that often afflict small-dish receiver systems.

According to ATI, the level of sparklies or impulse noise in a satellite picture has been a function of either the size and efficiency of the antenna, noise figure of the LNA/receiver, or the carrier-to-noise threshold of the receiver. ATI attacked the sparklie problem in a fourth area of the receiver system: detected video signal. The sparklies generated because of the smaller dish are identified by Dish Stretcher technology as improper video information and are removed from the picture signal after entering the receiver. The \$500 unit is available from ATI, 1846 N. 106th E. Ave., Tulsa, OK 74116.

R-E

THE POMONA PROMISE

We provide the design engineer with the best
"black boxes" made in this country.
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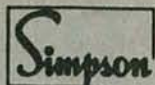
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\$329⁹⁵
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\$395⁰⁰
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PRICE DOES NOT INCLUDE PROBES

- Triggered and automatic sweep
- 18 calibrated sweeps
- On 1476 mode automatically shifts between CHOP and ALTERNATE
- Bright P31 blue phosphor
- Front panel X-Y operation using matched vertical amps
- Video sync separators
- Check most digital logic circuitry

BK PRECISION 15 MHz TRIGGERED SWEEP SCOPE



MODEL 1477

\$495⁰⁰
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PRICE DOES NOT INCLUDE PROBES

- Mode automatically shifts between CHOP and ALTERNATE
- Bright P31 blue phosphor
- Front-panel X-Y operation
- Differential input capability
- 19 calibrated sweeps—.5 μ SEC/cm to .5SEC/cm
- Sweep to .1 μ SEC/cm with 5x; 1.5SEC/cm with uncalibrated vernier

Simpson FUNCTION GENERATOR



MODEL 420

\$169⁹⁵
Reg. \$194⁰⁰

- Sine, square and triangle output
- Variable and fixed TTL outputs

Simpson DIGITAL MULTIMETER

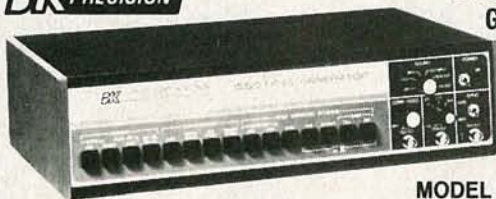


MODEL 467

\$229⁹⁵
Reg. \$259⁰⁰

- True RMS
- Analog display
- Peak hold

BK PRECISION NTSC COLOR-BAR GENERATOR



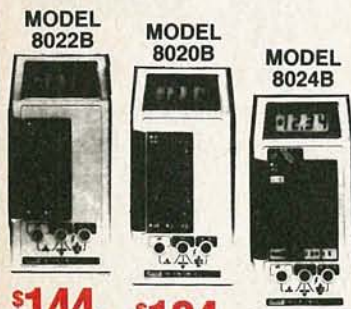
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Professional studio quality generator. Ultra stable. Ideal for VTR work.

- Generates NTSC color bars with or without—IWQ signal
- Generates 5 step linear staircase; staircase with high or low chroma
- External video input—modulates rf or i-f carrier outputs
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FLUKE HANDHELD DIGITAL MULTIMETER



MODEL 8022B

MODEL 8020B

MODEL 8024B

\$144. \$194. \$249.

8022B THE TROUBLESHOOTER ■ 6 functions ■ 0.25% basic dc accuracy ■ Overload protection

8020B THE ANALYST ■ 7 functions ■ 0.1% basic dc accuracy ■ LCD display ■ Safety design test leads.

8024B THE INVESTIGATOR ■ 9 functions ■ 0.1% basic dc accuracy ■ Peak hold on voltage and current functions ■ Selectable audible indicator for continuity or level detection.

FREE CASE FOR EACH FLUKE HANDHELD DMM

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MODEL 1601

\$279⁹⁵
was \$354.

- Isolated 0-50VDC, continuously variable; 0-2A in four ranges
- Fully automatic shutdown, adjustable current limit
- Perfect for solid state servicing



MODEL 1650

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- Functions as three separate supplies
- Exclusive tracking circuit
- Fixed output 5VDC, 5A
- Two 0 to 25VDC outputs at 0.5A
- Fully automatic, current-limited overload protection

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MODEL 1479B

\$595⁰⁰
was \$895.

PRICE DOES NOT INCLUDE PROBES

- Built-in signal delay line permits view of leading edge of high frequency pulse rise time.
- Triggers on signals up to 50MHz
- Rectangular CRT with P31 phosphor
- Mode automatically shifts between CHOP and ALTERNATE

BK PRECISION 70 MHz. Dual Time BASE SCOPE

MODEL 1570

\$995⁰⁰
WAS \$1570

PRICE DOES NOT INCLUDE PROBES

- 1 mV/division sensitivity to 70 MHz
- 500 μ V/division cascade sensitivity
- Four-input operation provides trigger view on 4 separate inputs.
- Alternate time base operation
- Switching power supply delivers best efficiency and regulation at lowest weight

FLUKE

4 1/2 DIGIT MULTIMETERS

MODEL 8060A



- Frequency measurements to 200KHz
- dB measurements
- Relative measurements
- True RMS
- High-speed Beeper

\$349⁰⁰**BK PRECISION** INDUSTRIAL TRANSISTOR TESTER**\$189⁹⁵**
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MODEL 520B

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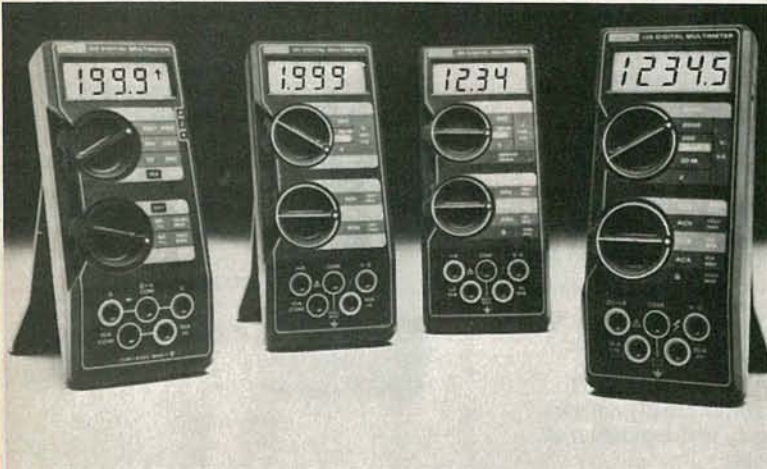
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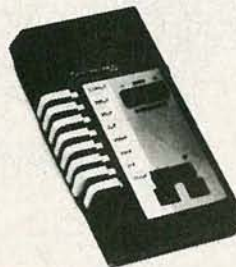
MODEL MS230

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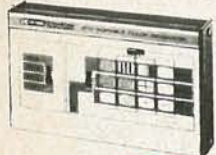


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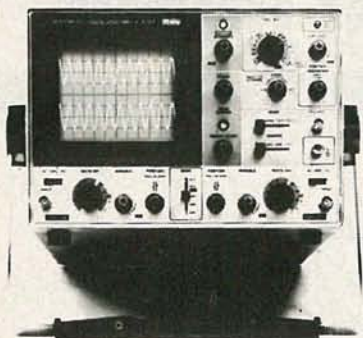
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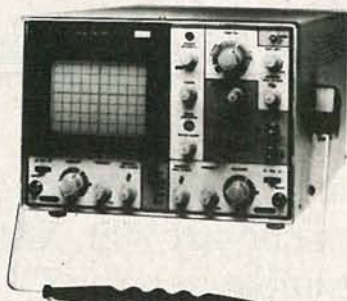
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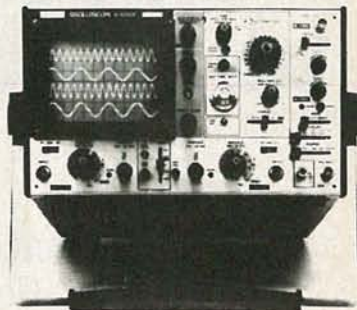
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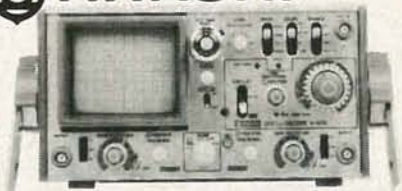
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LETTERS

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ROAD-INFORMATION SYSTEMS

I wish to respond to a letter in the July 1982 **Radio-Electronics** by Mr. Charles E. Koontz regarding road-information systems. First thing, I feel that such systems are really not needed in most areas of the United States. Just listen to any local radio station during the morning or afternoon drive-times, and you will hear plenty of traffic information. I think that the functions of roadside-information stations and auxiliary services that Mr. Koontz is discussing are already fulfilled by most radio stations. I also think that the FM system discussed in his letter is not as practical as the proposals for roadside stations that would have been located at the low end of the standard AM broadcast-band. Those AM stations could give better coverage and at lower transmitter power for the same area served. Also low-power and auxiliary services broad-

cast in FM would only be more difficult to receive in a moving automobile. I also wonder whether the public and the radio stations are interested at all in a system to transmit pictures over the existing FM band. That *sounds* interesting—but what real purpose could it serve?

Mr. Koontz must understand that the main-channel carrier space of a modern FM broadcaster is very valuable material. To the broadcaster, it is far more profitable to sell advertising time than to rent space on his carrier for SCA or other auxiliary services, possibly downgrading his signal. Mr. Koontz should remember that the broadcaster must give up about 10% modulation percentage for each auxiliary service. In most cases, that is not permissible. Note that most stations that carry SCA or other auxiliary services are non-commercial. The FM broadcast industry is more competitive now than it was 25 years

ago. Modern broadcast-equipment designers are really more concerned in allowing the broadcaster to get the most out of his main channel.

Now to a discussion on modern FM-broadcast technology: Mr. Koontz points out "reduced coverage" from Class-B and Class-C FM-radio stations; that is far from the real truth. Modern FM broadcasting antennas, built from the late 1960's to date, use a circularly polarized radiation pattern—that is, the antenna radiates in *both* the horizontal and vertical planes. Older antennas radiated only a horizontally polarized signal. Such horizontally polarized antennas are totally useless to the modern broadcaster. A majority of the FM receivers now used by the public use *vertically* polarized antennas. That includes portable radios and automobile receivers. A modern broadcaster is most concerned with "penetration"—the number of receivers that

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The Dyna-Mike Transmitter

It's smaller than a quarter. But DYNA-MIKE will transmit every sound in a room to an FM radio tuned to the proper unused frequency, from 1/3 mile to 2 miles away.

If you're at a neighbor's home a block from your own, you can hear your baby's cry, or you can tell the instant your spouse comes home. If two of you are driving tandem in two cars, one or both of you can communicate with the other even if other cars drive between you.

DYNA-MIKE has as many uses as your imagination can think of. For a business conference, let the tiny microphone sit unobtrusively on the table or concealed on a shelf, and you'll be able to record every word. For businesses, you can put an FM receiver in a warehouse or remote office and "broadcast" instructions or orders to be filled.

Public speakers never had a better friend than the DYNA-MIKE. No wires or setup — just turn on one or more radios and your speech will come through with perfect fidelity. Put one on the front porch. If you hear a suspicious sound, turn on the radio and you'll hear the doorbell or the ring of the telephone.

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New Horizons is introducing three models of the DYNA-MIKE supersensitive broadcast microphone. Model IC-18 is the world's smallest microphone — it's a miracle of electronic miniature power, with a high-fidelity range of 1800 feet. Introductory price is \$129.95 (two for only \$119.95 each).

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You'll hear it all.

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Ever put your ear to a railroad track to try to hear the train? Try it with SUPER-EAR. You'll hear that train many miles away. Use it as a powerful stethoscope on yourself, a friend, or a pet. You can even hear a bird's breathing.

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his signal can reach in the station's service area. It is true that horizontal-polarized antennas could give the station better range, but that would be useless to people who have vertically polarized antennas on their receivers.

As far as transmitters and exciters are concerned, any station still using a 25-year-old exciter and transmitter would not survive. You could get by with using a 25-year-old transmitter in AM radio and TV, but not FM broadcast. In fact, most of those old FM transmitters were relegated to auxiliary service, or even scrapped, when FM stereo came about. This is why: Old transmitters used modulator and multiplier stages that had insufficient bandwidth to handle the stereo-modulating signal. Also, the multiplier stages distorted the stereo information. A modern,

solid-state FM exciter is actually a marvelous instrument compared to the primitive exciters. Its solid-state modulator and AFC circuits require little or no adjustments, and are capable of far lower distortion and greater bandwidth. There are only a few or no multiplier stages that don't require tuning. The new exciters are compact and efficient, and not susceptible to microphonics.

The modern transmitter is a far superior performer compared to the older models. The newer and more efficient stages give wider bandwidth, easier tuning, and are much more efficient. They are less likely to cause distortion and harmonics. The modern transmitter is very "transparent" to the exciter's signal. As far as interference is concerned, such problems are very rare, because both transmitters and exciters are well shielded.

Here are some other considerations that broadcasters use in evaluating transmitters: They want something that is very energy-efficient, because the transmitter uses more electricity than any other device that the station uses. Older transmitters are just too inefficient, and every kilowatt the transmitter uses means bigger bucks each year on the power bill. That is getting very important. Also, parts and tubes for 25-year-old transmitters are getting very expensive, and difficult—if not impossible—to obtain, because so many of the manufacturers have long gone out of business. Even parts for equipment 10-15 years old have become hard to obtain. Also, old transmitters were not available in the power levels that broadcasters require now. The use of lower-gain, wide-beamwidth antennas requires a higher-power transmitter.

To sum it up: If Mr. Koontz would listen carefully to an FM radio station using new equipment, he would be surprised at how well it can perform: far better than 25-year-old or even 10-year-old technology. It's just like trying to say that a 20-year-old black-and-white tube-type TV set is better than a solid-state 1982 color receiver.

I can agree with Mr. Koontz on FM tuners. If someone asks me about them, I will reply that spending more than \$500 on a tuner is a waste of money. I laugh at people who spend \$1000 on a tuner and brag about the reception. These days, hi-fi FM listeners are in a minority; modern radio stations try to cater to those listeners who have portable radios. What sounds good on a portable may sound loud, dense, and harsh over your stereo. Whether you like it or not, that's the way it is—I don't agree with it, either.

To tell Mr. Koontz more: The transmitter is actually a minor cause of signal degradation for his station. The most probable cause is the telephone lines that the station may be using to relay program material to a remotely located transmitter. Those telephone lines can have unstable frequency response and distortion. Also, transients, intermodulation, and phase distortion over those lines are a problem. I don't want to downgrade the telephone companies that provide those lines. Most of them try to be cooperative and are sympathetic to the stations' needs; they do try very hard to provide adequate service under tough conditions.

If the station is lucky enough to be using a microwave STL system, the improvement in sound quality can be very startling. Another problem that stations have is distortion in phono and tape systems—those can have more distortion than a typical new transmitter. The problems are the same as those you have with phono and tape machines in your stereo at home.

I can also agree with Mr. Koontz about the Grundig FM receivers: They were high quality for their time. True, the older ones aren't stereo, but they could give excellent results. I have a Grundig radio-phono console that has an AM/FM shortwave tuner that suffered the same fate as some of Mr. Koontz's receivers: bad switches, old capacitors, and old age. The person who gave it to me said that he purchased it in 1959. I am now unable to repair it because the switches are unavailable, along with some of the tubes.

I apologize for the length of this letter, but I had to go to lengths to explain what is going on these days in modern radio stations. I

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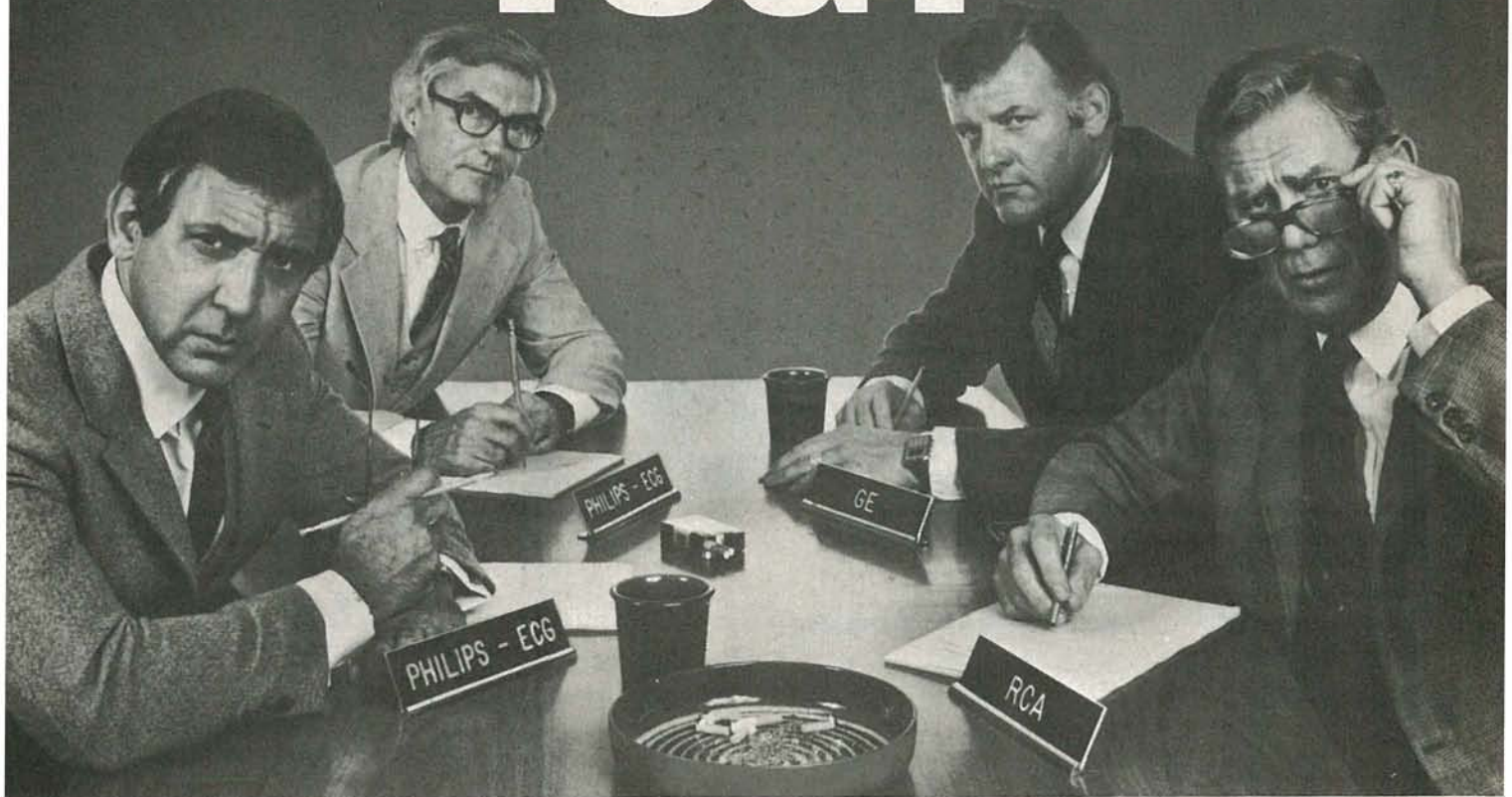
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know because I have worked, and still work, as a Maintenance Engineer for a prominent AM/FM station in the Washington, DC area. I have lived here for close to 10 years.
THOMAS REX OLIVER,
Forestville, MD

COLOR COMPUTER

I believe that Marc Stern made an error when he stated on page 51 of the October 1982 **Radio Electronics** (in the section dealing with Radio Shack's *TRS-80 Color Computer*) that "...the number-crunching capabilities of this system are slowed by its clock speed of .894 MHz."

While the CPU clock speed certainly affects the overall speed of a system, it does not tell the whole story. The *Color Computer*

can hold its own against a Radio Shack *Model III* with Z-80 Processor running at 2.03 MHz. When output to the screen is required during a benchmark test, the *Color Computer* gains almost a 2-1 advantage in speed because of the additional circuitry that it possesses to control the video. A quick glance at the benchmark article in the August 1981 issue of *Interface Age* will convince any reader that the CPU clock speed does not tell the whole story when number-crunching speed is being considered.

Another point: While it is true that a 32 x 16 screen format is certainly restrictive for word-processing tasks, many word-processing programs for the *Color Computer* now come with a 51 x 24 upper/lower case character generator in software.
GOLDEN RICHARD III

CHANGES

The October 1982 issue of *Popular Electronics* announces that its name is to be changed to *Computers & Electronics*. The editorial announcing that seemed proud of the change. I regret it. There are lots of computer publications, and there will be more, but it is becoming increasingly difficult to find major publications dealing with electronics concerns not directly involved with computers. There are readers whose interests who do not revolve around computers.

I hope that a similar change will not come upon **Radio-Electronics**. Your magazine has done a lot to emphasize and clarify basic principles, and for a long time has been very useful in any number of practical ways. It has looked to the future through those means. I hope that you do not allow it to become more specialized.

CARL F. HARTMAN
Newport Beach, CA

As we stated at the end of our October 1982 editorial: "...we do not intend to become a computer magazine. Radio-Electronics remains dedicated to the broad coverage of the entire electronics industry."—Editor

STANDARDS

I would like to comment on Dane E. Erickson's letter in the July 1982 **Radio Electronics**, concerning FCC-mandated standards for teletext, color TV, and other such things.

Among my other gadgets and toys, I own a Sony Betamax VCR, a cassette tape player (several, in fact), the typewriter upon which this letter is being written, and many other things that started out by competing in a free market with devices that did the same job differently.

All of those things have prospered and survived without any government agency mandating any standards at all. In fact, the struggle between VHS and Beta videotape formats goes on as we speak. I buy Beta tapes, while the owner of a VHS machine buys VHS tapes. It is clearly to my advantage to force the VHS people, by law, to change to Beta format so that I can always find the tapes I like in my format.

However, what if VHS turns out to be a better format in the long run? What if, in five years, a new system based on presently unforeseen technology comes out? A mandatory system freezes out such advances. Color TV here, which is seen to be such a success by advocates of mandatory standards, is actually quite inferior to the PAL standards in Europe, because they froze their technology a year or two later than the U.S. did. Who can tell what standards we might enjoy today if there had been no freeze, considering how much difference a couple of years' progress in Europe made in picture quality?

If we are to engage in predictions, let me make some: Teletext will prosper unless the space is needed for other uses (like captioning). AM stereo will be used and will prosper if people care. (I don't believe that it can compete with FM, because of bandwidth limitations.) If those and other ideas fail, it will be because people do not really need them, or want them, and not because the government failed to lead us to them by the nose.

MARWAN E. NUSAIR,
Cincinnati, OH

R-E

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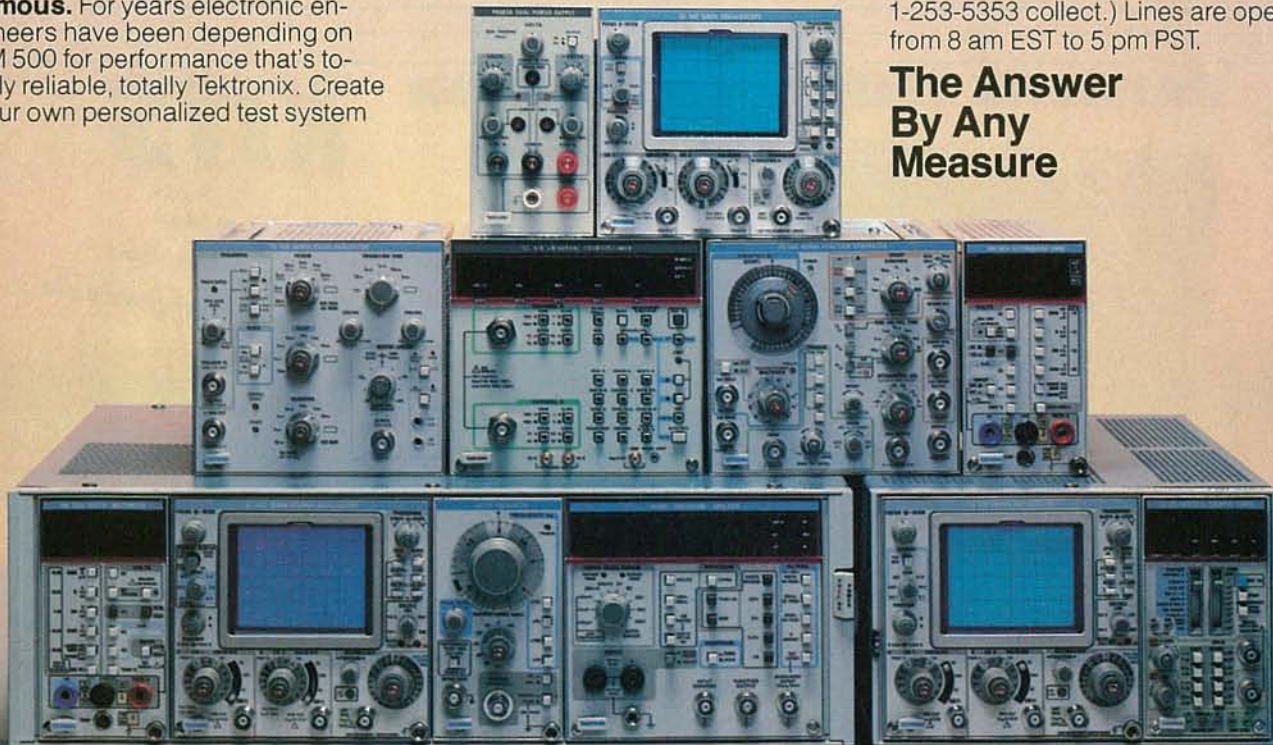
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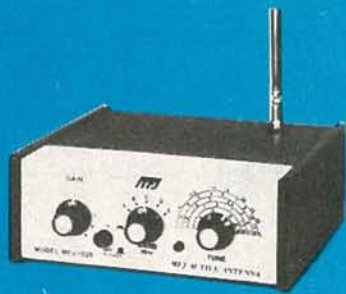
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LIVING IN AN APARTMENT CAN WREAK havoc on a shortwave listener's enjoyment. Usually, there's no place to string a decent antenna because of space and/or

lease restrictions, but some relief is here—at last—in the form of an indoor active antenna. It may not have the gain of a multi-wavelength long wire, tuned

dipole-antenna, or beam system, but it will still give a good account of itself.

Taking advantage of field effect and bipolar transistor technology, MFJ Enterprises (PO Box 494, Mississippi State, MS 39762) has introduced its model MFJ-1020 indoor active antenna; that device covers 300 kHz to 30 MHz in five bands.

Packaged in an attractive wood-grained and cream-colored case, it measures 6¼ × 6 × 2¾ inches and weighs only one pound—but the manufacturer has put a lot of performance into that small unit.

The device's frequency range includes all of the international shortwave bands as well as the high-frequency amateur-radio bands. If you wish, you could also use the device as a preselector for an indoor or outdoor wire antenna.

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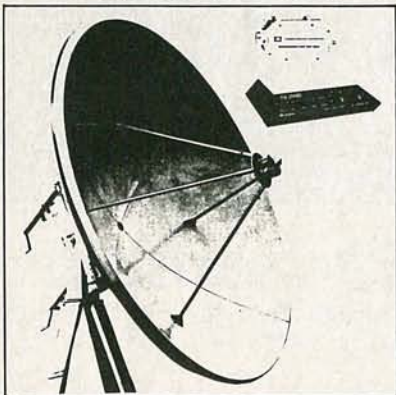
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Setting up the device is also straightforward. Since it is an active antenna, there must be a voltage source—in this case either a 9-volt battery or a 9-volt AC adaptor with a subminiature plug. Also required are a good earth ground and a coax jumper cable between the unit and your receiver. One note about that jumper—both the input and output of the device use phono jacks, not the standard SO-239 coax connectors you might expect. Thus, at least one end of the jumper will have to be fitted with a phono plug. Personally I would have preferred to see SO-239 connectors used here as they are less likely to be accidentally disconnected. MFJ does use them on several of their other products.

Once all of the connections are made, and you've tuned your receiver so that you are receiving a signal, set the antenna's GAIN control to about midrange and adjust the TUNE control for the maximum reading on the receiver's S-meter. If the signal level isn't high enough, you may find that you'll have to adjust the GAIN control to maximum and then adjust it to the proper level. Be careful that you don't leave the device's gain too high, because that could overload the receiver's front end.

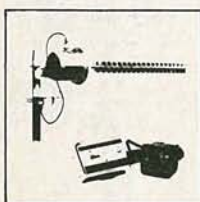
Our test

The first thing we noticed when we opened the carton was how well the unit was packed—certainly well enough to withstand the vigorous rough handling that it is likely to encounter during shipping. Once unpacked, very little time is required to set up the unit—in our case, it was operational less than one hour after opening the carton.

Of course, the important thing is how well it works. In a word, the results were amazing. When used with a receiver with poor sensitivity, the device improved that sensitivity by several orders of magnitude. On another receiver, not only did it help with sensitivity, it also improved selectivity as, with proper tuning, unwanted signals could be notched out. In fact, the active antenna tunes so sharply that the first few times you use the unit you'll find it easy to miss signals. Before long, however, the tuning procedure will become second nature.

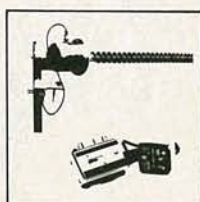
As mentioned earlier, the device can also be used as a preselector for an indoor or outdoor wire antenna. While we didn't use it for that purpose very much, we did

MICROWAVE TV SPECIALS



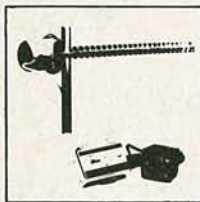
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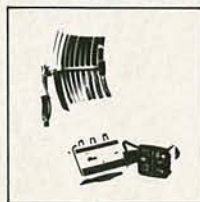
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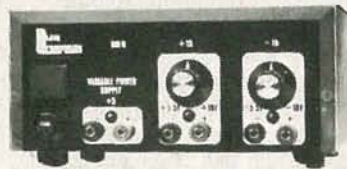


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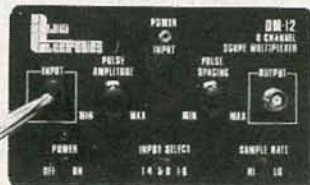


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make it a point to try it out and found that it is indeed useful in bringing a signal out of the noise.

Speaking of noise, one thing that must be remembered is that the unit will not only increase signal strength, but also the noise level in the receiver. That is particularly noticeable on the lower frequencies but, while it can get a bit bothersome at times, it is usually not enough of a problem to interfere with reception. Another and potentially more serious drawback is that the device is sensitive to man-made interference from such things as fluorescent lights and electric motors; of course the same would be true of almost any other antenna system that is located en-

tirely indoors.

The four-page instruction sheet includes all the information a user needs to have the device up and working in as little time as possible. Clearly and concisely written, it includes a schematic of the unit and a PC board component layout. The only problem with those instructions is that it provides little theory—a rather noticeable omission anyone who is the technically inclined.

Overall, however, the MFJ model *MFJ-1020* active antenna works as claimed, and is a good buy at its price of \$79.95. It is especially recommended for anyone who needs, but cannot erect, a full-size outdoor antenna. **R-E**

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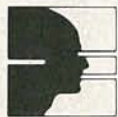
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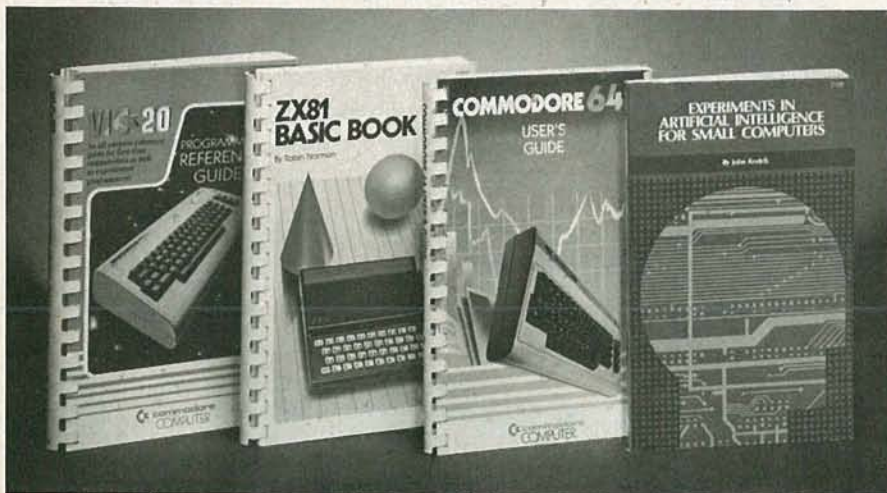
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	Poor		Fair			Good			Excellent	

BOTH ANALOG AND DIGITAL MULTIMETERS have been the workhorse in shops and on the bench for quite some time now. One common complaint associated with analog units, however, has been that the multiple scales make reading the device difficult at times. There have been several attempts to solve that problem, but the most novel approach I've seen is featured on the Sanwa model LCD-900.

That unit appears to be a standard analog multimeter, at least at first glance. It measures DC voltage in seven ranges, from 1- to 1000-volts full scale; AC voltage in five ranges, from 10- to 1000-volts full scale; resistance in four ranges, from 1000-ohms to 1-megohm full scale, and DC current in four ranges, from 300-microamps to 300-milliamps full scale. AC current is measured on one 0- to 3-amp scale. The meter's sensitivity is specified as 50,000-ohms-per-volt DC and 10,000-ohms-per-volt AC.

But one thing does appear amiss, however, when you first open the carton. The meter's face is completely blank. That's because an LCD is used there. Turn the meter on, with the range-selector switched to any range, and only the scale for that range appears. Neat—no more figuring out which range you're supposed to be reading as only one range is visible at one time. The LCD's readability is good, and the bright orange pointer is easy to spot against the display.

Panel markings are nice and clear. The

SWD-1 VIDEO CONVERTER

FOR CABLE TV



The SWD-1 Video Converter is utilized on cable TV systems to remove the KHz's signal from a distorted video (channel 3 in/out) and also pass thru the normal undistorted/detected audio signal. Rocker switch selects operating mode to remove KHz's distortion from the video or pass all other channels normally. Simple to assemble—less than 30 minutes. Pre-tuned. Input/output Channel 3. Impedance 75 ohms. 117VAC.

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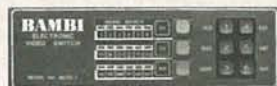
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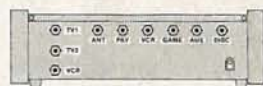
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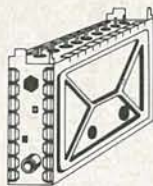
- Input/Output Impedance 75 ohm
- Signal Loss 3dB ±1dB
- Noise 4dB ±1dB
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Antenna Input 75 ohms
Channels 14-83 Output Channel 3



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4	FR35-SW	Resistor Kit, 1/4 Watt, 5% Carbon Film, 32-pieces	4.95
5	PT1-SW	Power Transformer, PRI-117VAC, SEC-24VAC, 250ma	6.95
6	PP2-SW	Panel Mount Potentiometers and Knobs, 1-1KBT and 1-5KAT w/Switch	5.95
7	SS14-SW	IC's 7-pcs, Diodes 4-pcs, Regulators 2-pcs Heat Sink 1-piece	29.95
8	CE9-SW	Electrolytic Capacitor Kit, 9-pieces	5.95
9	CC33-SW	Ceramic Disk Capacitor Kit, 50 W.V., 33-pieces	7.95
10	CT-SW	Variable Ceramic Trimmer Capacitor Kit, 5-65pfd, 6-pieces	5.95
11	L4-SW	Coil Kit, 18mhs 2-pieces, 22µhs 1-piece (prewound inductors) and 1 T37-12 Ferrite Toroid Core with 3 ft. of #26 wire	5.00
12	ICS-SW	I.C. Sockets, Tin Inlay, 8-pin 5-pieces and 14-pin 2-pieces	1.95
13	SR-SW	Speaker, 4x6" Oval and Prepunched Wood Enclosure	14.95
14	MISC-SW	Misc. Parts Kit Includes Hardware, (6/32, 8/32 Nuts & Bolts), Hookup Wire, Ant. Terms, DPDT Ant. Switch, Fuse, Fuseholder, etc.	9.95
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3	3TP11-PWD	PCB Potentiometers 4-20K, 1-5K, 2-10K, 2-5K, 1-1K, and 1-50K (11 pieces)	8.95
4	4FR-31-PWD	Resistor Kit, 1/4W, 5% 29-pcs, 1/4 W 2-pcs	4.95
5	5PT1-PWD	Power Transformer, PRI-117VAC, SEC-24VAC at 500ma	9.95
6	6PP2-PWD	Panel Mount Potentiometers and Knobs, 1-1KBT and 1-5KAT with switch	5.95
7	7SS17-PWD	IC's 7-pcs, Diodes 4-pcs, Regulators 2-pcs Transistors 2-pcs, Heat Sinks 2-pcs	29.95
8	8CE14-PWD	Electrolytic Capacitor Kit, 14-pieces	6.95
9	9CC20-PWD	Ceramic Disk Capacitor Kit, 50 WV, 20-pcs	7.95
10	10CT5-PWD	Variable Ceramic Trimmer Capacitor, 5-65pfd, 5-pieces	4.95
11	11L5-PWD	Coil Kit, 18mhs 3-pcs, 22µhs 1-piece (prewound inductors) and 2 T37-12 Ferrite Toroid cores with 6 ft. #26 wire	6.00
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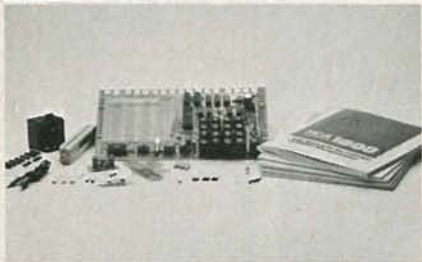
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input jacks here are rather interesting. The negative jack is the same for all of the meter's functions. The positive jack is used for all ranges except the 0- to 3-amp AC range; that range uses a separate input. A nice touch is that the leads need not be unhooked and reversed to reverse the polarity of the readings. The ON/OFF switch has a provision that will switch the polarity of the readings. What's interesting, however, is the jack marked OUTPUT. Long ago, we tuned up old radios by connecting an AC voltmeter to their audio output, using a blocking capacitor to "block off" the high DC voltage often present. That is what is done here; that jack is wired in series with an internal 0.47- μ F, 600-volt capacitor.

The meter requires two batteries for operation—1.5 volts for resistance readings, and 9 volts to power the LCD display. The display is driven by a CD4047AE IC. Battery life is good, although you might run them down just playing with this ingenious little instrument. Seriously, though, the automatic scale-ranging is really quite nice and a pleasure to use.

Soltec (11684 Pendleton Street, Sun Valley, CA 91352) is the exclusive agent for the Sanwa line of analog and digital multimeters. The Sanwa LCD-900 is available from your local distributor and lists for \$139.00.

R-E

converted portable typewriters. One notable exception to that general rule is the TP-1 from Smith Corona (65 Locust Ave., New Canaan, CT 06840). That unit, which carries a price of under \$900, is from start to finish a commercial-quality daisy-wheel printer.

The printer measures 6.4 x 12.4 inches and weighs 18.5 pounds. A hinged plate that serves as the paper support also doubles as a dust cover when it is folded flat. A fan built into the rear of the printer provides cooling.

The power switch and a Centronics-type connector are located on the rear apron. A small control panel on the front has the power lamp and a top-of-form positioning switch; that switch is used for positioning continuous-feed paper. Note, however, that at the present time the printer is unable to handle either tractor- or pin-feed forms. You can use friction-feed forms, but be prepared to monitor the printing closely, as that type of form tends to wander across the platen. In fact, the instruction manual warns against the use of that type of paper. So, in the meantime, the printer appears best for printing single sheets in an office-type situation.

The printer is available with a choice of either 10 or 12 pitch (characters-per-inch), and either parallel or RS-232C serial interface. The pitch and interface must both be chosen at the time of purchase. Printwheels with many of the more popular type-styles are available in both pitch and sizes.

The parallel interface has signal lines for 7 data bits, as well for STROBE, BUSY, and ACKNOWLEDGE. The serial interface has selectable baud rates, ranging from 50 to 19,200 bits-per-second—just about any conceivable baud rate ever used for personal and commercial computer equipment can be selected using jumpers on the interface PC-board. The serial interface can generate and acknowledge the following signals: RT (Request To Send/printer is ready), CTS (Clear To Send/computer is ready to send), and DTS (Data Terminal Ready/busy).

The printer mechanism accommodates paper widths to 13 inches, printing unidirectionally at 12 characters-per-second, which works out to approximately 120 words-per-minute. Just like a typewriter, at the end of each line, the carriage returns to the left for the start of the next line. The line spacing can be set manually for 6, 4.5, or 3 lines-per-inch. There is no provision to allow for program control of the line spacing; once set, it remains that way until changed manually.

Essentially, all the paper handling is done by an uncomplicated typewriter-style mechanism, with the usual end-roller knobs, paper release, paper bail with calibrations, and paper guide. The margins and the tabs, on the other hand, are under full program control, and can be set using PRINT statements from BASIC.

continued on page 38

Smith Corona Model TP-1 Daisy Wheel Printer



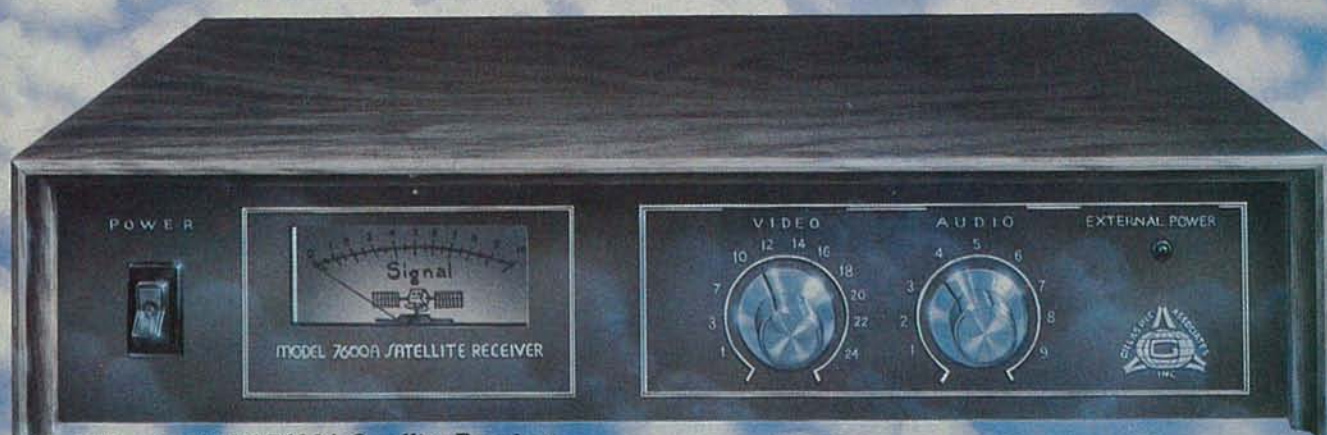
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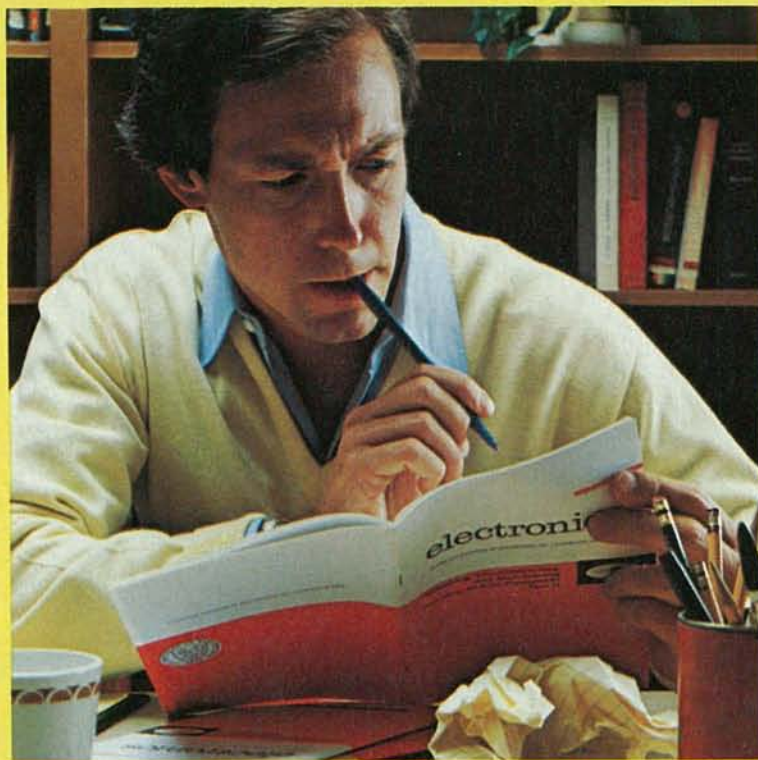
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RE-50

EQUIPMENT REPORTS

continued from page 32

If the word-processing software being used permits inserting ASCII control codes from within the program, the tabs and margins can be varied during the printout. If the software does not permit the insertion of control codes, the margins must be set through BASIC before running the word processor.

The unit can print 85 of the 128 ASCII codes and characters. It ignores the character codes for the caret (up arrow), accent grave, or tilde, and spaces for the "less than" (<) and "greater than" (>) symbols, the reverse slash, left brace, right brace, and broken vertical bar. One serious drawback with the printer is that many of the non-printing characters are used in BASIC program listings, and while the ones that generate a space can be written in by hand later, that is difficult to do for the ones that are totally ignored. That problem limits the printer's practical use to an office-type environment, and indeed that appears to be its intended use.

Of the ASCII control codes, the printer responds to CR (carriage return) and LF (linefeed), BS (backspace), HT (tab), DC1 (left margin set), DC2 (tab set), DC3 (right margin set) DC4 (tab clear), NAK (which prints ¼), SYN (which prints ½), CAN

(margin release), and EM (which turns the automatic underscore on and off).

The carriage return/linefeed arrangement is standard, and works much like most other printers: the machine looks for a carriage return followed by a linefeed. If it doesn't get a linefeed, it will put one in. If it does, the printer will assume that it is the correct linefeed and not insert another one. That will work fine with most computers, the exception being those from Radio Shack. Every Radio Shack printer always provides its own linefeed after a carriage return. Thus, since Radio Shack software is configured to work with the Radio Shack printers, if you use a Radio Shack word-processing program such as *Scripts*, and specify double spacing, the printer will produce only single-spaced copy. To get double-spaced copy, triple-spacing must be specified, and so on. Incidentally, that problem is not just limited to the *TP-1*, but will happen when you use almost any non-Radio-Shack printer with Radio Shack software.

The printer can use either multi-strike fabric or single-strike carbon ribbon cartridges. Both ribbons and printwheels are easily removed and installed because the operating controls are oversize and conveniently positioned. At each turn on, and whenever the cover is lifted, the machine runs a diagnostic procedure that causes the printhead to travel back and forth. If, for some reason, the cover is lifted during

a print, the printer interrupts the computer, the diagnostic runs, and the printhead relocates itself at its last print position so it can pick up correctly when the computer resumes output.

In operation, the printer runs exceptionally well. The mechanism is a little on the noisy side, but the print quality is superb. The characters print on a precise line, and equally well at either side of the paper. While the printer isn't the fastest one around, it's not all that slow either. Besides, anything that will work much faster also is going to cost significantly more.

The only real problem with the machine is the operator's manual. Saying that it leaves a lot to be desired is being kind. For example, you'll need a really good understanding of BASIC to be able to set the margins under program control as there is no description that makes sense. Specifically, the manual says "The printer carriage must be positioned to the desired margin set point...." — there's no way that can be done, either electrically or manually. A print statement filled with spaces directly in front of the margin-set ASCII code must be used to set the margins. The instructions for setting the tab are just as bad. Also, there is no list of how each ASCII code affects the machine. Finally, the ASCII control codes are given only in hex, not decimal, yet most software requires decimal input

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when imbedding control codes. Even BASIC sends the CHR\$ function in decimal. If you get the machine, be sure to dig up a standard ASCII code chart somewhere.

Except for that operator's manual, however, the TP-1 does a notably excellent job for the money. It has none of the bells and whistles of the super daisy-wheel printers, but it also doesn't cost anywhere near as much. If you're looking for something that will turn out a good-looking document and stand up to the rough handling of a typical office, be sure to look at this unit. And, if they ever come out with a retrofit to provide the symbols used in BASIC listings, this will be a dynamite all-purpose printer. **R-E**



"Hit me, Honey, I've got a system."

Non-Linear Systems Model TR-1 Signature Analyzer



CIRCLE 104 ON FREE INFORMATION CARD

NLS	TR-1
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EASE OF USE	
INSTRUCTION MANUAL	
PRICE/VALUE	
	1 2 3 4 5 6 7 8 9 10
	Poor Fair Good Excellent

HAVE YOU EVER WISHED THAT YOU HAD an in-circuit tester that could be used to identify instantly faulty transistors, di-

odes, capacitors, inductors, and the like? Well, new from Non-Linear Systems, Inc. (533 Stevens Ave., Solana, CA 92075) is a compact analyzer called the model TR-1; that handy little unit can do just that.

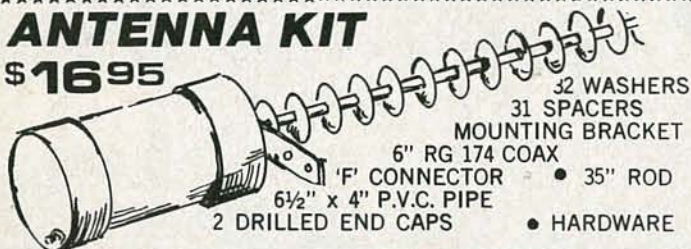
The device works on the principle that if a known signal is fed to a component, the resulting waveform, called a signature, can be used to diagnose a fault in that component. For a good transistor junction, the signature is a slanting line ending in a sharp hook. If, however, the corner of the hook isn't sharp, but rounded or looped, leakage is indicated. Good diodes, on the other hand may show a small loop on the hook if they are slow-recovery types; fast-recovery types, however, must show a sharp hook like a transistor's. Good inductors and capacitors are shown as a loop. For capacitors, that loop slants to the right; for inductors it slants to the left. For all components, a straight line (no loop or hooking) indicates a short or open—shorts show up as a vertical line; opens a slanting line. After a little practice, good and bad components can be spotted almost instantly by analyzing the displayed signature.

The analyzer has two input channels; one is labeled A and the other B. Either one can be selected using a front-panel switch. That two-channel capability is useful for many applications including checking such things as stereo amplifiers.

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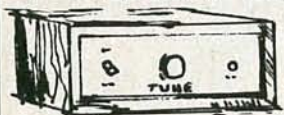
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ELECTRONIC RAINBOW

For instance, a component in the good channel can be compared to the same component in the bad one to locate the source of the problem. Two sets of color-coded test leads are supplied with the unit.

A third position on the selector switch is labelled A/B. When the selector switch is in that position, the input channels are automatically switched at a rate set by the A/B RATE control located on the front panel; that rate can be varied from 0.25 to 0.5 seconds. Synchronization of the patterns is completely automatic; no adjustments are needed. LED indicators are used to show you which channel is being displayed.

The test signal used by the unit is a 2-kHz sine wave. Either a 2-volt or 50-volt amplitude (P-P) can be switch selected. The resulting waveform is displayed on the analyzer's small (1-inch) CRT. The very bright blue-white trace can be centered using the positioning controls that are located conveniently on either side of the CRT. The graticule is calibrated to give approximate resistance readings. Resistances from 0 to 500 ohms can be read in the LO range; resistances from 500 ohms to 500 kilohms can be read in the HI range.

There are many applications where this device should prove to be useful. Among them are checking large numbers of identical parts to weed out the bad ones or cross-checking between a known good part and one suspected to be bad. It could also be useful for testing digital devices where many identical circuits are used. With use, you're sure to find many more applications where this device can save you a considerable amount of troubleshooting time and effort.

In using the model TR-1 we found drift to be negligible and the displayed trace to be rock steady. All-in-all this handy device worked exactly as claimed. Two models of the analyzer are available. One is a line-operated version for bench use and lists for \$795. The other is battery-operated for portable and field use; it lists for \$945 with battery charger. **R-E**



"He keeps reminding me he fixed his radio in 60 seconds flat. Now watch this: Helloph, Batphm, do youth reef me?"

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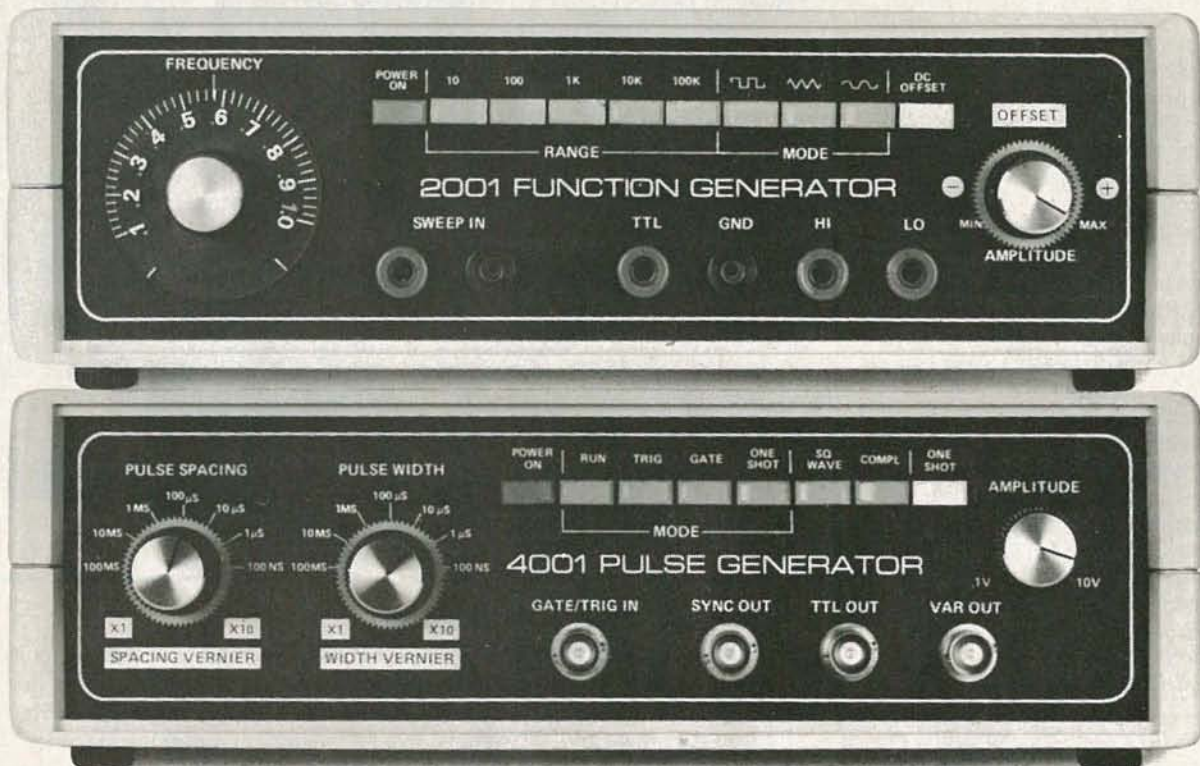
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CIRCLE 10 ON FREE INFORMATION CARD

BUILD THIS AUTOMATIC

COMMERCIAL EDITOR

Are commercials ruining your pleasure when you watch your favorite black-and-white classics on TV? Build this automatic commercial editor and make commercial-free tapes.



FOR YOUR VCR

GARY McCLELLAN

ATTENTION, ALL FANS OF BLACK-AND-white movies! If you like to tape those late-night classics for viewing at a more reasonable hour, you'll love this gadget. It detects color commercials, and stops your VCR (Video Cassette Recorder) while they are running. Then, when the commercials are over and the film begins again, the VCR restarts and continues to record. As a result, all the commercials are automatically edited out of the movie, and you can enjoy it without breaks.

Using the device is simplicity itself. You just connect two cables from it to the recorder's VIDEO OUT and PAUSE jacks. Then you turn it on, and start the recorder. The rest is automatic.

Besides removing commercials from black-and-white movies, this commercial editor offers several fringe benefits. For example, by eliminating commercials, the recorder uses less tape. That means that shorter (and less expensive) tapes can be used to record longer programs. (In my area, a one-hour tape will record almost 1½ hours of movies!) Still other features are a five-minute timeout, and a jack for your remote pause control.

The five-minute timeout feature restarts the recorder if the commercials run over five minutes at a time. That feature is included because some recorder manufacturers recommend that their machines be left in the PAUSE mode for no

more than five minutes. Longer than that can damage the tape, or cause excessive head wear, so the timeout feature protects both your tape and VCR. And, although the editor uses the PAUSE jack on the VCR, you can still use your remote PAUSE control as the editor contains its own jack.

You can build your own commercial editor in just a few evenings. Only four IC's are used, and there are only two easy calibration adjustments to be made. The only test equipment you need is an analog VOM and an RF signal-generator. (Don't be too concerned if you don't have the signal generator—there are ways around that, as you'll see.)

How it works

The automatic commercial editor works by detecting the presence or absence of the *color burst* signal, which accompanies all color broadcasting, but which is not transmitted (usually) when the material is in black-and-white. When that signal appears, a relay in the editor closes and switches the recorder to the PAUSE mode. When the color commercials are over, the color-burst signal disappears, and the relay opens. That switches the VCR to RUN, and recording resumes. When the next commercial appears, the cycle repeats all over again.

Note that the program being recorded must be black-and-white for the editor to

work (otherwise the color-burst signal will be present at all times). If the film is in color, the first five minutes of program or commercials will not be recorded. After that time, because of the timeout feature, both commercials and program will be recorded.

Some readers, no doubt, are wondering whether this editor can be made to work with color programs. The answer is that it is not likely. The reason, we found—after extensive research—is that there is not enough difference between color commercials and color programs, which makes detecting color commercials very hard to do. So forget about an editor for color-program commercials...at least for now.

Now let's take a look at the inside of the editor. Refer first to Fig. 1. There are three main circuit-blocks: an AGC-controlled amplifier, a comparator, and a timer. We'll discuss them one at a time.

The AGC-controlled amplifier comes first. The signal from the recorder's VIDEO OUT jack is supplied to the input of the editor. That video signal consists of horizontal sync, vertical sync, video, and a color-burst signal. The only portion that we are interested in is the color-burst, which is really a few cycles—a burst—of a 3.579545-MHz signal. That signal tells the TV receiver that it is receiving a color transmission, and determines what the

colors are, and how bright they are. To use it, we must amplify the color burst and filter out the rest.

Bandpass filter BPF1 separates the 3.579545-MHz (from here we'll call it 3.58 MHz for short) signal from the rest of the video, and that signal is amplified by IC1. The output from IC1 drives another bandpass filter, BPF2, which peaks the 3.58-MHz signal farther. Finally, the output from that filter is detected by D1 and converted to pulsating DC. The DC biases transistor Q1, which serves as an automatic gain control (AGC). The AGC-voltage controls the gain of IC1 via resistors R4 and R1 and, as a result, the output from D1 is at a constant amplitude whenever a color signal is present.

In our application, however, the color-burst signal itself is not used. Instead, the voltage drop across R4 is detected, which brings us to the second part of the circuit, the comparator.

The comparator circuit used here is rather unusual. To prevent false triggering, several circuit twists have been added. Resistors R6 and R7 provide a 0.26 volt hysteresis, and resistor R5 and capacitor C11 are used to introduce a short time-delay. The result of those additions to the circuit is to prevent noise—which can come from the video—from falsely triggering the device.

A color-burst signal applied to the input of the editor causes a voltage drop across resistor R4. When that voltage exceeds 0.26 volt, and lasts for at least 100 ms, comparator IC2 will be triggered, and its output will go from high to low. It will stay low for at least 100 ms after the color burst disappears, and that brings us to the third circuit block.

The last bit of circuitry is a five-minute timer, built around IC3. Whenever the output of IC2 goes low, both IC3 and the relay receive power. The relay closes immediately and, if the output of IC2 does not go high within 5 minutes, IC3 times out, removing power from the relay. The relay contacts, of course, control the VCR's PAUSE function.

Now that you have a basic understanding of how the project works, let's look at Fig. 2, the schematic diagram.

The AGC-controlled amplifier is at the top of the schematic. The video signal is input through C1. Coil L1, with capacitors C2 and C3, is peaked for the 3.58-MHz color-burst signal. The two capacitors also provide impedance matching to IC1. That IC is a differential input/output IF amplifier. It amplifies the signal by up to 60 dB, and its output appears across C6 and L2. That network is also tuned to 3.58 MHz, and provides greater selectivity. The amplified signal is detected by D1, and the color burst appears across R2 and C8. The DC component of the burst biases Q1, which draws current through R4. Capacitor C9 is included to prevent Q1 from amplifying the AC component of the color-burst signal so only a DC signal

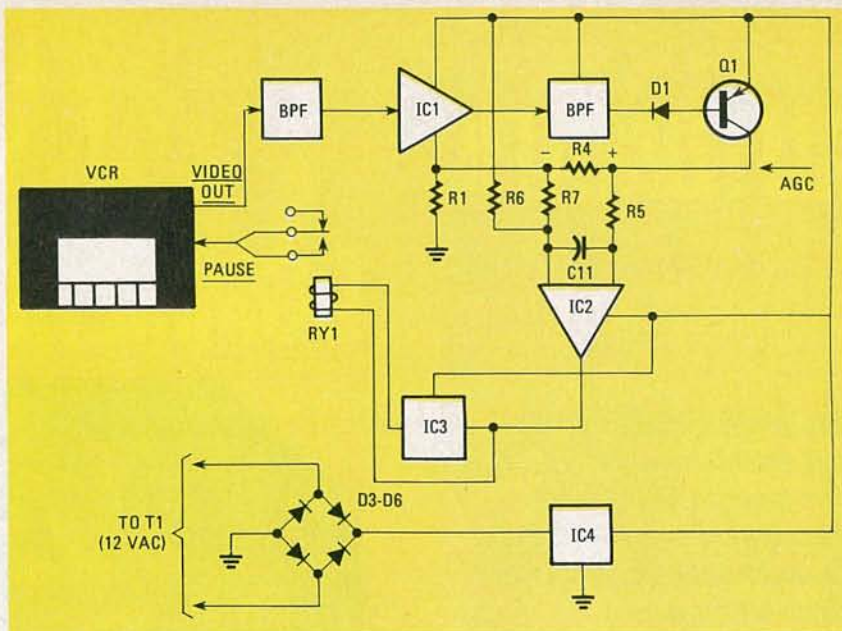


FIG. 1—BLOCK DIAGRAM OF COMMERCIAL EDITOR shows three principal sections: AGC-controlled amplifier, comparator, and timer.

is produced by Q1. From R4, that signal goes to IC1, where it controls the gain of the amplifier. In addition, IC1 also provides a 5-volt DC bias for the comparator; that 5-volt DC bias appears on pin 5 of IC1 and is necessary for the comparator to work properly.

Comparator IC2 monitors the voltage drop across R4. Resistors R6 and R7 form a voltage divider, biasing pin 2 of IC2 0.26-volt higher than the other input (pin 3). Thus, the voltage across R4 must equal or exceed the 0.26-volt hysteresis to trigger IC2. That extra circuitry was added to improve noise immunity.

With no color burst present, the output of IC2 will be high. When a color burst is present, though, there will be a voltage drop across R4. Typically, that will be 0.3 to 1.0 volt, depending upon the color intensity. When the voltage drop exceeds 0.26 volt, the comparator's output will go low. That, in turn, will trigger the timer circuitry.

In addition, the C11/R5 time-delay circuit provides a short time-delay (about 100 ms) before the comparator is triggered. That prevents noise spikes from triggering the comparator.

The output of the comparator drives a five-minute-timer circuit. Whenever pin 7 of IC2 goes low, a ground is provided for the IC3 circuitry, and that, in effect, applies power to the timer. Capacitor C14 and resistor R9 provide a trigger pulse for the timer, which causes relay RY1 to close, and C13 to charge slowly through R8. After about five minutes, the capacitor is sufficiently charged to trigger IC3, which opens RY1. Of course if the commercials last less than five minutes, the output of IC2 goes high sooner. In that case, the timer starts all over the next time it receives power. Note that the relay contacts are connected to the PAUSE jack on the recorder.

Finally, there's a simple 12-volt power supply. A wall-plug transformer supplies about 14-volts AC to D3-D6; from there the voltage is filtered by C15. It is regulated to 12-volts DC by IC4.

PARTS LIST—EDITOR

All resistors 1/4-watt, 5% unless otherwise specified

- R1-R3—10,000 ohms
- R4—1000 ohms
- R5, R6—1 megohm
- R7—48,000 ohms
- R8—10 megohms
- R9—100,000 ohms

Capacitors

- C1—10 pF, ceramic disc
- C2—220 pF, ceramic disc
- C3—470 pF, ceramic disc
- C4, C5, C7, C8, C14, C17, C18—0.01 μ F, ceramic disc
- C6—150 pF
- C9, C10, C12—0.1 μ F, ceramic disc
- C11—0.1 μ F, Mylar
- C13—22 μ F, 16 volts, low-leakage electrolytic

C15—470 μ F, 25 volts, electrolytic

C16—10 μ F, 16 volts, electrolytic

Semiconductors

- IC1—MC1350 IF amplifier
 - IC2—LM311 comparator
 - IC3—555 timer
 - IC4—78L12 12-volt regulator
 - Q1—2N3906 or equivalent
 - D1—1N6263 Schottky signal-diode or 1N60 germanium diode
 - D2-D6—1N4002
 - RY1—12-volt SPDT relay, 400-ohm coil (Mouser 43BC001 or equivalent)
 - L1, L2—10 μ H adjustable shielded coil (J.W. Miller 9052 or equivalent)
 - P1—RCA phono plug
 - P2—subminiature phone plug
 - J1—subminiature phone jack
 - S1—SPST toggle switch
 - T1—12 VAC, 250 mA, wall-plug transformer
- Miscellaneous: PC board, IC sockets, RG-174 coaxial cable, case, wire, solder, etc.

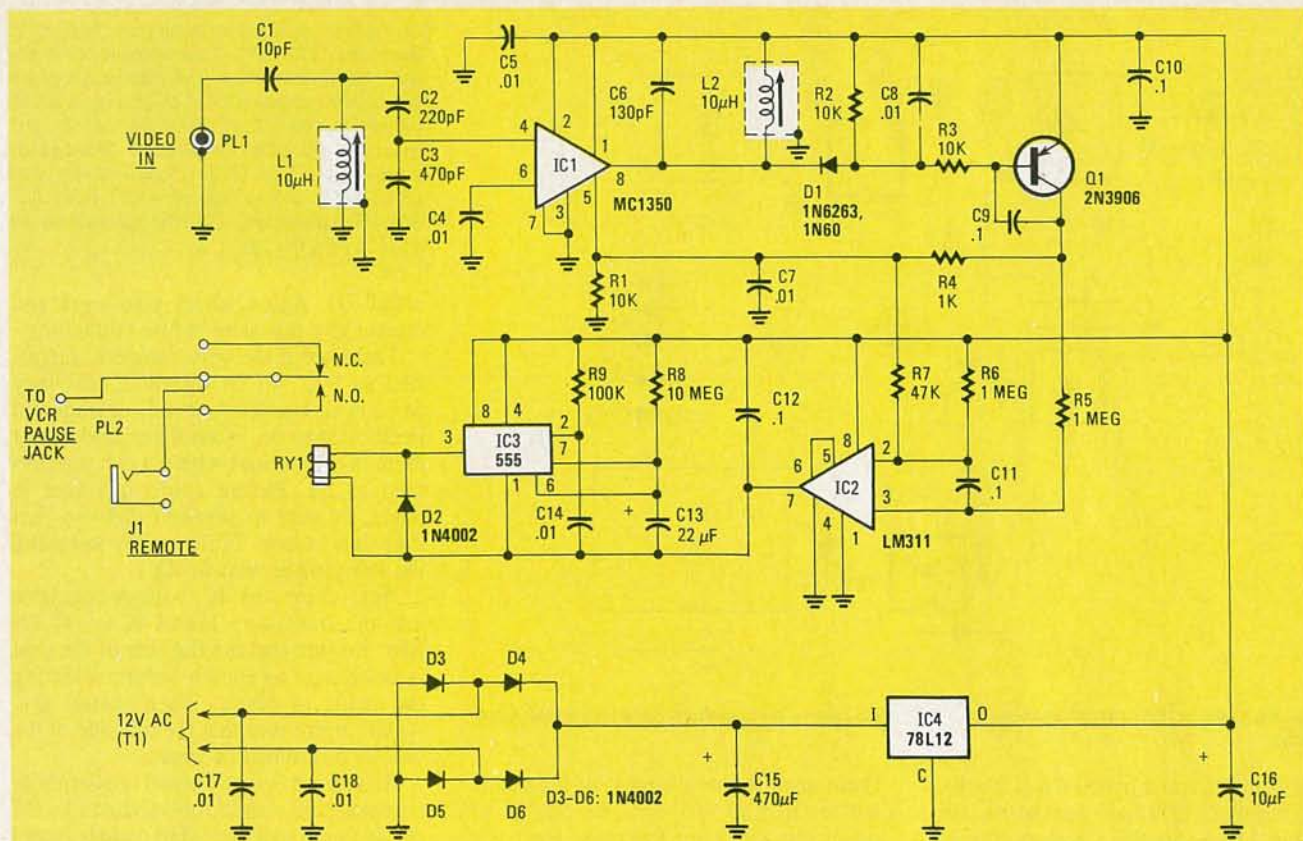


FIG. 2—AGC-CONTROLLED AMPLIFIER is preceded by two bandpass filters that isolate 3.579545-MHz color-burst signal.

Assembly

The commercial editor is intended to be built on a small PC board. In fact, you *must* use a PC board. If you don't, the AGC-controlled amplifier will probably oscillate, making it impossible to adjust the unit.

Figure 3 shows the foil pattern for the single-sided PC board, and Fig. 4 the parts-placement diagram. Start by positioning the board foil-side down so it faces the same way as the board shown in Fig. 4.

Start assembly with the IC sockets. Install 8-pin sockets at IC1, IC2, and IC3 as shown, and carefully solder them in place.

The coils and the relay are installed next. Install a coil at L1, and push it flush with the board; then solder it in place. Similarly, install L2. Finish up by installing relay RY1 in the appropriate position. Be sure to push it flush with the board before soldering.

The resistors are installed next. Start by installing R1 (10K) next to L2. Then install another 10K resistor at R2, and still another 10K resistor at R3 as shown. Move over to the IC2 socket and install R4 (1K) near it. (Be careful not to put it in the R5 position!) Then install R5 (1 megohm) between R4 and the IC2 socket. Move to the right of the IC2 socket and install R7 (47K). Again, make sure it's in the right place. After that, install R6 (1 megohm) next to R7. Move down to the

IC3 socket for the remaining resistors. Install R8 (10 megohms) just to the left of the socket. Then move down and install R9 (100K). That finishes up the resistors. Check your work carefully, and correct any mistakes you may find.

Install the capacitors next. Start by mounting C1 (10 pF) next to L1. Be sure to push the part flush with the board be-

fore soldering its leads in place. Then install C2 (220 pF) next to L1 in the same manner. Similarly, install C5 (0.01 μF) as shown. After that, install C3 (470 pF) between L1 and the IC1 socket. Move to the left of the IC1 socket and install 0.01 μF capacitors at C4 and C7.

Continue by installing C10 (0.1-μF ceramic disc) next to L2. Move to the

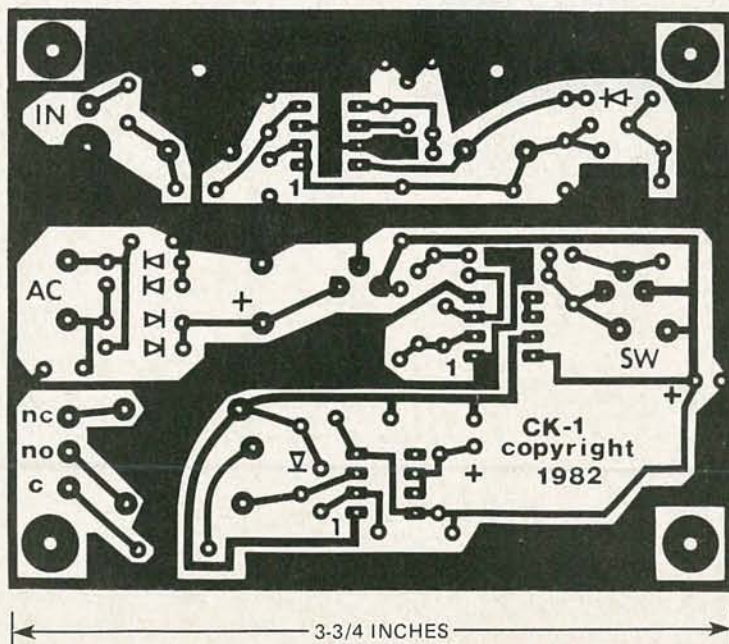


FIG. 3—FULL-SIZE foil pattern for single-sided editor PC board.

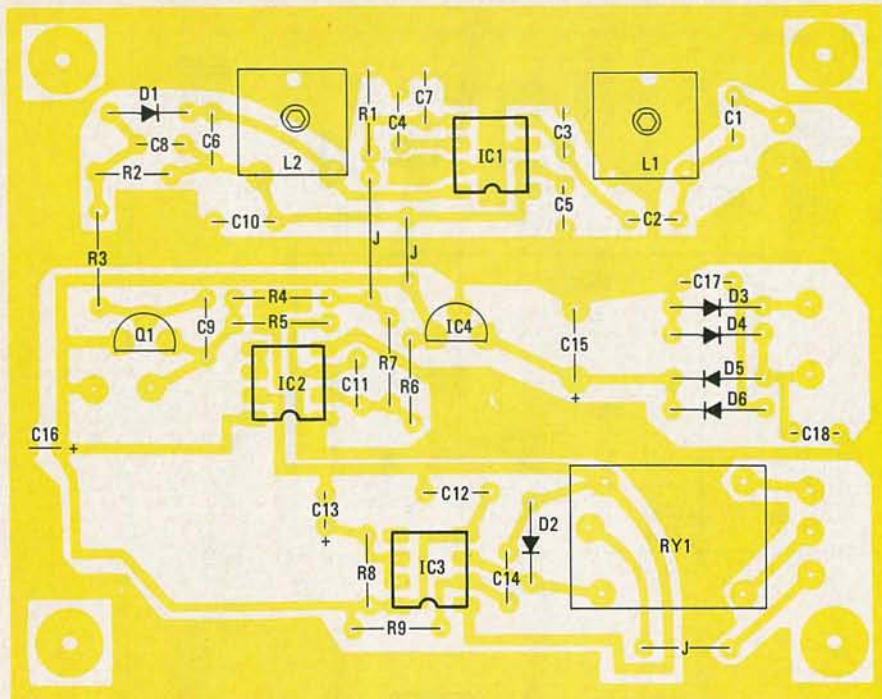


FIG. 4—BE SURE TO OBSERVE polarities of critical components. Be especially careful to install relay correctly.

other side of L2 and install C6 (150 pF). Then install C8 (0.01 μ F) next to the 10K resistor. Make sure that C8 goes into the correct holes. After that, install C9 (0.1- μ F ceramic disc) between the resistors.

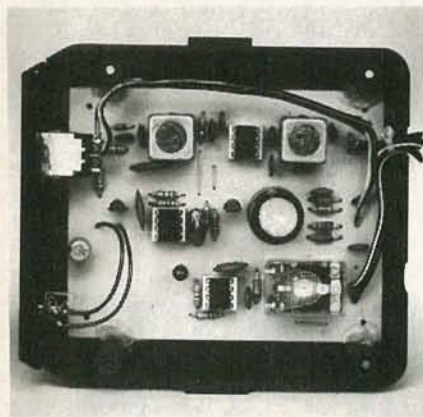


FIG. 5—COMPLETED circuit board should look like this. Board can be installed in small plastic clock-case.

Then move to the left edge of the board and install C16 (10 μ F). Be careful to install this capacitor the right way; it's polarized. Similarly, install C13 (22 μ F) next to the 10-megohm resistor. Then move up and install C11 (0.1- μ F Mylar) between the IC2 socket and the 47K resistor. Continue by installing C12 (0.1 μ F) above the IC3 socket and, after that, install C14 (0.01 μ F) to the right of the IC3 socket. Move to the right center of the board and install C15 (470 μ F) as shown. Be sure that this polarized part is inserted properly, then push it flush with the board and solder the leads in place. Finish up by installing C17 and C18 (0.01 μ F). Carefully check your work before going any further. Be sure to check the polarities of the three electrolytics, and correct any mistakes before continuing.

Install the diodes next. Start with D3-D6. Install the four diodes as shown, then solder them in place. Move over to RY1, and install D2 as indicated. Finally, move to the top left corner of the board and

The following are available from Technico Services, 1920 W. Commonwealth Ave., Box 20HC, Fullerton, CA 92633: kit of all parts, less case, (CK-2), \$54.00 plus \$3.50 postage & handling (nonrefundable); PC board only (CK-1), \$10.00, postpaid; assembled and tested calibration tool (3.58-MHz signal generator) (CAL-3), \$10.00, postpaid. California residents please add 6% tax.

install D1. Again, check your work and correct any mistakes before continuing.

There are three wire jumpers, identified as "J," to be installed, and they come next. Use leftover resistor leads and bend them to fit, insert them, and solder them in place. Start with the two jumpers next to L2. Before soldering them in place, be sure to position them so that they don't touch. Finish up by installing the last jumper next to RY1.

Next come the IC voltage-regulator and the transistor. Install IC4 (78L12) first. Be sure that the flat side of the case is positioned as shown before soldering the leads in place. Then install Q1. Again, make sure that the flat side of the case is positioned as shown.

As you can see, the board is essentially complete and should look almost like the one in Fig. 5. All that is left to do is install the IC's and attach the cables.

Start with the IC's and install IC1 (MC1350) first. Be sure to position it properly before you plug it in. Then install IC2 (LM311) in the same manner and, finally, IC3 (NE555). Be sure to check the IC's for proper installation before continuing.

Finish up by installing the cables (refer to Fig. 6). Start with the video-input cable. Cut a two-foot length of RG-174 coaxial cable, then strip and tin each end. Attach PL1 (RCA phono plug) to one end. Then connect the other end to the board. Note that the shield goes to the heavier ground foil on the board.

After that, connect the transformer leads. Measure about four inches from the end of the cable that is attached to T1 and cut just one of the wires at this point. Connect switch S1 at that point. Solder the free ends of the T1 cable to the appropriate pads on the circuit board.

Next, cut a two-foot piece of two-conductor wire (a short length of speaker wire will do). Attach PL2 (subminiature phone plug) to one end, and connect the other end of the cable to the board as shown.

Finish up the cable installation by wiring J1. Cut two two-inch lengths of hookup wire and solder them to J1 (subminiature phone jack). Then connect the other ends to the pads on the board.

That completes assembly of the editor's PC board. Once you've carefully checked your work, the next step is to align the device. That will be just one of the topics we'll cover in the next part of this article.

R-E

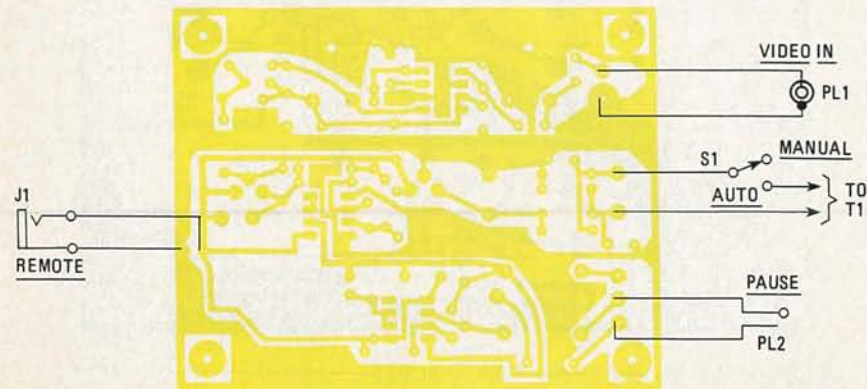


FIG. 6—OFF-THE-BOARD connections. Use RG-174 coax for VIDEO IN cable.



ELECTRONICS HOBBYISTS HAVE NEVER had it as good as they do today. The modern home workshop can now be equipped with laboratory-grade instruments at consumer prices and can produce projects of industrial quality. It's rather astounding to consider that the decision whether to build or buy is more a function of convenience than complexity. Ever since the IC manufacturers discovered the profits to be made in the consumer market, the components needed for virtually any project are rarely more than a postage stamp away.

So why haven't we seen a tremendous increase in the number of sophisticated projects built and developed on a backroom bench? Obviously there's been a problem, but the biggest stumbling block to home construction doesn't have anything to do with design or complexity—nine times out of ten it's the printed-circuit board. A hobbyist who wouldn't think twice about designing a circuit for bouncing signals off Neptune often cringes when he considers producing printed-circuit boards. While electronics technology gets more sophisticated almost by the minute, the process of making PC boards has remained just about the same for the last thirty years. And the reason why it hasn't changed is that the process is really quite simple.

No kidding, it's really easy to do. There's been a big push among manufacturers to develop wire wrapping as some sort of technology all on its own. If you're into wirewrapping, don't bother

ETCH YOUR OWN PC BOARDS

There are lots of ways to make printed-circuit boards, but if the pattern is complicated, nothing beats doing it photographically.

ROBERT GROSSBLATT

reading any farther because all I think wire wrapping is good for is baling hay. But making even the most complicated double-sided board isn't difficult—it takes time and requires attention to detail, but it's not hard. Let's take a look at the basic procedure.

Copper-clad board is coated with a light-sensitive chemical and allowed to dry. A mask is made photographically (the foil pattern) and contact-printed on the board. The light in the printing process causes a change in the chemical coating so that the unmasked areas become resistant to the developer. What happens then when the board is placed in the developer is that the chemical over the unmasked areas (the traces) remains, while the chemical over the rest of the board is removed. The result is a board ready to be dumped into the etchant. That is the basic procedure for so-called negative photoresist; there's also a positive resist but the idea is the same. As you can see, there are several steps in the process and we'll go over them completely later on in this article. By the time we finish, however, you'll be wondering why you thought it was difficult in the first place.

Layout

Before we explore the mysteries of board making, there's something that comes first—laying out the pattern. By the time you're ready to consider making PC boards you've probably pulled out a lot of hair in breadboarding. You've rearranged circuitry and substituted differ-

ent values to make sure everything is working exactly the way you want it. When you reach that point—when you decide that you're looking at the final design—you're ready to move on.

The first step is to draw a schematic of your breadboarded circuit. The chances are that you've already done that, but if you're anything like me the drawings are a mess. When I'm at that point, I usually have several pages filled with false starts and things that blew up. That's to say nothing about the most important sections, which I usually scribble on the back of an envelope. It's really important to stop and redraw everything. Make a clean drawing of the circuit, lay it out in a coherent fashion, assign numbers to the components, and carefully label everything on the page. Also, make sure you include the pin numbers of the IC's you are using. Use graph paper to help keep the lines straight and do as many versions as you need in order to make the drawing neat. Keep on checking your drawing against the breadboard as mistakes are easy to correct at this stage—it's much simpler to erase a line than to correct a trace. Make a couple of copies of the finished drawing in case you spill coffee on the original.

A completed drawing is a valuable tool because it will give you your first overview of your entire circuit. Don't be surprised if you wind up combining things or simplifying the design. Unused gates can be combined to eliminate whole IC's and a little reorganization can allow you to simplify complicated sections of your design. I usually wind up with several stages of "final" drawings until I get the point where any further changes are profitless. When you've reached that stage you are ready to begin the layout of the board.

At this point you have the schematic and the breadboarded circuit. **Whatever you do, don't fool around with the breadboarded circuit!** Not only will you still be checking the drawing against it, but later on, when the printed-circuit board is complete but perhaps doesn't work, you'll have something to compare it to. Remember an all-important rule of original design: the breadboarded circuit is the only one in the world that does what you designed it to do. The last step in creating a working circuit on a PC board, *the very last step*, is to pull the breadboard apart.

Special considerations

Examine the breadboard and the drawing. Are there sections of your design that have to receive special treatment? If certain sections need shielding, or have unusual power requirements, they should be physically close together on the board. Should the circuit be broken into several boards to conserve space? Is the printed-circuit board going to provide structural strength to the finished project? Does the board have to have a particular shape be-

cause of the case? Those are the sorts of questions you have to consider when you begin to do the layout. The answers will determine the materials you use, the size and number of the boards, how compact the layout has to be, and so on.

Before you start planning the printed-circuit pattern make sure that you have all the components that go on the board. Resistors, capacitors, and other components come in different shapes and sizes—and you're doing a layout for components of a particular size. If you plan on making three boards, make sure you already have three of everything you'll need. There are few things sadder than going through the trouble to make a complicated board and then finding out that certain small electrolytics aren't available any longer. It's always best to stick with standard, readily available parts.

The best way to work out your layout is to do it on graph paper, but the choice of

paper is important. What you want is a light-blue grid measuring ten squares to the inch. Get a large pad of at least 20 by 24 inches and tell the people in the art supply store that you want it to be non-reproducible blue. That is a particular shade of blue that won't register on the film you'll use to photograph the final drawings. The pencils you use for the initial drawing should be the same shade of blue and can also be obtained at an art supply store. The reason you want such a large pad is that you'll be doing the drawing double size and reducing it later on. Leave yourself lots of room to work and make sure that the paper you use is at least four times larger than the intended size of your board.

Drafting

Draw a horizontal line across the middle of the graph paper. You should be able to fit a double-size outline of your board's

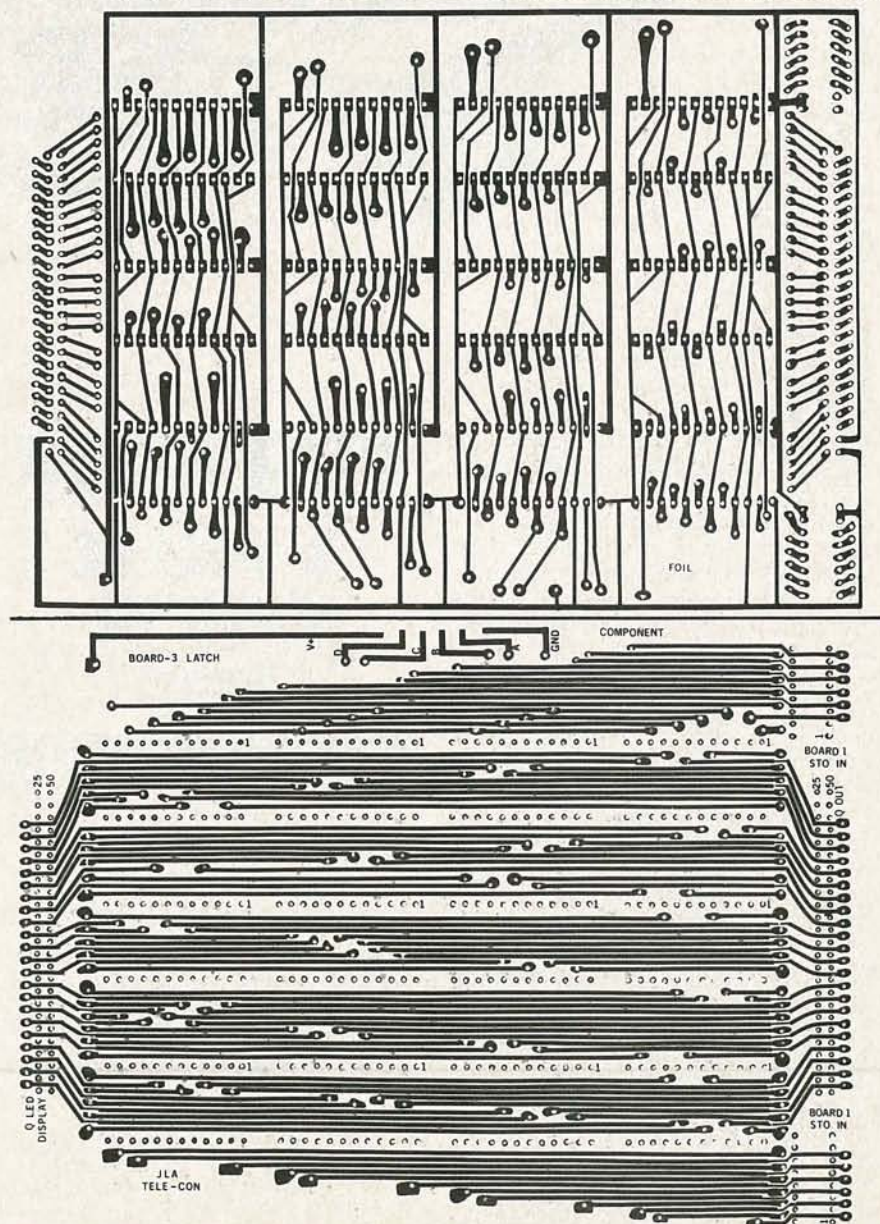


FIG. 1—DIGITAL-CIRCUIT FOIL PATTERNS often require long unbroken traces and double-sided boards with lots of interconnections. But even the most complicated patterns can be laid out and etched if you are patient and careful.

shape on both the lower and upper halves of the paper. If you can't, get larger paper. The reason for dividing it in half is that you'll be doing the foil pattern and the component layout at the same time. Not only that, but you'll be able to tell whether your board has to be single or double sided. Draw the outlines of the board in black using a fine-point magic marker. All the rest of your preliminary work will be done with blue pencils. This is the time to draw in the board extensions for edge connectors, or any other odd shape you need. Anything that has to be in a particular place on the board should be done in black. These are the "givens" around which the rest of the components have to fit. They are usually due to particular case requirements.

There are lots of schemes for doing a layout. Probably the best way to start is to draw the pins for the IC's. A good way to start the layout is to look at how the IC's were placed on your breadboard. Remember that you're working twice actual size so adjoining pins should be two squares apart. Since the physical dimensions of IC's are designed around 0.1-inch spacing, you'll find that everything fits neatly onto the paper. Place the IC's in neat rows and columns. That will make things easier when you start drawing in the traces and will make redundant connections obvious. Common connections such as power, ground, clocks, and so on will be much easier to lay in if the IC's are logically spread out on the paper. Make sure to indicate the IC pins on both halves of the graph paper and label pin number 1 on both halves to help keep everything straight in your mind.

Remember that you're not reversing anything—the bottom and top halves of the paper are being drawn as if you had laid both sides out flat and were looking down at them. Some layout schemes call for drawing the two sides of the board on top of each other in different colored inks.

Not only does that cause extra steps later on, but left and right get confused and it's really easy to make mistakes.

I wish I could give you some foolproof way of routing the leads but I don't know any other than trial and error. Be prepared to do a lot of erasing and redrawing because it takes a while to solve the topological problems caused by your circuit—there's no teacher other than experience. When you find that you absolutely, positively have to cross traces, break one of them and draw doughnuts at the ends. Locate the **exact same** positions on the other side of the board, draw two more doughnuts there, and draw a trace connecting them.

As you draw the connections for your circuit and put in the components, try to avoid using the component side of the board because single-sided boards are much easier to make. Unfortunately, however, digital circuitry often requires a double-sided board. If you accept that in the beginning, you can save a lot of grief by first laying out the address and data lines. The chances are that those traces will require a number of connections between both sides of the board. It's not at all unusual to have a trace jump back and forth from one side of the board to the other eight or ten times—if you can make the connection that way, do it.

The complexity of the foil pattern you generate will depend on a number of things. If the component density is high, the foil pattern will be busy. Obviously, the more connections you have to make, the more complicated the foil pattern will be. Interestingly enough, if your circuit has lots of passive components, you'll have an easier time figuring out a foil pattern. You'll wind up with lots of short traces and you'll be able to route the longer power and ground connections around them. Digital circuitry is tough because lots of direct connections have to be made between the IC's. That requires

long, unbroken traces and, unfortunately, lots of connections between both sides of the board (see Fig. 1).

Double-sided boards

Sometimes you can simplify things by rearranging the IC's but if you're dealing with power, ground, clock, address, and data lines, there's no way you're going to be able to get them all on a single-sided board. Don't throw in the towel, however, as double-sided boards *aren't* that much harder to make. Piggybacking components and using loads of jumpers is a poor alternative to making a double-sided board.

You have to be really careful when you're working back and forth on the two sides of the drawing—make sure that the holes and traces line up properly. Use a straightedge and calipers, as shown in Fig. 2, and don't feel silly about counting squares—I've made a lot of double-sided boards and I do it all the time.

Once you've decided to make your board double sided, take advantage of the decision. There's as much copper on the component side of the board as there is on the foil side. It's rather strange to jam as many traces as you can on the foil side and leave only a few on the component side. Not only that, but the more evenly you distribute your traces, the more likely it is that the board will etch evenly. If you have 90% of your pattern on the foil side and just a few traces on the component side, you're running a serious risk of undercutting the traces on the foil side during etching. What's more, having lots of traces on the foil side means that some of them are going to be thin. While you're waiting for a veritable sea of copper to be etched on the component side, the thin traces on the foil side will disappear.

If you plan things carefully, the connection from one side of the board to the other can be made using the leads of the components. It's a simple matter to solder the leads on both sides of the board. For those traces that wander from one side of the board to the other, you'll just have to leave doughnuts and connect them with small pieces of hookup wire soldered to both sides.

The big problem comes when you have to deal with IC's. There are a few options: For one thing, you can solder the IC's directly to the board and make the feed-through connections by soldering the IC legs on both sides—but **don't do it**—always use IC sockets. You can mount the socket so that it is raised about an eighth of an inch or so above the board and solder the the socket legs on both sides of the board; of course, you'll need wire-wrap sockets to do that. I don't like that method because wire-wrap sockets are more expensive and physically larger than others. Also, having the socket sit above the board leaves the plastic part of the socket free to move and puts an undue strain on the legs. All things considered,

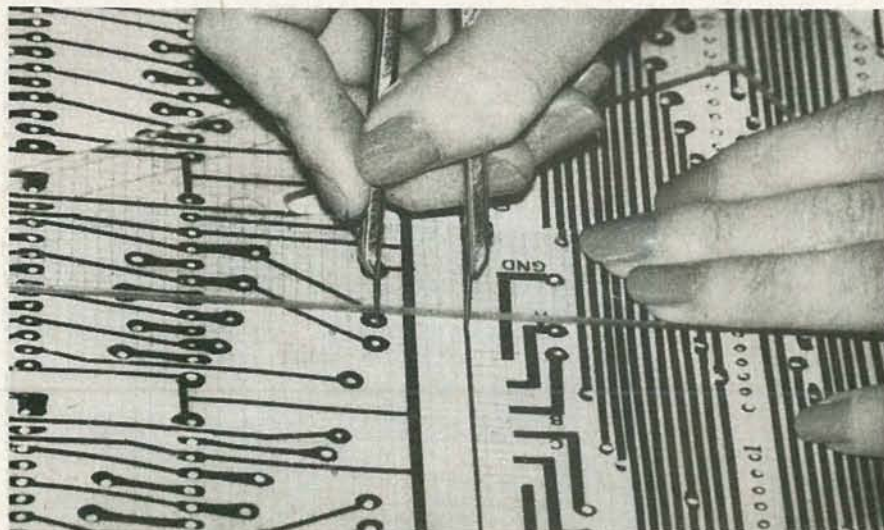


FIG. 2—WHEN LAYING OUT a double-sided pattern, one of the most important steps is to make sure that both sides line up properly.

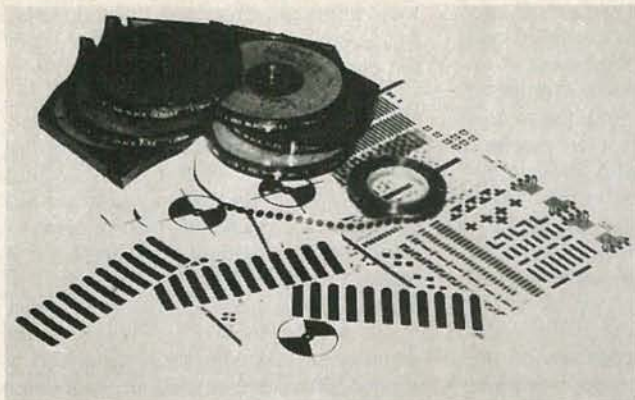


FIG. 3—DRAFTING AIDS, such as the ones shown here, can be very useful in completing your layout.

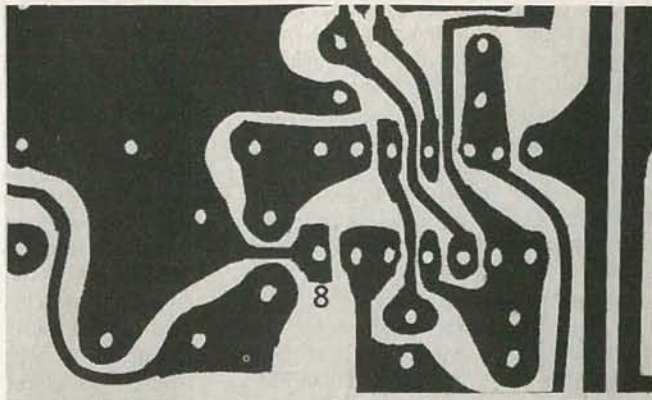


FIG. 4—IF TRACES MUST BE ROUTED between IC pads, elongate the pads to make soldering the components easier later on.

if your circuit is complex enough to require a double-sided board, you want to eliminate any potential problems, and mounting sockets above the board is a possible source of trouble.

What I do is to extend the trace on the component side and use small doughnuts on both sides of the board to make the connection from one side to the other. It's neater, and makes the board more secure and compact. After the board is etched, I thread hook-up wire through all the feedthroughs and then solder and snip them in one shot. Any other method leaves room for error and if you have a board with lots of connections, it's much too easy to forget the IC legs hidden under the sockets and then go crazy trying to figure out why the circuit won't work.

Further considerations

Whenever you add a component to your drawing, show the doughnuts on both sides of the board. Not only does that make the geography of the board easier to understand, but it will be the basis for a parts-placement diagram later on. When you have your drawing completed, draw the components in (remember you're working twice actual size) and make sure that there's enough room to fit them on the board. Count up the number of traces on the component side to see whether a double-sided board is justified. If there aren't too many, consider the idea of jumpers, instead. If you feel, though, that a double-sided board is necessary, make sure that you leave at least two squares between the traces and any component leads. That is particularly important around the IC's because you don't want the IC sockets to hide the traces on the component side. It's almost impossible to locate a small break in a trace if it's hidden under a socket, and it's even harder to correct it.

Once you have the drawing finished and have checked the hole alignment, make sure that the connections agree with the schematic. Take your time doing that because corrections are a lot easier to make at this stage than they will be later on.

Final artwork preparation

After everything has been verified, the next step is to prepare the board for photography. All that means is that you have to ink in all the doughnuts and traces you've done in blue pencil. Many of the commercially available drafting aids can come in handy here (see Fig. 3). Long runs can be made using black drafting tape and several companies sell IC pads and doughnuts in double-size patterns. When working double size, try to keep the traces at least $\frac{1}{16}$ -inch wide and separated by at least $\frac{1}{32}$ -inch. That's the absolute minimum, though—tiny traces have a nasty habit of disappearing in the etchant and small separations may not show up in the final foil pattern.

The only exception is when you find it necessary to route leads through the pins of IC's (which you should avoid like the plague—especially if you're doing a double-sided board). If you find that you absolutely have to route traces between IC pads, make the pads, rather than the trace, thinner, and elongate the pads parallel to the traces as shown in Fig. 4 to be sure you'll have enough copper to solder the pins to.

When you've finished that part of the layout make sure there aren't any breaks in the pattern. If you find any, color them in with a fine-point felt-tipped pen as shown in Fig. 5. Check the pattern carefully and if any of the traces or corners

look thin, use the pen to fatten them. If you make a mistake, cover it over with typewriter correction-fluid and redraw it.

Make sure your lines are as black as possible and are completely filled in—every bit of black you add now will become copper on the board. Also, try to anticipate any special problems. Will you be soldering any nuts on the board to mount a meter or power transistor? If so, blacken in an area large enough to provide copper for the nut. There are no hard and fast rules to tell you how much copper to leave on the board and everything depends on your application, and your patience. Just remember that the more copper you keep, the less you have to etch. Once I've gotten to this stage, I thicken every trace I can to make the etching process as simple as possible. It never hurts to have spare copper on the board and wide power and ground planes are never a bad idea.

There is one place where the doughnuts should be as large as you can possibly make them. That is wherever you have feedthroughs going from one side of the board to the other. The larger the doughnuts, the more tolerance you're going to have in keeping the top and bottom of the board in register later on.

When we continue this article, we'll put the finishing touches on the layout. We'll also show you how to transfer the layout to the copper board. **R-E**

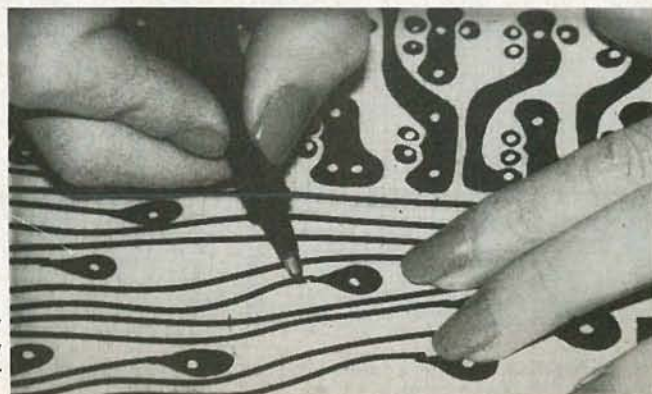


FIG. 5—REPAIR ANY BREAKS in the layout by filling them in with a felt-tipped pen.

SPECIAL SECTION

VIDEO Radio- Electronics® ENTERTAINMENT

A GUIDE TO VIDEO ENTERTAINMENT IN THE HOME

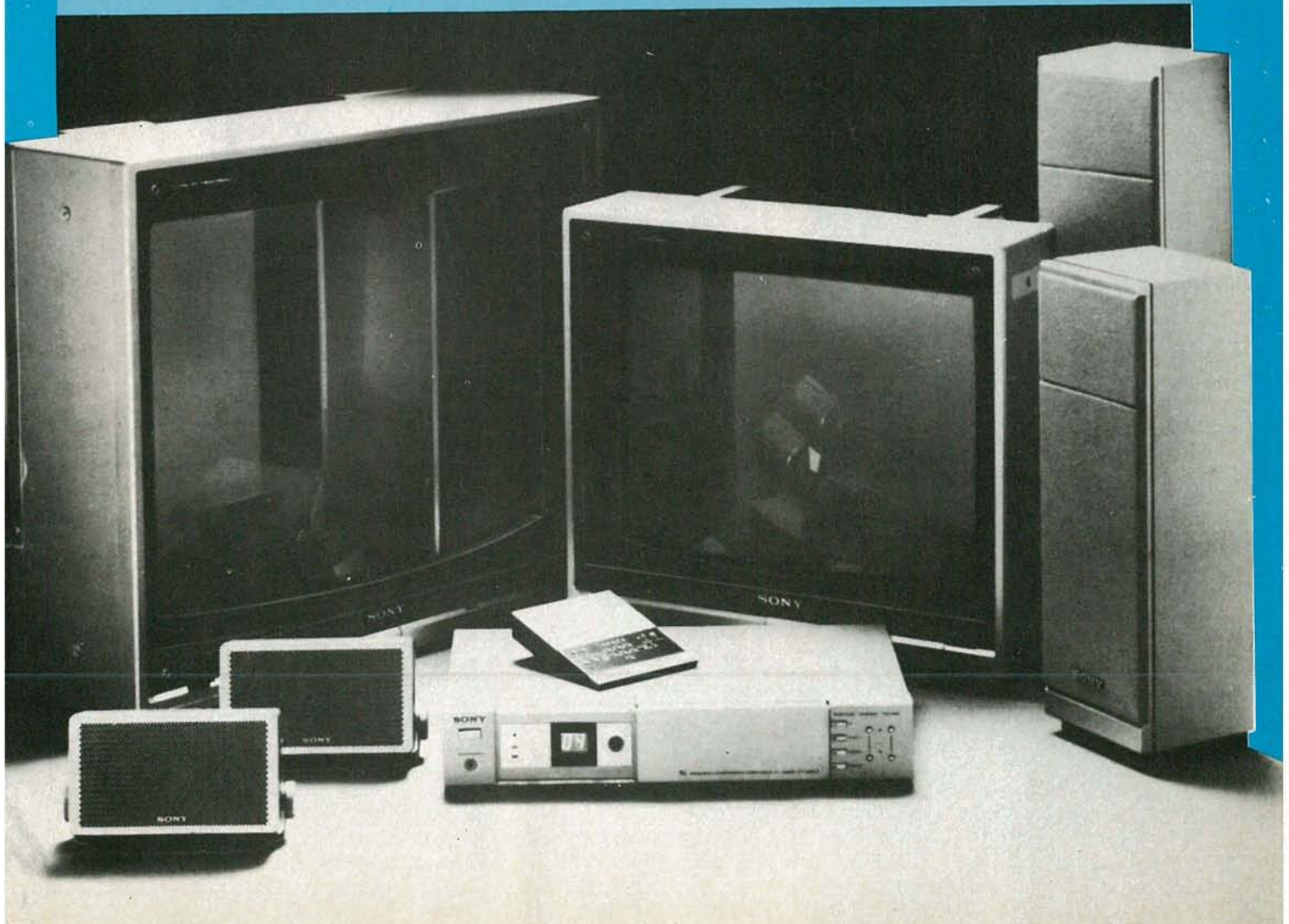
Video Entertainment in The Home

New Video Components

Direct Broadcast Satellite Television

Video Accessories

How to Connect Video Components



VIDEO ENTERTAINMENT

The video revolution has changed the way we look at and use television. But has your set kept up, and will it keep up in the future? The following will help answer that question.

A. LEVIS

IT WAS ALL SO SIMPLE BACK IN THOSE EARLY DAYS OF VIDEO, BEFORE SONY'S *Betamax* and Atari's *Pong* began to change the way we viewed the lowly TV set. In that ancient era, which began drawing to a close in the early 1970's, video simply meant television. The only choice one faced when shopping for video equipment was whether or not to blow the entire budget on that huge 25-inch console.

Unless you've been asleep these past few years, you know that those days are most certainly gone for good. Walk into any video store or department and you are likely to see VCR's (VideoCassette Recorders), videodisc players, projection televisions, television-component systems, VCR and cable-TV accessories, and perhaps even a complete satellite-TV earth station. You're also likely to find a great many computers and video games; we'll not say much about those products in this section, but you can be certain that the next year will feature the kind of new products and exciting advancements we've all now come to expect. What we will be looking at is video—the latest developments in the field, as well as what the near future holds. But first, let's look at what our old family TV-set has grown into, and why.

Alternative programming

In the old days, television "quality" generally referred to a set's reliability rather than to the kind of image it reproduced. After all, few consumers would have been willing to lay out the extra dollars for such things as high resolution—even if the manufacturers had been capable of producing sets with such features—just to watch the generally low-quality fare offered by the networks.

That began to change with the advent of cable-TV services such as Home Box Office. That service, and others like it, offered viewers a real alternative in the form of recent movies, live sports events, and specials.

The real beginning of the so-called video revolution, however, began with the introduction of the VCR. That device gave the viewer almost complete programming control. Today, programming can be drawn from a vast array of pre-recorded videotapes, including the very latest movies (sometimes the videotape is released simultaneously with the movie's premiere); from both free (broadcast) and pay-TV services, or can be original—

the video camera allows your video programming to be limited only by your imagination. For those who demand the very best, there are now video accessories that help clear up many picture, color, and stability problems; we'll take a closer look at those elsewhere in this section.

A relatively new program source is the videodisc. Although those devices lack the recording capabilities of the VCR, and have not proven as successful in the marketplace as was predicted, there is still considerable hope that it will succeed as an alternative program source. Among its advantages are its lower initial outlay, as compared with a VCR, and its potential for use with a computer.



IN THE HOME

Currently, satellite-TV reception is out of reach to all but a few. That too is about to change as an exciting new service, direct broadcast satellites, will be in place by the end of this decade. Aside from being yet another source of alternative programming, it may also offer such attractive features as high-definition TV and stereo sound.

A new kind of TV

Although TV set manufacturers were slow in responding to the explosive growth of alternative programming and video-related products, they soon realized that their product was destined to become the centerpiece of a sophisticated, highly complex, and demanding electronic home information and entertainment system. They also realized that the conventional TV-receiver was inadequate for that purpose. There were many reasons for that, but among the most important was the way a signal is fed into the receiver. Up to recently, the only signal that the TV was required to handle was a combined over-the-air audio/video RF signal; that

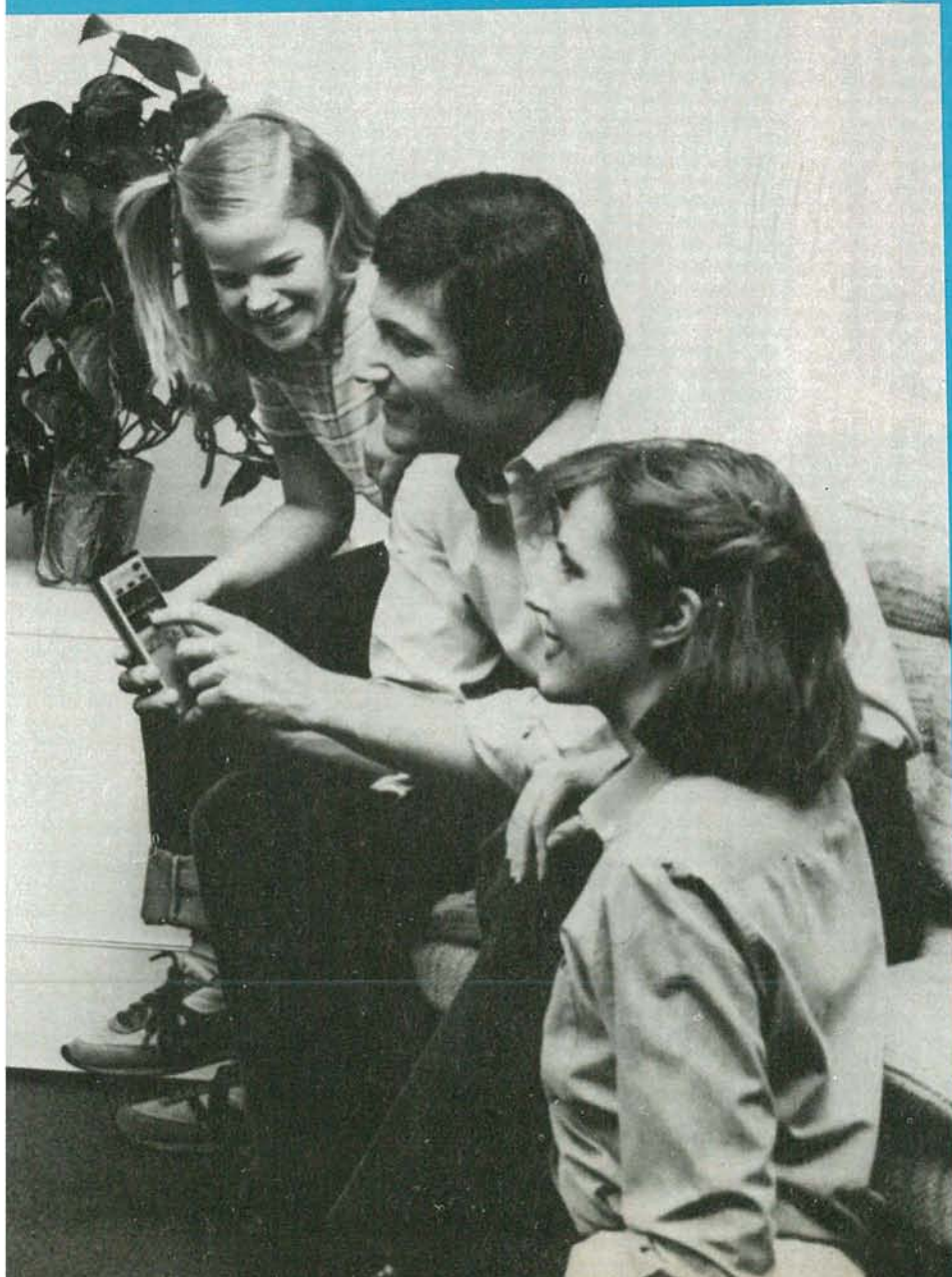
signal was simply fed into the set through the antenna terminals. Once inside, the RF signal was demodulated, with the video information extracted and displayed on the screen. The result of that process was degradation of the signal that showed up as poor resolution, video noise, and the like.

With some recent exceptions, that's exactly how it's still done. But now, as we pointed out before, the TV set is being called upon to handle the audio and video information from a variety of sources. Those signals are still fed in through the set's antenna terminals, and on an RF carrier. As a result, that same modulation, demodulation, and extraction process still causes significant signal degradation. And what may have been acceptable for watching your favorite soap opera certainly is not when dealing with computer-generated graphics, videotapes, videodiscs, etc.

The standard television set also fails in another important area. The audio quality of most sets is poor. And while high-quality audio may not be particularly important for getting the most out of your video game or computer, some things that should happen in the near future will make TV audio more important than it has ever been before. One of those is multichannel, or stereo, TV-audio. Such audio has been available in Japan for many years now, and last year was introduced in West Germany. In both cases, the set manufacturers were prepared and had sets on the market by the time the service was approved. In this country, three TV-stereo systems have been proposed; as this is written, though, no commitment has been made to any of them, so neither we, nor the set manufacturers can be certain of which standard(s) will eventually receive acceptance. In the meantime, however, VCR's and videodisc players with stereo-audio capability are already on the market.

But help is on the way, thanks to a new generation of TV equipment designed to meet the high requirements of new and future video technology, and to be the centerpiece of that home-entertainment/information center of the future. The age of component television is here.

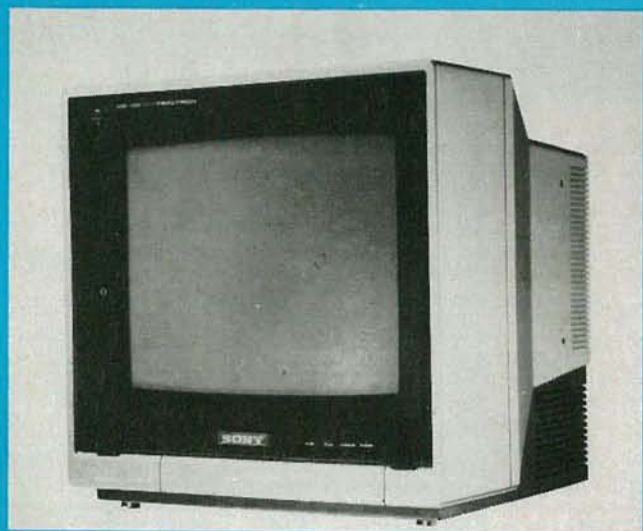
The term component television is currently used to refer to a wide range of products. In the past year, many of the leading set manufacturers, and some manufacturers that previously had dealt with audio products only, have introduced at least a few models designed to solve some of the video and audio problems we've mentioned. In some cases, they've simply taken an existing TV chassis and added separate audio and video input jacks. Doing that allows direct access to both video and audio circuits, getting rid of many of the problems associated with feeding an RF signal to the set. That helps, but you're still stuck with a receiver with limited flexibility, and limited growth potential.



A better alternative is the true component approach. Systems of that type were first available last year from Sony (1 Sony Plaza, Park Ridge, NJ 07656) and Teknika (1633 Broadway, New York, NY 10019); now they are available from a variety of manufacturers. Those systems are by no means identical, and each manufacturer has its own idea as to how a component system should be configured. For instance, some offer a separate amplifier and monitor while others offer a combined package. Also, in some the tuner doubles as a switcher while in others those functions are handled by separate components. Since you now have a bit to choose from when picking your component system, it's important to match your needs to what's available.

What's available

The *Profeel* system from Sony remains unchanged from last year except for the addition of a new 12-inch monitor to the series. Like the larger 19- and 25-inch *Profeel* monitors, this one boasts good video resolution and offers an RGB (Red, Green, Blue) input. While currently that input will only be of interest to some computer hobbyists, it is hoped that will change with the eventual availability of RGB outputs on



THE NEWEST ADDITION to the *Profeel* line, is the model KX-121HG shown here. Like all *Profeel* monitors, it boasts RGB inputs.

consumer video equipment. The main advantage of an RGB input is that it allows you to bypass all of the signal processing stages in the receiver. The RGB signals are fed directly to CRT grid amplifiers. For the time being, however, such outputs are available only on professional equipment.

Sanyo (1200 W. Artesia Blvd., Compton, CA 90220) enters the field with their *Pro-Ponent* system. It features a 19-inch monitor with a built-in 5-watts-per-channel stereo amplifier, a separate TV-tuner with remote control, a companion stereo-audio sound system, and a pair of accessory speakers. The tuner has multiple inputs designed to accept a wide variety of sources (VCR, videodisc, etc.).

Zenith (1000 Milwaukee Ave., Glenview, IL 60025) was the first U.S. manufacturer to offer a component TV system. It includes a 19-inch monitor, TV tuner, a separate source-selector, a separate stereo-amplifier, and speakers. The monitor also has its own built-in audio amplifier and speaker and features both composite-video/audio and RGB inputs. The source selector allows the connection of up to six sources to the system. The tuner features an infra-red remote control and has 112-channel capability, including 42 for cable TV.

Several high-fidelity audio-component manufacturers have also come out with systems. Pioneer (200 West Grand Ave., Montvale, NJ 07645) is offering two screen sizes, 19- and 25-inches, with the larger monitor offering 400-line resolution, which is the highest claimed by any manufactur-



THIS NEW VIDEO-COMPONENT SYSTEM from Jensen offers a choice of tuners either with (shown) or without AM and FM-stereo.

er. The tuner is housed separately and features 127-channel capability and an infra-red remote control; the tuner also serves as the system's source selector, handling all of the switching functions.

Jensen (4136 N. United Parkway, Schiller Park, IL 60176) has a new entry that offers the consumer an additional choice. While many manufacturers are marketing systems with more than one monitor size, Jensen is marketing two tuners—one for AM and FM-stereo as well as TV, and one with TV capability only.

Even some smaller companies are getting into the act. NAD (675 Canton St., Norwood, MA 02062) and Proton (1431 Ocean Ave., Suite B, Santa Monica, CA 90401) are marketing systems built around the same 19-inch monitor, but will sell them through different outlets.

Component-TV systems are also available from Mitsubishi (7045 N. Ridgeway Ave., Lincolnwood, IL 60645), JVC (41 Slater Drive, Elmwood Park, NJ 07407), and others. And you can be sure that the list will grow considerably in the near future.

The future of video

Will your new video-entertainment/information system soon become obsolete? The answer to that is yes and no. No, because the U.S. is almost certainly tied to the 525-line NTSC broadcast standard for many years to come, although just about everyone involved with the technical end of the television industry—from broadcasters to set-makers—agrees that it is less than perfect. (For one thing, the resolution of current sets is not limited by technology, but by that standard.) Yes, because there are several developments that may soon result in higher definition television, with signals compatible with the NTSC standard.

Work is continuing on high-definition TV (HDTV), with much of it concentrating on developing a system that is NTSC compatible. RCA, for one, is involved in an intensive NTSC-compatible HDTV project, and believes such a system can be operational within this decade.

Another thing to look forward to is the arrival of digital TV. IIT in Europe is scheduled to begin offering set makers digital-circuit kits sometime next year, and such international companies as RCA, Sanyo, Philips of Holland, Sharp, Blaupunkt and Thomson CSF are involved in intensive research and development projects in that area.

But one of the chief advantages of a component system is its adaptability. It's relatively easy to modify your system to keep in step with the latest changes in technology, or your needs. So, don't let what's coming tomorrow spoil your fun today.

R-E

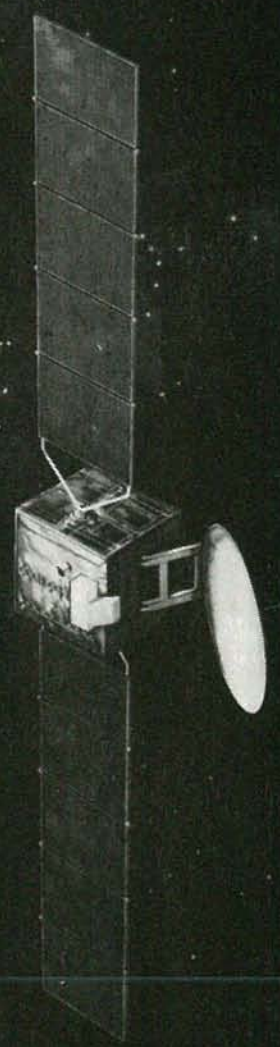
DIRECT BROADCAST SATELLITE TELEVISION

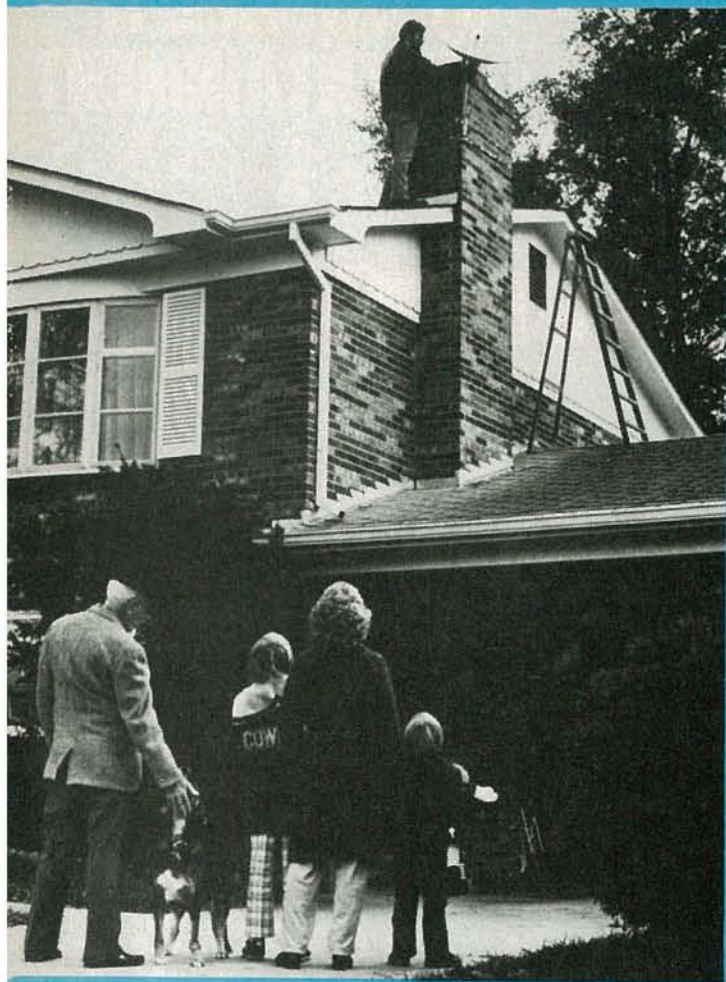
In this age of cable television, a proposed cableless system is getting quite a bit of attention. Here's a look at what DBS television is, and what it can mean to you.

DANNY GOODMAN

AS THE WIRING OF AMERICA CONTINUES, CABLE TELEVISION is reaching more and more densely populated areas. Despite that, however, one industry report projects that as many as 48 million households will not be wired for cable TV by 1990. What's more, the U.S. government estimated that in 1980, 4.6-million homes could receive only one or two channels, while over one million homes were located in areas served by no television whatsoever.

But by the mid-1980's we'll see the beginning of a new type of television service—that service will have a potential of about 30 channels of programming and will be available in all parts of this country, whether it be the crowded island of Manhattan or the most sparsely pop-





THE SMALL SIZE and relatively low cost of a DBS dish antenna will bring satellite-TV reception within reach of most families.

ulated areas of Nevada or Montana. All you'll need to receive that service is line-of-sight exposure to a satellite parked in orbit some 22,300 miles up, a 2½-foot-diameter dish antenna, a satellite receiver, and your TV set.

At least eight companies have unveiled a variety of plans for Direct Broadcast Satellite (DBS) services, including advertising-supported programming (free to the viewer), subscription television (pay TV), and advanced video services such as teletext and *High Definition TeleVision* (HDTV). The success of those plans depends on a number of technical and regulatory issues that are as yet unresolved, but substantial amounts of money are being invested in the belief that DBS services will eventually become a reality.

Mini dishes

DBS services will differ greatly from the satellite-TV services currently in operation. The present generation of TV satellites generally receive high-powered signals from Earth in the 5.9- to 6.4-GHz range (called the *uplink* frequency), and retransmit them in the 3.7- to 4.2-GHz range (called the *downlink* frequency) using low-powered transmitters. The downlink transmitters are low powered primarily because each satellite carries as many as 24 transponders and each of those must share the limited power provided by the satellite's solar panels. The transponders carry TV and radio relays, as well as two-way communications traffic.

The retransmitted signal is so weak by the time it travels through the atmosphere that the receiving antenna dish needs to be relatively large (10 to 20 feet in diameter) to collect as much signal as possible and focus it on the receiver's sensitive amplifier, which is suspended over the dish's center.

Large dishes and highly advanced receiving circuitry built in small quantities make for expensive setups, ranging from under \$5,000 to over \$25,000 depending on your location and the number of different satellites you want to tune in. While that high cost, and the large size antennas required, may not present a problem to a cable-TV company, it does put home satellite reception out of the reach of many individuals and families.

The proposed DBS services, however, will use frequencies in the 17-GHz region for the uplink and frequencies in the 12-GHz region for the downlink. More important, however, is that the satellites will have transmitters capable of outputs of 150 watts or more. Contrast that to the satellites that are currently in service; those can only deliver outputs of about 5 watts or so. The greater power levels will allow a sensitive receiver to gather a useable signal from the smaller 2½- to 3-foot dish. At that size, a chimney-mounted satellite dish is no more cumbersome than a high-gain terrestrial-TV antenna.

Programming on the DBS channels may not be different from present satellite fare, but its intended audience will be. Currently, satellites are used primarily as a relay for broadcast and cable networks. In the original plan, retransmitted signals were to be picked up only by properly licensed ground stations for further distribution over a local terrestrial TV station or cable network. While the FCC has removed the necessity of securing a license for a TVRO (*TV Receive Only*) station, there are still some gray areas about the legality and/or ethics of individuals "eavesdropping" on relay satellites. Home Box Office, one of the more popular and outspoken satellite services has begun scrambling its signal, making it difficult for satellite aficionados to view their programs. Other satellite-distributed services may feel likewise threatened by the relatively small number of TVRO owners and scramble signals.

In contrast, DBS signals will be received by each individual subscriber, bypassing the need for an earth-based distribution system such as a cable system. Some cable systems may, however, wish to "subscribe" to a DBS service so that they can, in turn, offer the DBS services to their own subscribers.

Variety of DBS services

The proposals submitted to the FCC by the initial eight DBS applicants contained a number of different approaches. Satellite Television Corporation (STC), a subsidiary of COMSAT, was the first to submit a proposal, having done so in April 1981. Their plan (see Fig. 1) calls for a system of four active satellites, one for each of the U.S. time zones (the Pacific Time Zone satellite will also serve Hawaii and the more heavily populated areas of southern Alaska), plus two orbiting spares. The spares are intended to provide uninterrupted service in the event that one of the primary satellites fails. As early as 1985, STC hopes to have its first satellite (and one spare) in place, providing service initially to the Eastern Time Zone.

It is expected that each STC satellite will have three operating transponders (one for each channel of programming) and three spare transponders. With the high cost of each satellite and the long waiting list for room aboard a Space Shuttle or Ariane rocket, spare transponders and satellites are particularly important to prevent delays or interruptions in service due to equipment failure.

Three distinctly different program channels will be beamed from the 185-watt transponders. Only one, the so-called "Superstar" channel will run nearly 24 hours a day. Planned programming includes recent movies, popular concerts, theater, and family entertainment. The "Spectrum" channel will offer 15 hours daily of children's programming, classic films, performing arts, cultural programs, and public affairs. "Viewers Choice" will offer 15 hours of sports, educational/instructional programming, ex-

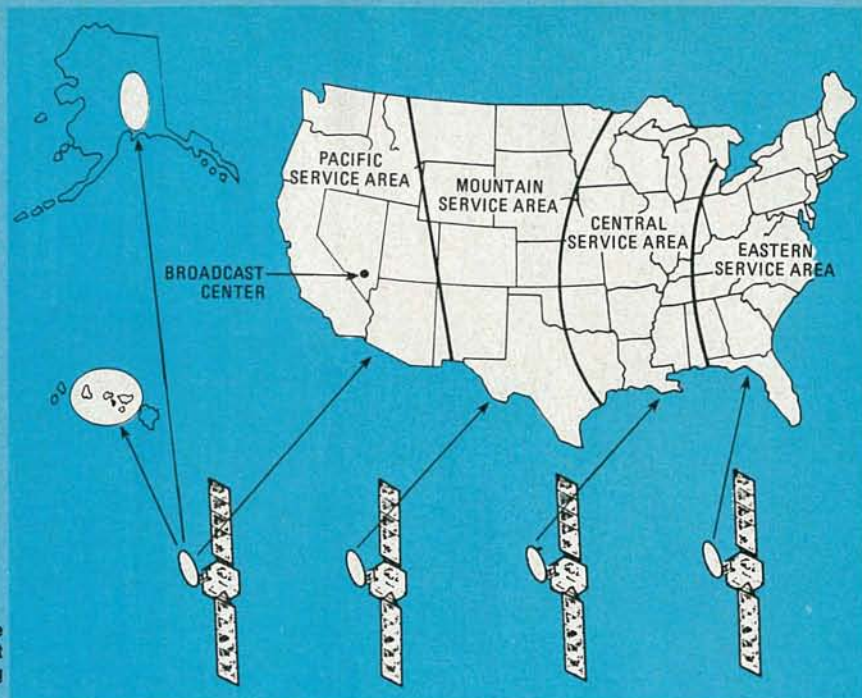


FIG. 1—STC'S PROPOSED DBS SYSTEM will cover the entire continental United States, as well as the most heavily populated regions of Alaska and Hawaii, using four operational satellites and two spares.

perimental theater, and the like. STC plans to "counter program," a scheduling method that tries to satisfy as many different viewing tastes during a given viewing period as possible. The STC satellite network will also have provisions for stereo sound for concerts, simultaneous second language transmission for minority viewers, and closed captioning for the hearing impaired. It also will present a full range of teletext information services.

All three channels will be scrambled, and TV set decoders will be individually addressable (digitally coded signals corresponding to each home decoder's serial number will engage the decoder), allowing the possibility of pay-per-view or pay-per-series specials requiring an extra subscriber fee.

Free satellite TV

United States Satellite Broadcasting (USSB) is taking a different approach. Their initial plan calls for only two satellites to cover the four time zones. Each satellite would carry

six operating transponders, three for each time zone. An orbiting spare would have 12 transponders ready to fill in where needed.

But what is more striking about their proposal is that the three channels would be advertiser-supported—like our present commercial TV—and would therefore be free to any viewer equipped with a dish and satellite receiver. According to Stanley Hubbard, president of USSB, the company's DBS service will be like a fourth national commercial network, with member TV stations around the country. Local stations will act as news bureaus and will be able to sell original programming to USSB and advertisers for national broadcast. Local member stations, whether broadcast or cable TV, will be able to re-transmit the USSB programming in those areas where DBS dishes aren't practical or popular. That proposed system is shown in Fig. 2.

Planned programming on USSB's channels include a full-service TV channel (news, weather, entertainment, children's programming, etc.), a 24-hour news and information

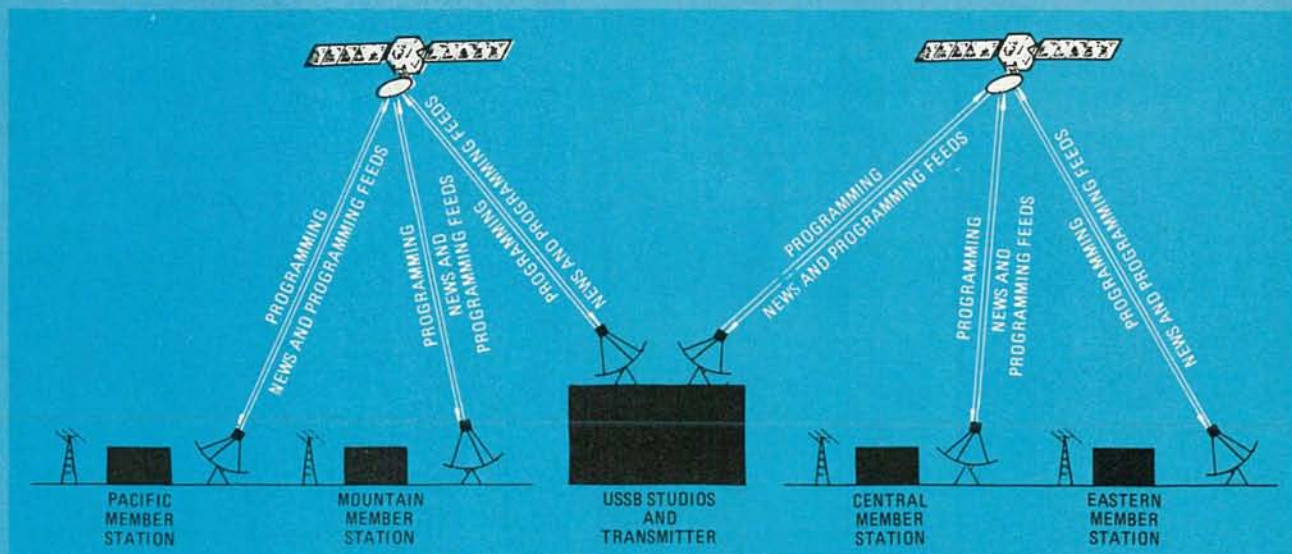


FIG. 2—IN THE USSB PROPOSAL, local member stations will feed news and original programming to the satellite "network" for national distribution. The local stations will also be able to rebroadcast the DBS services in areas where home satellite reception is either not practical or unpopular.

channel, and a third channel. Hubbard would not reveal what programming was planned for that third channel, but he did state that it will be "something original and unique."

Another way DBS satellites may be used is as common carriers. In other words, satellite companies will build, launch, and control the DBS satellites, but will lease the transponders to programmers or networks as is currently done with C-band satellites.

Pat McDougal of Direct Broadcast Satellite Corporation (DBS Corp.) foresees his company dividing the country into three zones, with one satellite covering each zone. Each satellite will have six channels, some of which could be used to transmit programming only to specific regions within the satellite's broader reception area; that reception area is often called a satellite's "footprint."

Although, as a common carrier, DBS Corp. would have no say in the programming going through its satellite, McDougal anticipates a mixture of advertiser-supported (free) and scrambled subscription or pay-per-view programming. DBS Corp., he says, is working toward a universal decoding scheme for all DBS satellites so viewers won't need a separate decoder for each service. There will also be a huge potential for FM stereo and teletext services through the DBS Corp. system.

High resolution

For its part, CBS is looking to the sky to test and implement a major video development expected in the mid- to late-80's called HDTV (*High Definition Television*). The intent of HDTV is to improve the resolution of the video picture in the U.S. from the current NTSC standard of 525 horizontal scanning lines to 1125 lines. The aspect ratio (the ratio of horizontal to vertical measure) would also change from the present 4:3, to 5:3, which more closely resembles the aspect ratio of theatrical films. Unfortunately, HDTV

signals will require a wider bandwidth than NTSC television signals because they contain much more information.

The CBS plan calls for four satellites (one per time zone), each capable of transmitting three HDTV channels. One channel would be used to feed HDTV programming to CBS affiliates for rebroadcast. A second channel would probably feature subscription and/or pay-per-view programming for individual subscribers. The third HDTV channel would be a professional service, most likely used for distributing HDTV movies to theaters, where they would be shown using a HDTV projection-television system.

As it appears now, HDTV signals will be incompatible with current TV receivers, and a report prepared for CBS indicates that HDTV sets will cost about 20% to 30% more than a standard television. But CBS is also looking at ways to make HDTV signals "transcodable," that is, to allow them to be viewed on NTSC sets also. In addition, work is being done by CBS and other HDTV developers to establish international standards to avoid the NTSC/PAL/SECAM compatibility problems broadcasters face today; work is also being done to find ways to reduce the bandwidth required for HDTV signals.

Home receiver

To receive the DBS programming, each subscriber will need a home DBS receiver. To prompt potential suppliers to develop those receivers, STC early this year issued provisional specifications for their system's home receiver, a block diagram of which receiver is shown in Fig. 3.

The eventual cost of the home receiver is difficult to predict at this time. One reason for that is that work is continuing in an effort to reduce the cost of several of the critical microwave components. The GaAs FET's used in the circuit, for instance, cost between \$10 and \$20 each when bought in large quantities. Among other things, ways are being sought to improve the manufacturing yield (the ratio of the good circuits or devices to the total produced) and to place as much of the circuitry as possible—including multiple GaAs FET's—on a single IC. That is turning out to be more difficult than with ordinary semiconductors, but researchers are confident that solutions will be found by the mid-1980's, when many of the DBS services are expected to begin.

Even with the uncertainties that remain, STC figures that consumers will need to pay "several hundred dollars" for the receiver electronics, or will be able to lease the receiver from the company for about \$25 per month, including the subscription to the basic three-channel service. Those that chose to buy the electronics can expect to pay \$14 to \$18 for the subscription alone. In addition, a one-meter dish antenna will have to be bought and installed at a cost of about \$100. Stanley Hubbard, of USSB, predicts that mass-produced receivers, capable of tuning all possible DBS channels in the 12-GHz range, will be available for \$250 or less. R-E

(continued next month)

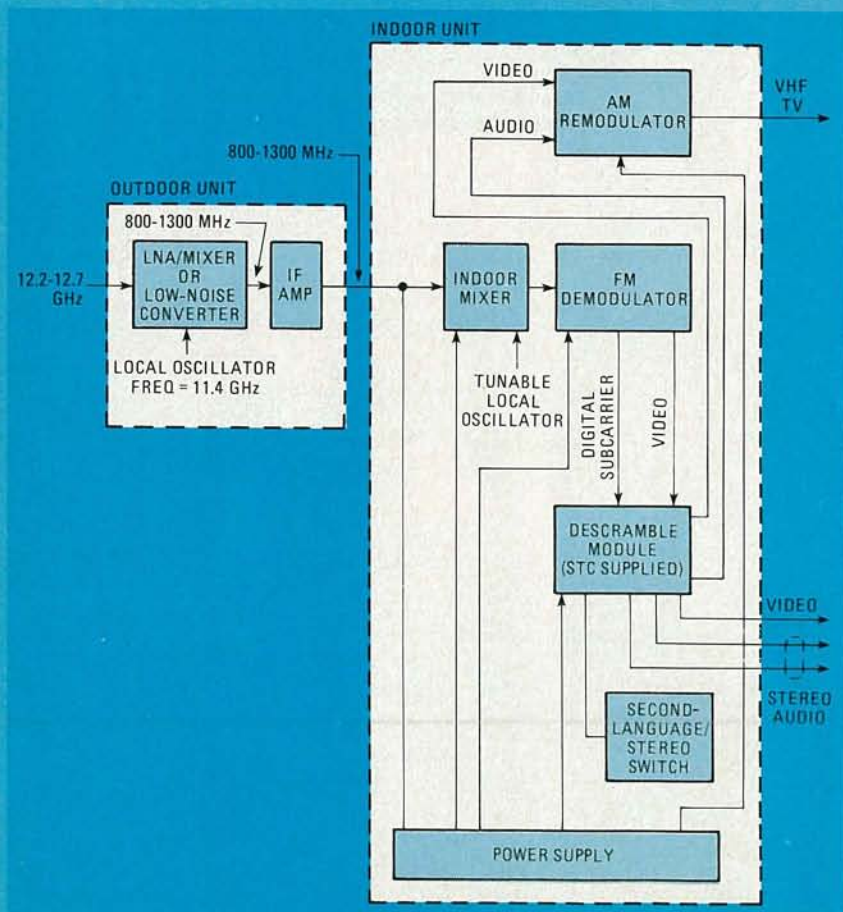


FIG. 3—BLOCK DIAGRAM of STC's proposed receiver. Descrambling will be handled by an STC-supplied module.

NEW VIDEO COMPONENTS

From the looks of things, the video revolution is nowhere near over. Here's what's new in home-entertainment equipment, and what to expect in the next couple of years.

DANNY GOODMAN

IN THE LAST FEW YEARS, WE'VE SEEN THE INTEREST AND activity in video grow at an almost explosive rate, and there is no sign of that letting up. Consider, for instance, that VCR's continue to sell briskly; pre-recorded videotape rentals are way up; and, according to an RCA estimate, those who own videodisc players also own, on the average, a library of about 30 discs. It's also expected that video camera sales in 1982 may double those of the previous year.

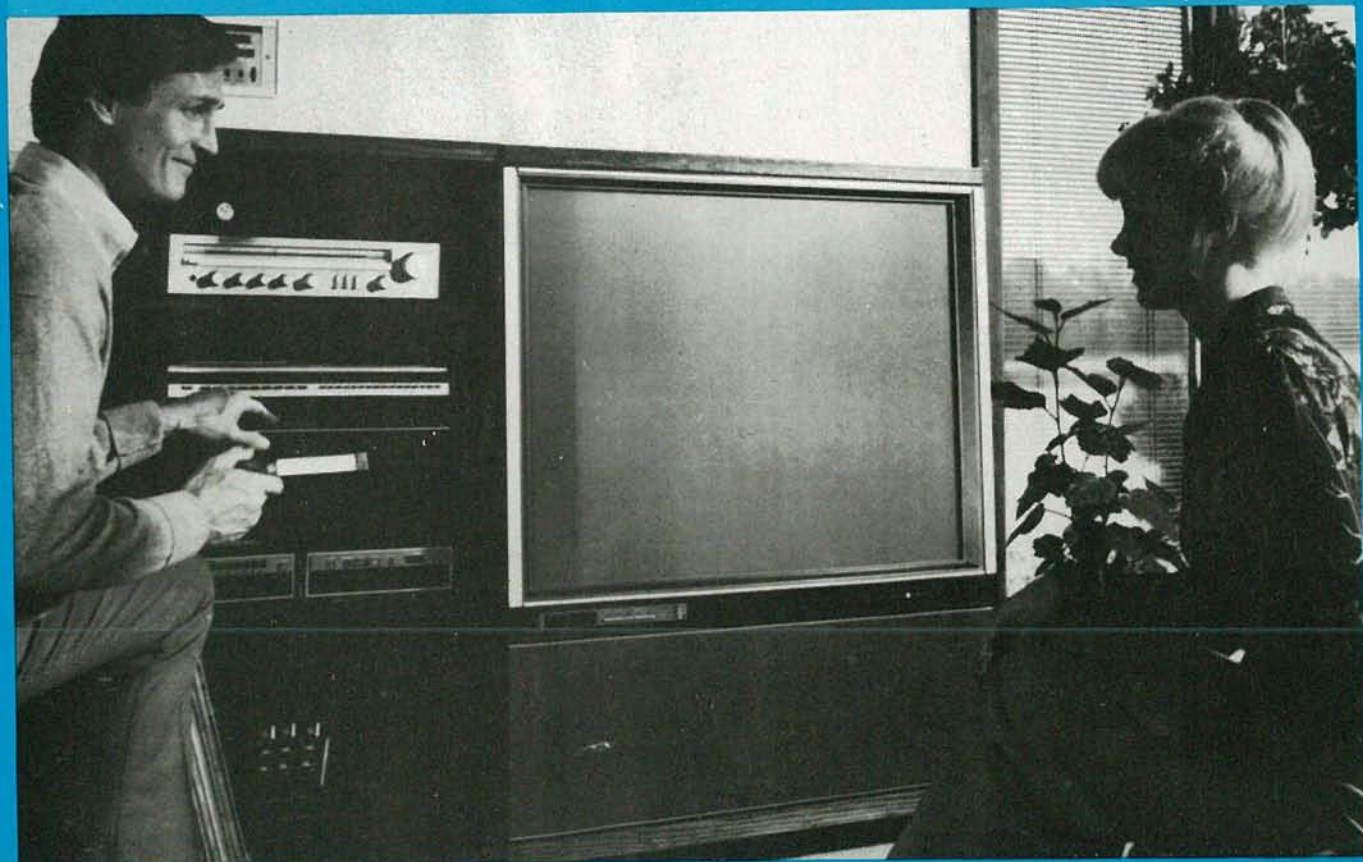
Manufacturers of home video equipment are very much aware of the consumer's love affair with video. They continue to produce better performing, more advanced equipment in new shapes, sizes, and concepts—all designed to lure more of us into the video picture. And, from the looks of the trend-setting video innovations already here and expected soon, they're likely to succeed.

HI-FI VCR

Stereo-audio for broadcast TV is still only in the planning stage in the U.S., but the absence of a broadcast

standard isn't preventing stereo from being included in several home videocassette-recorders. For now, stereo is limited to the upper end of the VCR price scale. In the VHS format, for example, top-of-the-line models from Panasonic (PV-1780), JVC (HR-7650), Quasar (VH5623UW), and General Electric (IVCR3018W) all feature stereo-audio inputs and outputs, including Dolby noise reduction. As high-end models, costing roughly \$1500 to \$1600, those units also have the deluxe special effects features and remote control offered on last season's expensive recorders. In the Beta format, Marantz has a home deck, the model VR200, with stereo and Dolby C noise reduction. That deck lists for \$1295.

Although there are few pre-recorded stereo-videocassettes currently available, there are other possible sources for two-channel audio, including FM simulcasts of TV concerts, stereo videodisc players, and your own dubbing. But, as stereo VCR's become more popular, you can be sure that much more pre-recorded material will become available. The quality of the audio, howev-



er, may not measure up to what you might expect from a hi-fi tape recording. The audio tracks on videotape are narrow, and the tape speed relative to the fixed audio tape-heads (ranging from 1.31- to 0.44-inches-per-second depending on format and selected speed) is slower than an audio cassette's 1 7/8-inches-per-second. But if the stereo is played through a hi-fi amplifier and speakers, it will provide a spatial feeling and left-right separation that you can't get from any of the TV stereo-audio adapters currently on the market. And stereo isn't just limited to the home—a new portable VHS-format VCR, the JVC HR-2650, lets you record your own stereo material in the field. That unit sells for about \$1150.

Finding stereo in such compact recorders is not surprising considering that the trend in new portables is to equip them with as many features as the stay-at-home models. Today, the full-featured portable essentially breaks the home VCR's tape playing and tuning/timing elements into separate, mobile components. That arrangement offers the most in flexibility, since the VCR system can be a compact portable-recorder for remote taping with the addition of a camera, but can also act as a sophisticated home deck with such features as infrared remote control, multi-day/event time-shift programming, and special effects.

The lightweights

The race is on in the industry to see who can develop the smallest and the lightest VCR. Sony's SL-2000 (\$1150) and the Zenith VR-9800, which is sold with its companion tuner for \$1425, are the lightest Beta portables, weighing just 9.5 pounds. But several lighter (8.5 pounds) VHS-portables were recently introduced; those are built on the same Matsushita chassis and carry brand names like Magnavox, Panasonic, Quasar, General Electric, and Canon. Those units range in price from about \$750 to \$1500. The high-end models are full-featured VCR's that include, among other things, 14-day, 4-event programmable tuner/timers.

An intriguing new portable VCR was announced this year; it was developed by JVC and is called the Ultra Compact Machine, or UCM for short. It is just what the name implies. Though not yet available, Sharp has demonstrated a model of that machine, the VC220, which weighs just 5.8 pounds, including battery, and measures approximately 7 x 2 3/4 x 9 1/4 inches. The key to the UCM is the VHS-C ("C" for compact) format cassette. About the size of a pack of cigarettes, a VHS-C cassette contains enough 1/2-inch videotape for about 20 minutes of recording on a UCM recorder. You can playback the tape through a TV, or edit on to a longer-playing regular-sized deck. You can also slip the VHS-C cassette into an adapter that makes the tiny tape playable on a standard VHS recorder.



SOLID-STATE IMAGE SENSORS, such as the one used in this NEC TC-100 color camera may eventually replace the cumbersome and power-hungry image tubes of today.

Many more home video-recordists are eagerly awaiting the all-in-one camera-recorders currently under development by several Japanese companies. Those companies are working toward a uniform 1/4-inch videotape standard to help avoid another battle such as the one over the Beta/VHS format. But the way things are going, even if a standards agreement is reached in 1983, it may be 1985 or later before finished products reach the store shelves.

Still looking to the future, most of you are probably familiar with the 1/4-inch VCR format—CVC—from Japan's Funai Electric. In the U.S., their small, five-pound portable recorder is sold by Technicolor. But an even smaller recorder, using the same CVC format, will be marketed in the U.S. in 1983 by Grundig. Grundig, a popular brand in Germany, has been selling their five-pound VP-100 successfully in Europe. It measures only 4 1/4 x 2 3/4 x 1/2 inches and can be easily mistaken for a portable audio-cassette player; among its features are variable speed and freeze frame.

Lights, camera...

The producer and director in us all will appreciate the features found on advanced color cameras. General Electric's two newest cameras, for example, both feature automatic focusing, relieving the cameraman of that chore. The more expensive of the two, the model 1CVC3035E, which sells for \$1350, uses a Newvicon-style pick-up tube that is designed to reduce "picture lag"—the tendency of an image to persist, thus streaking when the camera pans away from it. In addition, a control panel on the camera gives the operator fingertip command of fade-in/fade-out between scenes, and a built-in character generator capable of adding titles of up to 60 characters.

An exciting bit of news is the development of advanced solid-state imaging devices. Those, it is hoped, will eventually replace the power-hungry and cumbersome image tubes found in most of today's cameras. The heart of Hitachi's VK-C2000 color camera, for instance, is a one-IC MOS image-sensor. That sensor, which is about the size of a postage stamp, consists of wafers of color filters over an intricate grid of semiconductor material. The material appears to be impervious to the burn-in that many cameras suffer when the lens is directed at a bright light. With its electronic viewfinder, the compact camera weighs only 3 3/4 pounds, but carries a heavy price tag—about \$2,000.

Another type of solid-state image sensor is a CCD (Charge Coupled Device). NEC, for one, is currently developing a CCD color camera. The CCD in that camera will consist of a mosaic of 384 x 490 picture-elements and is expected to produce 250 horizontal lines of resolution. And, while the cost of cameras with solid-state image sensors will initially be very high, as manufacturers find ways to produce those sensors more efficiently, that cost should drop significantly.

For the do-it-yourself recordists, several new accessories can be very helpful for producing more varied programming. Although converters that let you videotape 35-mm slides for home-spun travelogues have been available for some time now, Sony has introduced a more advanced version, the HVT-3000 (\$179.95), that gives you the option of doing the same with print film negatives when used with their model HVC-2400 (\$1250) color camera. That camera converts the negative images to positives before they're recorded using a standard color video-camera.

More dramatic and professional productions can be done using Sony's new HVS120K special-effects kit, which consists of a model HVS-2020 black-and-white camera, model HVS-2000 special-effects generator, graphics aids, etc. If the kit is used with a color VCR-camera, you can superimpose either titles and graphics (from a six-color palette), or any image from the black-and-white camera, over the color-camera image. A slide-control fades titles in and out for quite a professional look. The generator also lets you alter color

keying (up to six different colors), which can make for some interesting "sci-fi" footage—as does electronic black-and-white reversal. The kit, which should be available by the time you read this, is expected to sell for around \$550.

Anyone with real creative talent will want to be first on the block with the *Video Scribe* electronic color-video titler (\$795), from Comprehensive Video Supply Corporation. You can be a real artist, using the device's 80 letters, numbers, and graphic elements, all addressable by a flat-membrane keyboard, to draw colorful titles and animation sequences in 8 colors—red, green, yellow, blue, buff, powder blue, orange, and magenta. Graphic output goes directly to the video input of your VCR.

Videodiscs

Videodiscs aren't generating all of the excitement they did about a year ago, but the two current formats, RCA's CED and Philips' LaserVision optical format, have been making some news as more consumers are selecting videodiscs as an alternate program source.

New on the scene are stereo CED players from RCA, Toshiba, and Hitachi. RCA now offers two stereo models, both of which contain CX noise-reduction decoders (stereo CED discs are CX encoded). The top-of-the line *SGT250* sells for about \$449. That model features infrared remote control and an automatic start function, which replaces the front-panel control lever. Toshiba offers a \$90 accessory that allows owners of their monaural CED-players to get the full benefit of CX-encoded stereo discs.

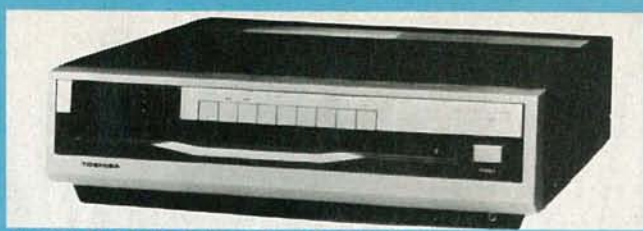
The LaserVision group also has chosen CX noise-reduction. Two North American Philips companies (Magnavox and Sylvania) have newly styled stereo LaserVision players, featuring remote controlled random frame-access and special effects. Those players sell for about \$750 and are made for N.A.P. by Pioneer.

LaserVision backers are still looking toward the future with the knowledge that their optical videodisc-players can be more than simple playback devices. North American Philips (Magnavox and Sylvania) has been demonstrating how their players can be used interactively with home computers. Interfaces exist now that allow computers like the *Apple II* and Texas Instruments *TI-99/4A* to control a videodisc player's search and frame-access functions. Soon, we're likely to see very realistic video-game-type interactive video based on such a setup. The use of a videodisc will allow for realism unmatched by any other home or arcade game.

Video separates

With all of the new video sources available, you need something to use for display. That display device is, of course, your home television set—but even it is undergoing some major changes. For one thing, more and more sets are including separate video and audio input jacks in addition to the antenna-input terminals. Previously, the video and audio signals from a VCR or other video source had to be fed into an RF modulator, with the modulator's RF-output fed to the set's antenna terminals. The set would then demodulate the RF signal to extract the audio and video information. The addition of the separate audio and video inputs skips all of those steps and eliminates the losses caused by them.

Even more significant, however, is the trend toward video separates or components. In a component TV-system, the TV set is broken up into separate units; just how it is broken up varies from manufacturer to manufacturer, but often the system is made up of a separate video monitor, tuner, controller/switcher, and speakers. The monitors used in component TV-systems are capable of high-quality displays, but there is a premium to pay for that quality, and for the flexibility that those units can offer. For instance, Pioneer's *Foresight* 19-inch monitor lists for about \$850, while their 25-inch monitor lists for \$1500; the companion tuner for those lists for about \$500. Also, you should bear in mind that



AMONG THE FEATURES of this VP-500 CED videodisc player are stereo audio and CX noise reduction.

the displayed video can only be as good as what is input.

But, on the positive side, with a component approach to video you have the flexibility to build your system one piece at a time. With Sanyo's *Pro-Ponent* series, for instance, you can start with their model *AVM195* 19-inch monitor (\$600) and their model *AVT95* tuner/controller (\$400). The audio outputs can be fed to a separate stereo audio-amplifier and played through stereo speakers; the monitor also has its own built-in five-watt-per-channel stereo amplifier.

Zenith takes the component approach a bit farther. In its *Hi-Tech* system, all of the components feed into a central source selector, the model *CV-540*. That is one system in which the tuner and selector are separate units. The tuner here, the model *CV-510* lists for \$279.95, and features an infrared remote control. Getting back to the source selector, that \$169.95 component can accept up to six video and audio inputs. The video inputs and mono audio-inputs are routed by the selector to the system's model *CV-1950* 19-inch monitor, which lists for \$469.95; stereo audio-inputs are routed to a 20-watt-per-channel stereo amplifier that lists for \$149.95. That amplifier, the model *CV-520*, is designed specifically for the component TV-system; some manufacturers simply use one from their standard audio line. The amplifier also has phono and auxiliary inputs for adding hi-fi audio sources to your system.

Like many of the other new component-TV monitors, the Zenith monitor has an RGB (Red-Green-Blue) input. While such outputs are currently available on only a few consumer products, such as the IBM *Personal Computer*, they offer a great advantage. Unlike video-modulated RF (such as broadcast TV) or composite video signals, an RGB signal is fed directly to the video amplifiers, bypassing many stages of lossy signal conversion. With an RGB input, most component-TV monitors are capable of high-resolution displays.

A component-TV system, with its stereo amplifiers, speakers, and controller functions sounds much like a stereo-audio setup. In fact, it's getting harder and harder to classify entertainment as either all video or all audio; instead, it's simpler to consider the two as part of a broader home-entertainment category. Consider, for instance, Jensen's model *AVS 1500* audio/video receiver; it's an AM/FM- and TV-receiver, as well as a stereo amplifier, all in one package. The AM/FM receiver section features a synthesized AM/FM-tuner and a digital frequency-readout. On the TV side, there's a 133-channel (including cable channels) TV tuner and National Semiconductor's DNR (Dynamic Noise Reduction) circuit. The amplifier has an output of 50-watts-per-channel. The unit lists for \$990. Two companion monitors are available—the 19-inch model *AVS-3190*, which lists for \$800, and the 25-inch model *AVS-3250*, which lists for \$1030.

Teknika is another company that blends audio and video into the same components. Their \$1200 *ATV-19* system consists of a 19-inch monitor, AM/FM/TV tuner, and 2 speakers. A 25-inch monitor is also available.

Tiny TV's

There's been a great deal of interest in small-sized TV and this year the first commercial flat screen (or pocket) TV's may become available. Sony's 2-inch diagonal screen



POCKET-SIZED TELEVISION, such as the Sony *Watchman* shown here, may finally become available this year.

Watchman uses a new picture tube design which places the electron gun below and parallel to the screen surface. The scanning electron beam is magnetically "bent" to hit the screen at a perpendicular; the result is a TV only 1 1/4 inches thick. The estimated retail for that set is around \$300.

Sinclair, developers of another bent electron-beam flat TV-tube, may get their 3-inch TV on the U.S. market soon; its price is expected to be in the \$100 range. Both the Sony and Sinclair tubes are black-and-white only.

The LCD flat-screen display is still under development, but may be still a year or two away. Seiko has been demonstrating a prototype of a wristwatch TV. Only the LCD screen is worn on the wrist. The bulk of the receiver, including an FM-stereo receiver, is housed in a vest-pocket sized unit.

Turning to small-sized color TV, Panasonic is offering what may be the world's smallest color TV/monitor; its *CT-3311* sells for \$499.95 and features a screen size of just 2.6 inches. It even has separate audio and video input jacks.

Projection TV

From the very small, we turn to the "giants" of home video—projection TV. More companies are heading toward one-piece, rear-screen projection systems. They are typically more expensive than front-projection units, but recent scaling down of chassis size, as RCA has done for example, allows 40- to 45-inch screens to be placed in a cabinet that takes up no more floor space than a console TV. New for this year are many cabinet styles and sizes from General Electric, Mitsubishi, NEC, Panasonic, Quasar, Sharp, and Sony, to name a few; prices for those are in the \$3,000-\$4,000 range. Many like GE's new *Widescreen 40* feature built-in stereo amplifiers and twin, two-way speakers.

It may sound odd to put the words projection and portable together, but Kloss Video Corp has developed a portable (i.e. movable) projection-TV called the *Novabeam Model Two*. The unit is about the size of a 19-inch portable TV (though only 12-inches deep), and opens to project a 5-foot diagonal image on any wall. Room lighting needs to be controlled for best viewing, yet the ability to use a plain wall as a screen allows the unit to be played anywhere in the house and makes the viewing angle much less critical. Assuming that most projection-TV buyers have VCR's with TV tuners, Kloss eliminated the tuner completely from that model, which helped keep the price to about \$2200.

It appears, then, that there's more available in video this year than ever before. The so-called video revolution is continuing, and it's getting easier all the time to become a part of it.

R-E

LIST OF MANUFACTURERS

Canon U.S.A., Inc.
One Canon Plaza
Lake Success, NY 11042

Comprehensive Video Supply Corp.
148 Veterans Dr.
Northvale, NJ 07647

General Electric
Portsmouth, VA 23705

Grundig AG
Kurgartenstrabe 37
Furth, W. Germany 8510

Hitachi
401 W. Artesia Blvd.
Compton, CA 90220

Jensen Sound Laboratories
4136 N. United Parkway
Schiller Park, IL 60176

JVC
41 Slater Drive
Elmwood Park, NJ 07407

Kloss Video Corporation
145 Sidney Street
Cambridge, MA 02139

N.A.P. Consumer Electronics Corporation
1-40 and Straw Plains Pike
Knoxville, TN 37914

Marantz
20525 Nordhoff St.
Chatsworth, CA 91311

Mitsubishi
7045 N. Ridgeway Ave.
Lincolnwood, IL 60645

NEC Home Electronics
1401 W. Estes Ave.
Elk Grove Village, IL 60007

Panasonic
One Panasonic Way
Secaucus, NJ 07094

Pioneer Electronics
1925 E. Dominguez St.
Long Beach, CA 90810

Quasar Co.
9401 W. Grand Ave.
Franklin Park, IL 60131

RCA
600 N. Sherman Dr.
Indianapolis, IN 46201

Sanyo Electric Inc.
1200 W. Artesia Blvd.
Compton, CA 90220

Sharp Electronics Corp.
10 Sharp Plaza
Paramus, NJ 07652

Sinclair Research, Ltd.
50 Staniford St.
Boston, MA 02114

Sony
1 Sony Dr.
Park Ridge, NJ 07656

Technicolor
299 Kalmus Dr.
Costa Mesa, CA 92626

Teknika
1633 Broadway
New York, NY 10019

Toshiba
82 Totowa Rd.
Wayne, NJ 07470

Zenith Radio Corporation
1000 Milwaukee Ave.
Glenview, IL 60025



VIDEO ACCESSORIES

With the three devices described below, there's no need to put up with poor-quality videotapes any longer. While they can't work miracles, they can make an amazing difference in the recorded image.

GORDON McCOMB

THERE WAS A TIME NOT TOO LONG AGO WHEN THE AMATEUR video enthusiast endured videotapes with poor or shifted colors, fuzzy details, and picture instabilities. But as the home-video industry has grown to include millions of owners of videocassette recorders (VCR's), add-on components have been introduced to help clear up the annoying color, detail, and picture problems that have plagued videophiles for years.

There are three video components that no top-quality video system should be without (at least not for long): a color processor, an image enhancer, and a sync stabilizer. Color processors can shift, exaggerate, even delete the color information on signals from a tape, disc, camera, or any other standard NTSC format video source. Image enhancers improve the high-frequency response and help restore clarity and sharpness to old tapes, dubs, and less than perfect video equipment. The vertical sync pulses on many pre-recorded videotapes have been altered in an attempt to prevent unauthorized VCR-to-VCR duplication. Unfortunately, the circuitry in some TV's can't handle those altered sync pulses. That results in vertical hold instability and horizontal 'tearing' of the picture. To eliminate the problem, sync stabilizers are used to restore the altered sync pulses to their original shape.

Let's examine each of those components further, and discuss the ways they can be used to improve your video pictures.

Color processors

Color processors enable full manipulation of the color and brightness components of any standard NTSC video signal. Colors can be shifted so that reds become green, greens become blue, and blues become red. The intensity of the image and colors can also be modified. A videotape with washed out colors can be rejuvenated by increasing the CHROMA or COLOR controls on a color processor. Special effects such as fading from a black-and-white to a full color scene can also be created.

The main function of the color processor is to correct for errors in the brightness, color, and tint of video signals, particularly with equipment lacking color and tint controls, as when dubbing between one VCR to another.

Three important fundamentals must be considered when reviewing the operation of television signals and color processors. Those fundamentals are luminance, hue, and saturation. The luminance signal occupies a bandwidth of 0 to 4.2 MHz, and includes all the brightness information seen in a black-and-white or color-TV picture.

Hue (adjusted by the TINT control on most color-TV sets) is the actual colors (red, blue, green, etc.) seen in a picture. Saturation (adjusted by the COLOR control on most color-TV sets) is the degree to which a hue is diluted by white light, enabling differentiation between vivid and weak shades of the same color. For example, a vivid blue is a saturated hue; a pastel blue is an unsaturated hue.

A color-TV picture consists of hue and saturation information, in addition to the black-and-white luminance information. The combination of hue and saturation is called chroma. The chroma signal is transmitted as a 3.58 MHz subcarrier of the luminance signal and consists of two color-difference signals, designated I and Q. The I signal determines the saturation, and the Q signal determines the hue. The I signal is transmitted as an amplitude modulation of the 3.58 MHz subcarrier signal, and has a bandwidth of 0 to 1.5 MHz. The Q signal frequency modulates the 3.58 subcarrier, and has a bandwidth of 0 to .5 MHz. A 3.58 MHz color-burst signal is inserted after every horizontal sync pulse in the composite video signal. The color-burst signal is used to synchronize the phase of the 3.58 MHz oscillator in the TV receiver, enabling proper color demodulation.

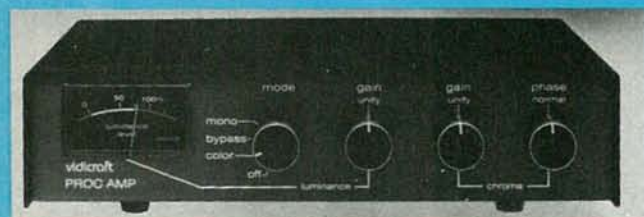
The instantaneous phase angle of the modulated chroma signal, with respect to reference color burst, controls the hue. The variations in the amplitude of the modulated chroma signal, in relation to the corresponding brightness



THIS COLOR PROCESSOR from Showtime Video Ventures features independent chroma and burst amp controls.

or luminance level of the scene, determines the saturation of the color.

Color processors can manipulate the amplitude and phase of the chroma signal, and thus can vary the tint and saturation of any scene. Old movies, either recorded or broadcast with weak color, can be rejuvenated by boosting the amplitude (saturation) of the chroma signal. Poorly taped home movies with incorrect colors can be corrected by altering the phase angle of the modulated chroma signal. Of course, there is no way to selectively boost, retard, or shift only certain colors with this type of color processor. If skin tones are excessively yellow, shifting the scene to normalize the skin color may also change the color of blue swimsuits to green. But those side effects, usually slight, are more than compensated for by the overall improvement of the video image.



A TRUE PROCESSING AMPLIFIER, this unit from Vidicraft allows full control over both the color and sync of the video image.

Operating a color processor, like the one manufactured by Showtime Video Ventures (2715 Fifth Street, Tillamook, OR 97141), is relatively easy. Showtime's color processor, model VV-777P, sells for \$377 and uses a MODE switch along with four front-panel controls for manipulation of the brightness, color saturation, and tint of a TV picture.

The LUMA control determines the relative amplitude of the luminance signal being processed. Decreasing the LUMA control darkens the scene; increasing the control brightens the scene. Color saturation is determined by the CHROMA and BURST AMP controls. The CHROMA control adjusts the amplitude of the color subcarrier with respect to luminance (hence the LUMA and CHROMA controls are interactive), and so alters the degree of color intensity. A BURST AMP control, unique to the VV-777P, adjusts the color-burst amplitude without actually affecting the subcarrier level itself, and is used as a fine adjustment for resolution and proper skin tone. Rotating the BURST PHASE control counter-clockwise a quarter-turn shifts reds toward blue, blues toward green, and green toward red. Rotating the BURST PHASE a quarter-turn clockwise will have the reverse effect. Secondary and complementary colors undergo a similar change. Switching the MODE SWITCH to MONOCHROME squelches the color-burst signal, rendering a black-and-white picture.

Showtime's VV-777P color processor, like nearly all video components made by Showtime and others, operates with a demodulated composite video signal. To use the color processor with an RF signal (from the RF output of a VCR or from an antenna download, for example) the signal must first be demodulated by use of an RF demodulator. An RF modulator or converter is then used to remodulate the processed video to a suitable RF channel (usually Channel 3 or 4) for reception on a standard TV set. Most VCR's have separate RF IN/OUT and VIDEO IN/OUT connections, and in some instances, an RF demodulator and RF modulator may not be required.



AN RF CONVERTER or modulator is used to modulate separate audio and video signals for viewing on a standard TV.



SYNC STABILIZERS, such as this one from Vidicraft, are used to eliminate side effects caused by tape-duplication prevention schemes.

Image enhancers

Image enhancers are used to improve the detail and sharpness of programs viewed off-the-air, while recording or playing videotapes, and during tape-to-tape duplication. Image enhancement involves boosting or accentuating high-frequency video signals, thus increasing overall detail.

The picture displayed on a TV screen is produced by varying the intensity of an electron beam (or three beams in the case of a color TV) as the beam sweeps across from left to right, and from top to bottom on the face of the picture tube. Extremely fine detail and outlines require the beam to vary its intensity more often—up to several million times per second. The smaller the detail in the image, the higher the frequency required to produce it. Likewise, the sharper the outline of an object, the faster the beam must cycle from light to dark or vice versa. Again, high frequencies are involved.

The bandwidth of the TV luminance (brightness) channel is from 0 to 4.32 MHz. However, high frequencies are often distorted and rolled off by the less than perfect circuitry used in most home VCR's and TV sets. That combined with the maximum bandwidth of most VCR's of about 2.0 MHz, leaves only a 0 to 2 MHz bandwidth for all the picture detail and information displayed in any given video frame. Reduplication of a videotape—three or four generations from the original—will yield a muddy, hardly discernable image



THE SHARPNESS AND DETAIL of both recorded and broadcast video can be improved by using an image enhancer.

where only the largest objects can be recognized.

The easiest way to think of an image enhancer is to look at it as a video version of an audio-frequency equalizer. But instead of dealing with audio frequencies, video image enhancers handle frequencies of 1.0 MHz and beyond. By boosting high-frequency signals, much in the same way as an equalizer or treble control boosts audio high frequencies, sharpness and detail can be improved. Low-frequency signals (0 to .5 MHz) are ignored by the image enhancer as they do not contain any detail or sharpness information.

Most image enhancers split the high-frequency signals into separate bands by using a comb filter, then selectively amplify the signals giving them an extra boost. Preshoots and aftershoots are added to the signal, driving the signal transitions (dark-to-light and light-to-dark) at a steeper amplitude than normal. That helps eliminate the soft edges in light-to-dark and dark-to-light transitions in the picture.

Of course, an image enhancer cannot produce something that isn't there in the first place. If the delineation of each blade of grass no longer exists in the image recorded on a videotape, no enhancer—no matter how well designed or made—will restore the lost detail.

Boosting the high-frequency content of a video signal also brings out noise, seen as snow. Most enhancers incorporate noise-reduction circuitry in addition to the main enhancement circuitry. The noise-reduction circuit reduces the level of enhancement to compensate for the introduction of snow, and usually works in direct opposition to the enhancement circuit. Noise, usually of high-frequency content, has a lower amplitude than most desired signals. Noise reduction circuitry is designed primarily to suppress low-amplitude, high-frequency signals.

For example, Vidicraft's (Box 13374, Portland, OR 97213) *Detailer I* (selling for approx. \$149) and *Detailer II* (selling for approx. \$295) incorporate a VNX control that suppresses noise. It performs the opposite function of the DETAIL control, but uses a different set of amplitude thresholds, thus removing a majority of the increased noise and a smaller portion of the enhanced signal. There are trade-offs to be considered when using an image enhancer, as excessive noise cancellation can totally counteract all efforts to enhance detail. The proper enhancement level for any given program, with relation to detail and snow, is performed by interactive adjustment of both the enhancement and noise-reduction controls.

Connecting an image enhancer to a video system is similar to the hook-up procedures for the color processor, in that all units on the market can accept demodulated video signals only. Depending on the installation, RF modulators and/or RF demodulators may be required.

The operation of an image enhancer, such as the *Detailer II* from Vidicraft, uses several knobs for enhancement control. The DETAIL control is used to compensate for detail loss encountered in VCR recording and playback. It can be used to exaggerate detail before the signal is recorded (and so precompensating for high frequency roll-off inherent in home VCR's) or to help restore a VCR signal on playback. The SHARPNESS control increases the overall sharpness of the picture whether the signal be from a tape, camera, videodisc, or off-the-air. The SHARPNESS control is adjusted for maximum edge detail without creating false, black outlines. The VNX control (also called CORE, NOISE SUPPRESS, or NOISE CANCEL on some other models and brands) helps reduce snow and other low-amplitude, high-frequency interference—however, at the expense of detail. It's function is very much opposite that of the DETAIL control and is most helpful when sharpness-only enhancement is used. Normally, fine adjustments of all three controls must be made to obtain the best results, and the settings must be optimized for each program.

It should be noted that image enhancers improve the horizontal resolution of a video image—that is, fine details that occur from left to right in a TV picture. Image enhancers, unfortunately, aren't tested for any given horizontal resolution standard or minimum, since the quality of the original program will determine the results achieved by the enhancer.

Sync stabilizers

Many pre-recorded videotape program distributors, to protect against unauthorized duplication of their product, have altered the vertical sync pulses recorded on their videotapes. That technique, most often called *Copyguard*, (although similar systems with different tradenames are available) disables the sync circuits in a VCR. However, several side effects of sync alteration can appear on the screen of a standard television receiver. *Copyguarded* tapes have a tendency to make the picture on a TV screen roll and shake, as well as "tear" at the uppermost portion of the picture. Adjusting the vertical-hold control on the TV can correct for most of the problems, but many of the newer sets often have the vertical-hold control mounted inconveniently on the back of the chassis, or may even lack a vertical-hold control altogether.

Sync stabilizers restore the vertical sync pulses and provide sync stability to TV's and VCR's. In the NTSC broadcast standard used throughout the United States (as well as Canada, Mexico, and Japan), an electron beam scans 525 horizontal lines 30 times each second. Every picture then consists of 30 frames composed of 525 vertically stacked lines. The scanning process is actually done in a 2:1 interlace pattern, so 262.5 lines are scanned in one pass over the tube,

and the second 262.5 lines in a second pass. The scan lines of the first pass or field is comprised of all the odd-numbered lines; the second field is comprised of all the even-numbered lines.

When the beam reaches the bottom of the screen after each field, it is then directed to retrace back to the top of the screen, and repeat the scanning process again. The synchronization pulses required for the circuit to direct the beam to the top of the screen are contained in the vertical blanking interval.

Within that vertical blanking interval, six vertical sync pulses, bracketed on both sides by equalization pulses, are used to trigger the vertical sync circuits in the receiver (those pulses are at one-half the normal horizontal line rate). Vertical sync pulses have extremely long duty cycles of 87% (the longest duty cycle involved in TV sync timing). In most TV's, only one vertical sync pulse is required to trigger the vertical-sweep circuits. In VCR's, however, nearly all of the vertical-timing pulses are required to ensure full sync stability while recording.

Sync-alteration techniques, (again, there are numerous types currently in use), normally pass just one pulse—the remaining five pulses are changed or distorted.

Nearly all sync stabilizers currently on the market, such as MFJ Enterprise's (921 Louisville Rd., Starkville, MS 39759) model 1400 (selling for approx. \$80), reshape the vertical sync interval to the point where a stubborn TV or VCR will trigger properly. A timing circuit is normally used that triggers when the circuitry in the stabilizer detects the first vertical sync pulse. The stabilizer circuit then generates a single long pulse equal to the normal vertical sync interval (three horizontal lines long), overriding the altered signal produced by the tape. The sync stabilizer has a frequency adjustment to allow the unit to match its frequency with that of the incoming vertical sync pulses.

The actual half-line equalizing pulses of the vertical blank-

ing interval are lost by use of most sync stabilizers, meaning that the horizontal sync circuits in the TV lack a timing reference for three lines in each field, which can cause horizontal sync loss during the vertical blanking interval. This can create some picture instabilities as well as horizontal tearing at the top of the screen. Only one sync stabilizer, the Showtime Video Ventures' model VV-170P, recreates the pulses contained in the vertical interval. Usually, however, simply adding one long pulse to the vertical interval compensates for most of the serious side effects caused by *Copy-guard* or similar encoding, and produces a relatively stable, jitterless picture.

Sync stabilizers work only with demodulated video signals, although some units are now available with built-in RF modulators. To use a video-only stabilizer, the VIDEO OUT jack on the VCR is connected to the VIDEO IN on the stabilizer. An RF modulator is then used at the output of the stabilizer. It should also be noted that video processor components can be combined in one installation along with a sync stabilizer. Normally, there are no restrictions as to the order in which the components should be placed in the signal path. Be certain, however, that a sufficient video level (1 volt peak-to-peak) is supplied to the input of each component. Many video accessories incorporate distribution amplifiers (some with adjustable gain) that can be used to overcome excessive signal loss when passing the signal through several components. Video distribution amplifiers are available separately as well.

Quite a few video components have provisions whereby more than one input source can be connected to the unit at one time. A front-panel switch is used to select the desired input. Additionally, many units have several outputs allowing for multi-point signal distribution.

Operating a stabilizer requires the user to adjust one knob, the STABILIZE control. The control is usually readjusted for each tape viewed.

R-E



NEARLY ALL of the sync stabilizers on the market, including the one from MFJ enterprises shown above, reshape the vertical sync interval.

HOW TO CONNECT VIDEO COMPONENTS

Watching TV is no longer a matter of simply turning on the set and sitting back. Here's a look at some devices and ways to take the nuisance out of what should be a pleasure.

GARY McCLELLAN

THANKS TO CABLE AND SUBSCRIPTION TV, TV GAMES, AND personal computers, television is becoming more interesting and useful than ever before—not to mention a lot more fun. Furthermore, VCR's (Video Cassette Recorders) and videodisc players are making available to us an even wider range of entertainment. And, on the horizon, there is the promise of interactive videotex, which can allow us to shop and be informed from the comfort of our living rooms.

Hooking up all the equipment needed to an existing TV set can become a bit confusing, especially if several devices are involved; this situation is becoming more and more common these days. Most people start out with a VCR, and over a period of time add other equipment, turning a TV set into a video system. Perhaps there's a cable box with a decoder for sports and movies; then a programmable TV-game for the kids, a personal computer for Dad, and so on.

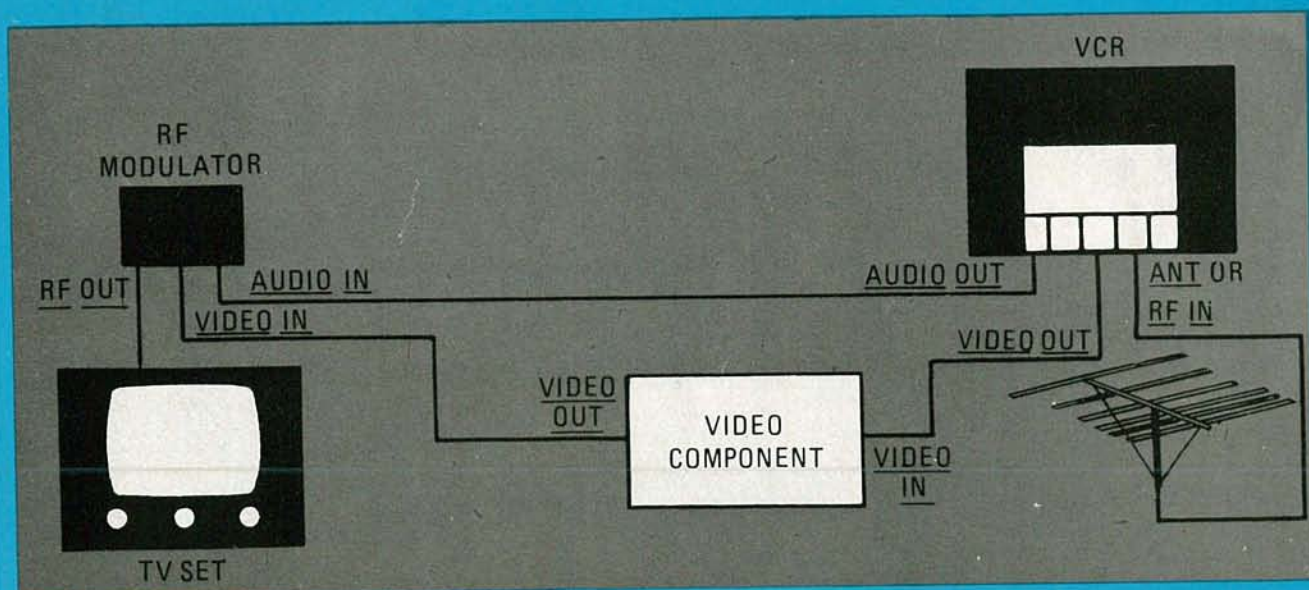
The result is often a rat's nest of wiring centered around a complicated switching arrangement. And, while you may understand it, it can make life very difficult for other people who use the system, especially if they have to change cables! Also of concern is the fact that every connection in that maze of cables causes signal losses that

can degrade picture quality. What's needed is an easy and simple way to mate all that new equipment with the TV set, creating a true integrated video-system that's easy to use and provides the highest quality possible.

In this article we'll discuss how to connect additional equipment to your TV set. We'll start with the simple addition of a VCR and build up to a "full house" video system featuring almost every device you are likely to want. We'll finish up by adding useful accessories such as cameras, enhancer/stabilizers, commercial editors, and others.

The hookups described will bring many benefits. For example, your video system will be easier for the family to use (no cable swapping), and there will be fewer connections and less signal loss. All you have to do is to read through the article to find the descriptions that best match your equipment, and then follow the diagrams to hook it up. It's as easy as that!

Of course, it isn't possible to cover every possible combination in this short space; there are just too many variations. Instead, we'll concentrate on the ones that most people are likely to use. With a little thought you should be able to adapt those ideas to your particular situation.



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CIRCLE 3 ON FREE INFORMATION CARD

Basics

Before you do anything, there are few simple rules to remember when making these hookups, and they boil down to one thing—*go first class, inside and out*. Lots of people will spend a thousand dollars or more for a VCR, \$15 and up per month for subscription-TV service...and then use a 15-year-old antenna and rotten cables. Naturally, they get angry when their expensive equipment performs poorly. The moral is to save a little extra cash for a new antenna and lead-in. If you don't, you'll wind up paying for your "economy" with poor reception. Read the following basics, and apply them to your own situation. Make any improvements required.

Start outside with the TV antenna and lead-in, if you use them. They are often overlooked because rooftop mounting makes antenna inspection an unpleasant chore. Yet, if your antenna and lead-in have been up 4-5 years or longer, you may not be getting the best picture possible. Why? To begin with, the sun's heat and ultraviolet radiation degrade the insulators and lead-in, which means increased signal-losses. Also, soot deposits and humidity will cause breakdown of the cable insulation, and wreak havoc with connections. The result is predictable; more losses and noisy pictures! And, in extreme cases, corroded connections, combined with a nearby CB or ham transmitter, will invite interference. So if your antenna and lead-in are four years old or more, check them out carefully, and replace them if necessary. The same goes for lightning arrestors and guy wires.

If you are going to replace your lead-in, you should use coaxial cable, rather than twin lead. For one thing, it tends to reduce noise pickup, and for another, it is more compatible with the newer equipment. In my own case, using coaxial cable cut ignition-noise pickup noticeably and reduced ghosting; the extra cost was worthwhile.

Once the antenna and lead-in have been taken care of, the antenna should be carefully oriented for the best picture-quality on all channels. Often that requires a compromise, but with patience, things can be worked out. Of course if you have a rotator, antenna positioning will be no problem!

Go first-class inside the house, too. The lead-in cable should be routed to the area where the video equipment is located, and F-type connectors used. If necessary, use a VHF/UHF signal-splitter between the cable and TV set.

Now turn your attention to the TV set you will be using. It should be in good condition and provide a good color picture. Adjust the fine-tuning control for the best sound and picture quality. If you don't get a good picture and sound, repairs or adjustments may be required; have them made. You should then be getting first-rate results, and will be ready to add other equipment.

You are going to need baluns (impedance-matching transformers), short lengths of cable with F-type connectors on each end, and other parts. Refer to Table 1 for a list of parts of this sort that you will probably require.

Adding a VCR

Video cassette recorders (VCR's) are very popular, and thanks to decreasing prices, are becoming quite widespread. Figure 1 shows a simple VCR/TV hookup. The antenna cable connects to a VHF/UHF splitter that goes to the VHF and UHF inputs of the VCR. From the VCR, separate cables are connected to the antenna inputs on the TV set.

Several remarks concerning that arrangement are in order. First, many older TV sets have balanced 300-ohm antenna inputs. To use the cable hookup shown, a 75-300-ohm matching transformer (balun) will be needed between the 75-ohm RG-59 cable and the TV set's VHF terminals. Don't try to do without the transformer or you may get a noisy picture. Second, some early VCR's don't have UHF output-terminals. If that's so in your case, run a short length of 300-ohm twin-lead from the VCR's UHF input-terminals to the UHF antenna-input on the TV set.

Adding CATV/MATV

In some parts of the country, cable TV (CATV) or master-antenna TV (MATV) takes the place of an outside antenna. Figure 2 shows a hookup for such systems. Note that the switch box is not required if just one of those services is available.

Your setup may differ somewhat from the one shown, but can probably be handled in a similar fashion. For example, your area may have CATV and still permit outside antennas. If that's the case, just substitute the antenna cable for the MATV connection to the switch box.

Subscription-TV service has become popular, thanks to its uncut, no-commercials, format of movies and sporting events and the fact that the programming is sent over the air and no cable-service is required. Figure 3 shows two simple ways to connect a subscription-TV decoder to your VCR and TV set.

The hookup shown in Fig. 3-a will work well except for one thing: when the decoder is turned on, the VCR or TV will receive *only* the decoder! That can be a problem if there are two good programs broadcast at the same time on subscription- and free TV.

The solution is shown in Fig. 3-b. By adding the a device like the *Channelizer* (made by Energy Video Corporation, 20371 Prairie St., Chatsworth, Ca. 91311), it is possible to receive all channels without any switching. You can tape one program for viewing later, and watch the other program right now. Special circuitry inside the unit makes that possible without interference, and the decoder output appears on a channel unused in your area (channel 3 in mine).

There may be an unused input on the subscription-TV decoder that can be used to connect the output from any other video equipment (game, computer, videodisc player, etc.) to it, and you can then use those devices without throwing any

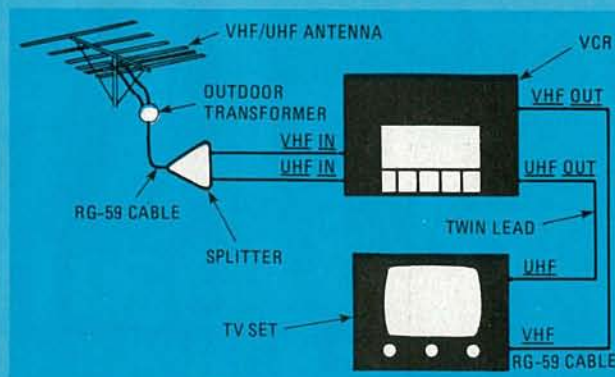


FIG. 1—CONNECT THE TV ANTENNA to your VCR, and then connect the recorder to your TV set. Keep the VHF and UHF signals separate.

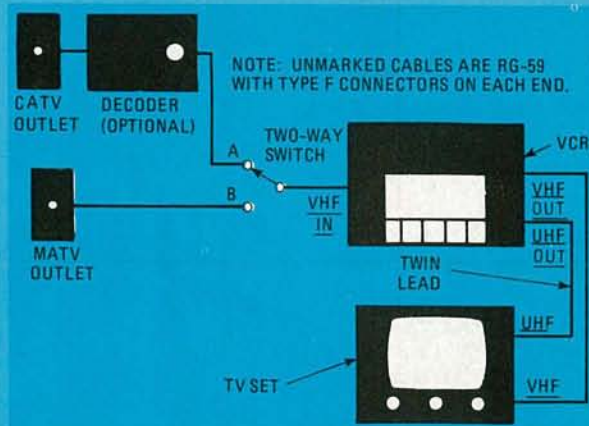


FIG. 2—A SIMPLE 2-WAY SWITCH will allow you to select between two TV-signal sources like CATV and MATV.

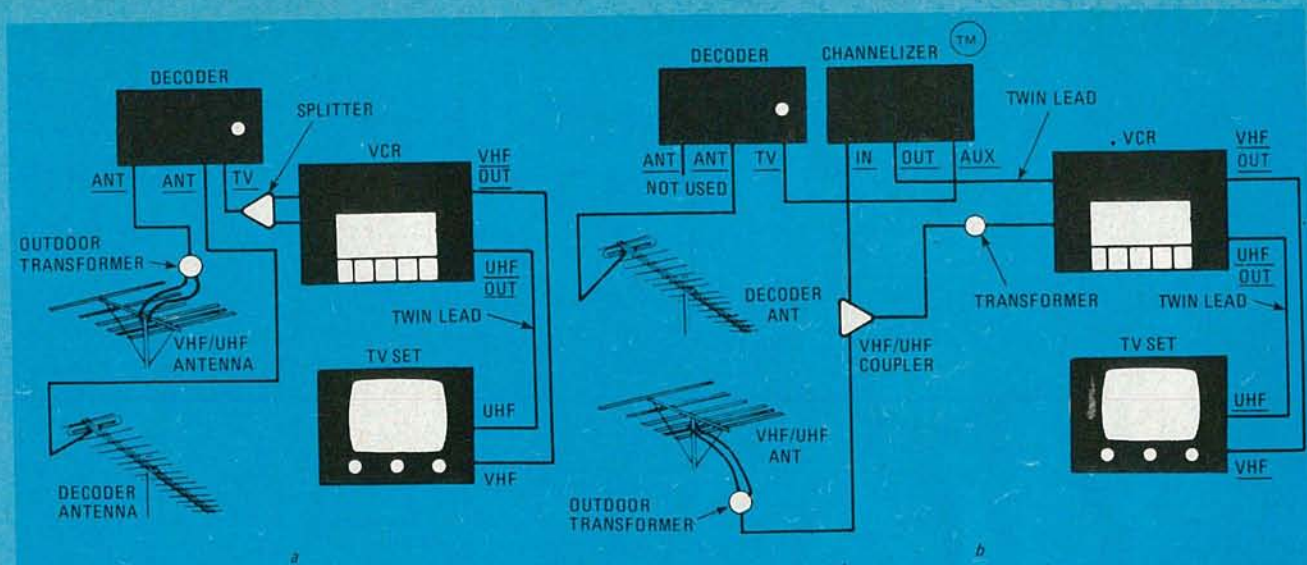


FIG. 3—A SUBSCRIPTION-TV DECODER in the line usually restricts you to receiving only one channel at a time (a). A device like the Channelizer (b) removes that restriction.

switches. Just set the TV set to your local "vacant channel," turn on the equipment, and go. That's great for the non-technical members of the family who hate to figure out switches!

Adding other equipment

Add-on devices like video games, personal computers, and videodisc players are rapidly increasing in popularity, and making one part of your video system is easy—just use a two-position switch as shown in Fig. 4.

That arrangement does have a drawback; it allows you to connect only one device at a time. If you're lucky enough, however, to have a device with a VHF antenna-input (perhaps a videodisc player), you can "daisy chain" it with another device. Figure 5 shows how that kind of hookup can be made. You might be able to connect the output of your video game to the antenna input of the videodisc player, and the output of the player to the switch. Often that method will work well, and will eliminate a second switch. However, since not all video equipment has an antenna input, that hookup may not be practical for you! The solution, then, may be a video switch, our next topic.

Getting sophisticated

All the hookups described so far have assumed that you would be adding just one device to your video system but the chances are that, with time, you will be adding several pieces of equipment. That can complicate things greatly if you are not careful, and you can end up with a rat's nest of cables. But take heart; there's a way around that problem.

This solution is simple enough: Use a multi-position video switch to select the output of a particular device for viewing on the TV set or for recording.

Actually, the term "video switch" is a misnomer. Those devices actually switch VHF/UHF RF signals, *not* low-frequency video, as their name would imply. However, since most manufacturers refer to their products as "video" switches, we'll do so, too.

The switches are sold under various names: *Bambi Electronic Video Switch* from Simple Simon Electronics (3871 S. Valley View No. 12, Las Vegas, NV 89103), *Video Organizer* from Zenith (1000 Milwaukee Ave., Glenview, IL 60025), and many others. All those products have one purpose: to switch in or out a number of program sources with a minimum of loss and crosstalk. They neatly replace the array of 2-position switches that are often used, and eliminate cable swapping. So, once your video system reaches the point where a 2-position switch won't be enough, or where

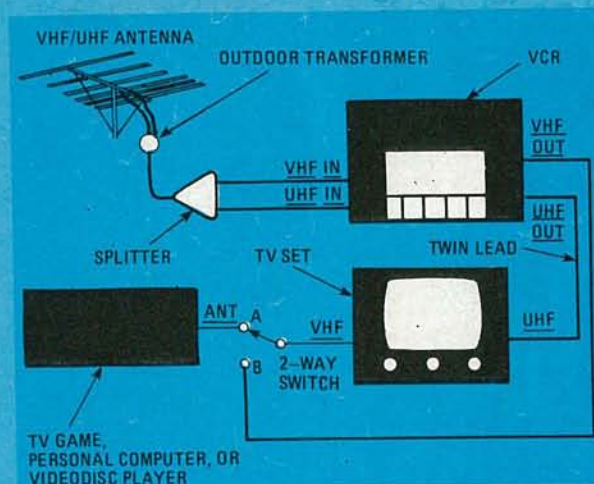


FIG. 4—USE A 2-POSITION switch to select between off-the-air TV reception and using an add-on video device.

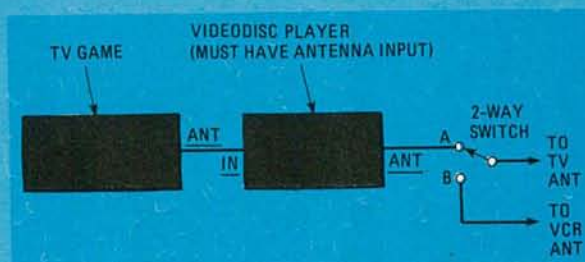


FIG. 5—"DAISY CHAINING," when you can do it, can reduce or eliminate the need for switches.

cable swapping is required, a video switch would be a wise investment.

To really appreciate how neatly a video switch simplifies cabling, suppose you add a video game, personal computer, and a videodisc player to your system. You could make the whole thing work by using the scheme shown in Fig. 4 and using three 2-position switches. Or, you could daisy-chain the devices as shown in Fig. 5. A better way, though, is shown in Fig. 6. There, a single video-switch takes care of everything and fewer cables and connections are required. That means less fuss in hooking the devices up, and less signal loss. But, best of all, using the switch is easy—all that you need to do is press the appropriate button to get what you want.

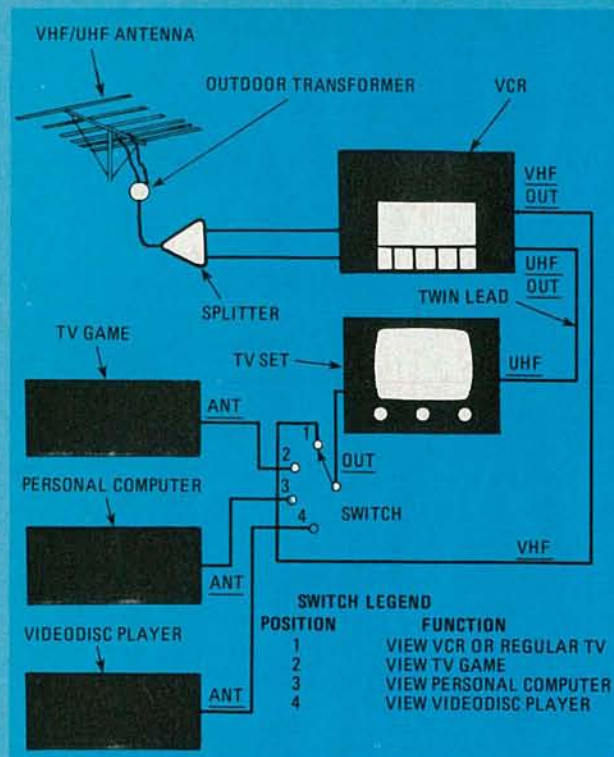


FIG. 6—A MULTI-POSITION switch will allow you to hook up several pieces of equipment to a single receiver easily.

A full-house system

There are many people who take video very seriously, and, as a new item is introduced, they will add it to their systems. With each addition, the switching becomes more complicated, until most video switches become inadequate.

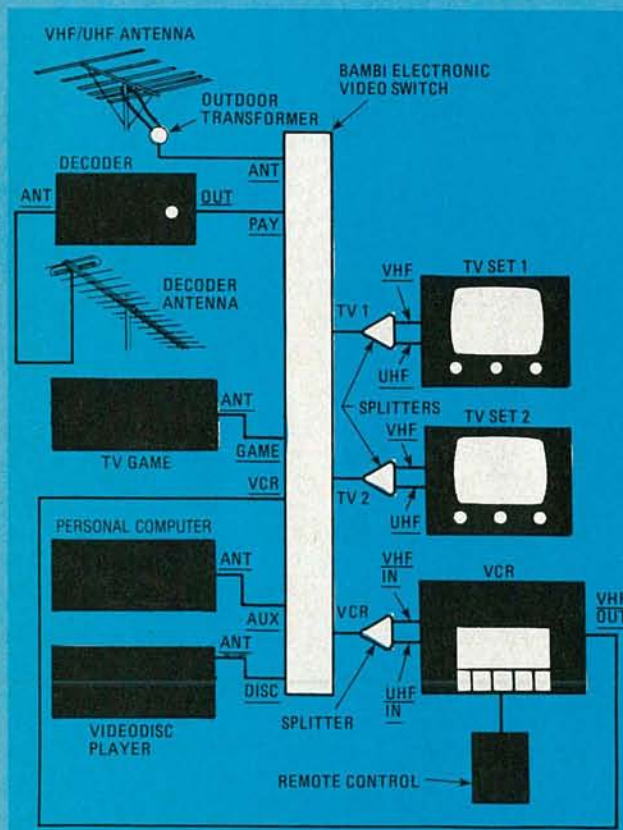


FIG. 7—"FULL HOUSE" VIDEO SYSTEM is easy with an electronic video switch. Any input can feed any output.

The solution to that problem is to select a more versatile video switch.

A setup for those who have "everything" is shown in Fig. 7. That particular arrangement is based upon the *Bambi Electronic Video Switch*. It can handle up to six inputs, and can route the signals from any of those inputs to any of three outputs.

The manufacturer quotes an especially high isolation-figure (65 dB) for the unit, which is important because that reduces interference among sources. Interference can occur because most inputs to a video switch are on the same channel (typically channel 3), and that can cause multiple pictures and interference lines to appear on your TV set(s). The solution is to use a switch with a high isolation-figure, indicated in dB. Watch for that specification when you buy.

As shown in Fig. 7, an outdoor antenna, subscription-TV decoder, VCR, videogame, personal computer and a video-disc player provide inputs to the video switch. Those devices can be switched independently, and the VCR can tape a program while the kids use the TV game and you watch another program off the air. A video switch makes it easy; in the case of *Bambi*, just press two buttons and go!

Selecting a "full-house" video-switch is easy. First, figure out how many devices you will be hooking up to your system. You will need a video switch with *at least* that many inputs, and—I would recommend—a spare or two. If you buy a switch without allowing for future expansion, you'll soon regret it.

There are plenty of video switches on the market, with a wide number of inputs and outputs, so scan the ads when you get ready to buy a switch. When you have narrowed your choice down to a few units, select the one with the greatest isolation figure, if those numbers are available to you. Once you have a video switch, just hook it into your system as shown in Fig. 7.

Copying

There are times when it becomes necessary to copy material recorded on one VCR onto another. For example, I recently replaced my Beta-1 machine with a newer VHS model and had to re-record my old lectures and family outings onto VHS-format tapes as the new machine couldn't handle Beta-format tapes.

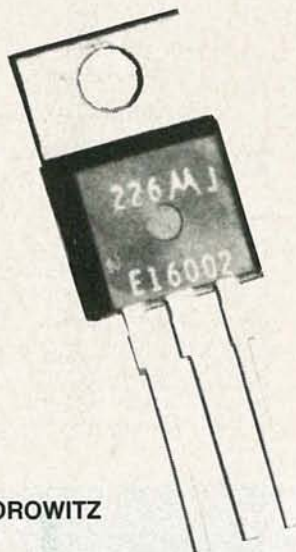
There are two methods that can be used for dubbing videotapes. You could use an RF hookup to the antenna terminals of the second recorder, but that would put a lot of signal-processing circuitry between the two machines and cause a reduction in picture quality. A direct-video/audio hookup using separate audio and video cables is preferred.

You may have to make the cables if you can't buy them locally. The audio cable can be standard audio or microphone cable with RCA phono plugs, or miniature phone plugs on the ends. The video cable should be RG-174 coax with RCA phono plugs on each end. The reason for using coaxial cable is to reduce losses. While a very short length of microphone cable could be used, it is very lossy and will reduce picture definition.

In some cases you may want to make a copy of an old tape that is of poor quality. Sometimes a video enhancer/stabilizer can make improvements in the quality, if the tape isn't too bad.

Enhancer/stabilizers are available from a wide range of suppliers (many of whom advertise in *Radio-Electronics*), and are very useful. The enhancer section of the device boosts—or peaks—the high-frequency portion of the video, which makes the picture look better. The stabilizer portion is used to stop vertical roll when some commercially available tapes are viewed on certain TV sets—especially some of the newer ones. Thus, the enhancer/stabilizer can be a valuable tool when it becomes necessary to copy tapes. **R-E**

How to Design Analog Circuits —Multi-Stage Amplifiers



MANNIE HOROWITZ

For most applications, you'll need more gain than you can get from a one-stage amplifier. This month, we'll show you how to combine one or more amplifier stages to get a more useful circuit.

A SINGLE STAGE OF AMPLIFICATION MAY provide enough gain for some applications, but rarely, if ever, is the gain from a single-stage amplifier sufficient for audio or RF circuits. For those applications, the output from one amplifier stage must be fed to one or more subsequent stages to obtain enough gain. If that is done, the voltage gain of the first stage is multiplied by the voltage gain of the succeeding stages to determine the overall gain of the entire circuit.

Let's look at some of the methods that are often used to connect several amplifier stages together. Transformer coupling was, at one time, used quite often with audio amplifiers—especially in the power-output stages. Although it is still used for that, transformer coupling is now used mainly in IF and RF circuits.

Another method of coupling is through an R-C network. In that type of circuit, the output of one stage is connected to the input of another through a capacitor. In the previous article in this series (see the November 1982 issue of **Radio-Electronics**), we indicated that a capaci-

tor is used to couple a signal from its source to the input of an amplifier to isolate the source from the bias circuit. If the signal source and the bias circuit were not isolated in that fashion, the bias on the transistor would be altered by the signal source. That will also occur in a two-stage (or more) coupled circuit, but with an additional factor—in a circuit of that type, the collector voltage at the first transistor would affect the base current of the second. In turn, the current demanded by the base circuit of the second would affect the collector voltage of the first.

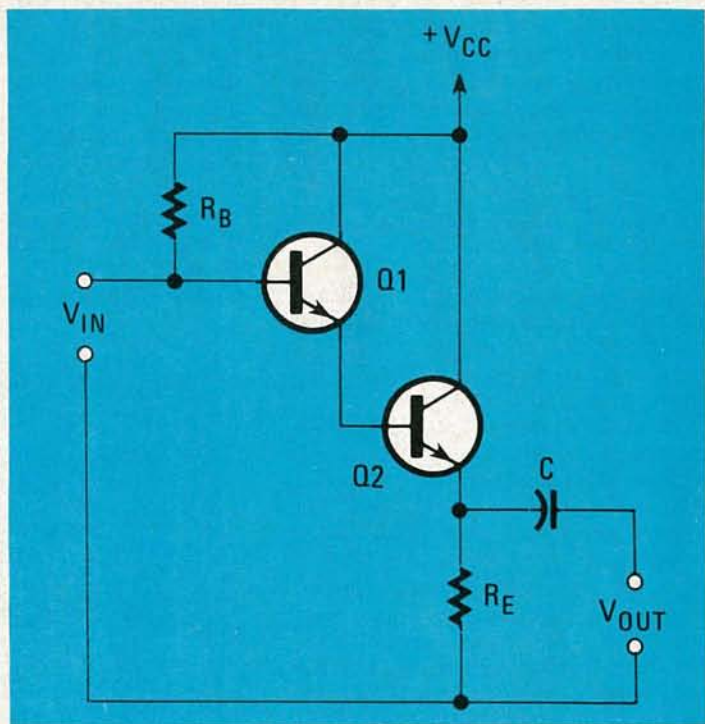
But that interaction can be accounted for and the circuit can be designed so that the use of a capacitor between the two stages is not required. That type of circuit is referred to as a direct-coupled amplifier and there are a number of advantages to using that arrangement, including better low-frequency response.

The circuits we will describe in this article all use bipolar transistors. Many of the problems encountered are not as severe in FET circuits. Thus, once you have mastered the design procedures for

bipolar-transistor circuits, applying what you've learned to FET circuits is relatively easy.

Transformer-coupled circuits

The transformer has characteristics that make it useful as a coupling device. As you know, a simple transformer is made up of two coils of wire wound around a common magnetic-core. For the rest of this discussion, we will assume that one coil, called the primary, consists of N_1 turns of wire, while the other coil, called the secondary, consists of N_2 turns of wire. If a DC current were fed to the primary, a steady magnetic field would be induced in the primary and coupled through the core to the secondary, but no current would be induced in the secondary. If an AC current were fed to the primary, however, an AC current would be induced in the secondary. The signal in the secondary would have the same frequency and vary identically to the signal fed to the primary (assuming an ideal transformer, of course). If there were perfect magnetic coupling between the two



coils, the current appearing in the secondary (I_2) due to the input current (I_1) would be inversely proportional to the ratio of the turns in the coils, or:

$$\frac{I_2}{I_1} = \frac{N_1}{N_2}$$

The ratio of the voltages across the two coils is directly proportional to the turns ratio, or:

$$\frac{V_1}{V_2} = \frac{N_1}{N_2}$$

Since impedance, Z , is equal to the product of V and I , by multiplying equation 1 by equation 2, we come up with the fact that the impedance ratio is proportional to the turns ratio squared, or:

$$\frac{Z_1}{Z_2} = \left(\frac{N_1}{N_2}\right)^2 = \left(\frac{N_1}{N_2}\right)^2$$

That equation states that an impedance in the secondary, Z_2 , will appear reflected into the primary as an impedance, Z_1' , equal to Z_2 multiplied by $(N_1/N_2)^2$, the square of the turns ratio.

Now, let's turn to a practical design example. Let us say that we have a ceramic phonograph cartridge that must see a resistance of at least 22,000 ohms. The average output voltage from the cartridge is 0.25 volt. After amplification, however, 1/2-watt must be delivered to an 8-ohm loudspeaker at the output. Transformer coupling is to be used between the amplifier stages and between the output stage and the loudspeaker. To do that, we'll use the circuit shown in Fig. 1. Assume that the beta of both Q1 and Q2 is 100.

If the transformer is 60% efficient, at least $0.5/0.6 = 0.833$ watt must be available from Q2. However, there are other factors such as transistor leakage, losses due to the emitter resistor, and so on. Considering those potential losses, we should be safe if we design the circuit to deliver 1 watt.

To start our design, we must first draw the load lines for output transistor Q2. When we draw the load lines, we must take into account the effect of the transformer in the collector circuit of Q2. Since the primary of the transformer is an inductor, it will present a DC resistance, equal to the resistance of the wire in the coil, but a different AC impedance. Because of that, two load lines, one DC and one AC, must be plotted on the same graph. The supply voltage is specified to be 9 volts.

Let's assume that the resistance of the primary winding of T2 is very low and just about equal to zero. We will also assume that R_{E2} is very small and let it equal zero. Making those assumptions, the vertical line in Fig. 2 is the DC load

line. (To make the graph clearer, the transistor's characteristic curves have been omitted here.) That is, of course, determined by the equation for the load line, $V_{CC} = I_C(R_P + R_{E2}) + V_{CE}$. Since R_P , the resistance of the primary winding of the transformer, and R_{E2} were both assumed to be zero, $V_{CC} = V_{CE}$ for all values of collector current I_C . That, of course, is the equation of a straight line.

The AC impedance is R_L reflected into the primary of T2 as a resistance R_L' . Before we go any further, let's take another look at Q2's collector circuit because something interesting happens here. First, as we said before, the primary of the transformer is an inductor. Just as a capacitor opposes sudden changes in voltage, an inductor opposes sudden changes in current. We will show how the quiescent collector current, I_{CQ} , is chosen shortly, but for now let's just say that it will be chosen to allow for equal swings between 0 and I_{Cmax} . With no input-signal applied to the transistor's base, I_{CQ} (the quiescent collector current) flows through Q2's collector circuit and through the primary of T2. When an AC

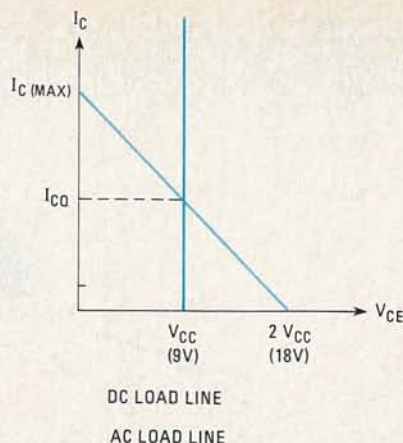


FIG. 2—THE LOAD LINES for the transistors of the circuit shown in Fig. 1 are shown here.

swing between 0 and 18 volts. Plot 18 volts on the V_{CE} axis; that determines V_{Cmax} on the load line.

The amount of power delivered from the transistor, 1 watt, determines the I_{Cmax} point on the load line. Because the signal swings from the transistor's quiescent current point, I_{CQ} , to a peak current of I_{Cmax} , and then to a minimum current

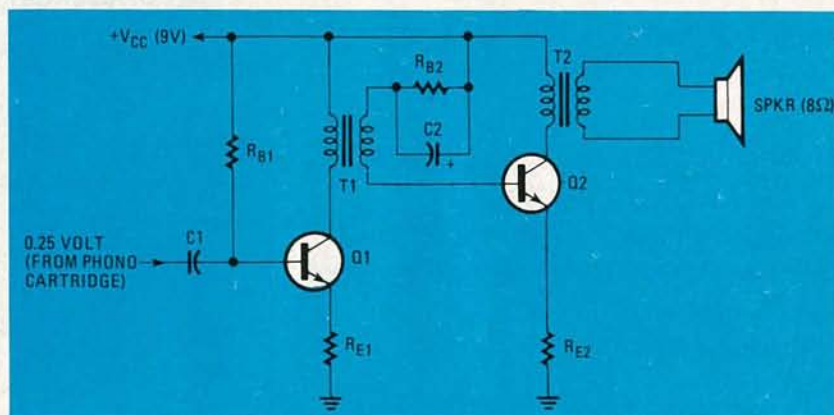


FIG. 1—THIS BASIC TRANSFORMER-COUPLED CIRCUIT uses transformers to couple the two amplifier stages, and to couple the entire circuit to the load, in this case a loudspeaker.

signal is applied to the transistor, however, things begin to change.

Let's first consider what happens when the AC-input swings negative. When that happens, the transistor tries to decrease the collector current by increasing its collector-to-emitter resistance. The inductance of T2's primary, however, opposes the change in I_C and tries to maintain the flow of current at I_{CQ} . The result is that I_{CQ} now flows through a higher collector-to-emitter resistance, and the transistor's collector-to-emitter voltage *increases* to a level higher than the power-supply potential. That same phenomenon occurs when the input signal starts swinging positive. In that case, however, the collector-to-emitter voltage subtracts from the power-supply voltage. As a result, the transistor should be biased so that V_{CEQ} is higher than $V_{CC}/2$. We will design the circuit in Fig. 1 so that the collector-to-emitter voltage of Q2 will

of 0, the RMS current delivered by the transistor is $I_{Cmax}/2\sqrt{2}$. Similarly, the RMS voltage due to the 0- to 18-volt swing is $V_{Cmax}/2\sqrt{2} = 18/2\sqrt{2} = 6.36$ volts. The product of the RMS voltage and the RMS current must be equal to the required 1 watt needed from Q2, so $6.36(I_{Cmax}/2\sqrt{2}) = 1$, and $I_{Cmax} = 0.44$ amps. Plot that point on the I_C axis and draw the AC load line. Idling current, I_{CQ} , is half of I_{Cmax} , or 0.22 amps.

The slope of the AC load line is the negative of the load resistance as seen by Q2, or $(18 - 0)/(0.44 - 0) = 41$ ohms. Because the load that the loudspeaker presents to the secondary is 8 ohms, the impedance ratio of the primary to secondary is $41/8 = 5.13$. The turns ratio of T2 is the square root of the impedance ratio, or 2.3:1.

For stability's sake, we want to make R_{E2} as large as possible. But we are limited because of gain. Let us assume, for

the moment, that we can accept a gain of 8.2. Then the sum of R_{E2} and r_e , the AC emitter resistance of the transistor, can be $41/8.2 = 5$ ohms. (That is so because the gain of a stage of amplification is equal to the ratio of the output impedance to the resistance in the emitter circuit. Here it is $41/5 = 8.2$.) Because the idling current is 0.22 amps, or 220 mA, $r_e = 26/220 = 0.12$ ohms. Resistor R_{E2} can be a maximum of $5 - 0.12 = 4.88$ ohms. Thus, a standard 4.7-ohm resistor can be used for R_{E2} . Now the actual gain of that circuit is $41/(4.7 + 0.12) = 8.33$. That is just a bit above our minimum desired gain of 8.2.

If the transistor is to idle at 0.22 amperes, the base current is $I_{CQ}/\beta = 0.22/100 = 0.0022$ amp. That current is determined from the circuit consisting of V_{CC} , R_{B2} , the resistance of the secondary of T1 (assumed to be 0 ohms), the base-emitter voltage of Q2, and the resistance R_{E2} multiplied by β . Knowing everything but R_{B2} , we can determine it from the equation relating all these factors, $0.0022 = (V_{CC} - V_{BE})/(R_{B2} + \beta R_{E2}) = (9 - 0.6)/[R_{B2} + 100(4.7)]$; solving for R_{B2} , we find it is equal to 3348 ohms. A standard 3300-ohm resistor can be used. If that resistor is only to affect the DC bias, no signal voltage must appear across it. For that to happen, R_{B2} must be bypassed by a capacitor. The reactance of that capacitor, X_C , should be about .01 of the resistance of R_{E2} , or 330 ohms at the lowest frequency, f_L , to be amplified. The value of the required capacitor can be found from $C = 1/(2\pi f_L X_C) = 1/2072.4f_L$.

Because the gain of the circuit is 8.33 and the RMS voltage at the collector of Q2 is 6.3 volts, the voltage at the secondary of T1 must be at least $6.3/8.33$, or 0.76 volts RMS to drive the output circuit to its desired level. The resistance across the secondary winding of T1 is the 3300 ohms of R_{B2} in parallel with βR_{E2} (r_e is negligible), or 3300 ohms in parallel with 470 ohms. That is equal to 411 ohms.

A reasonable load for T1 to present to the collector of Q1 is 2000 ohms. As the secondary of the transformer sees 411 ohms, and as the turns ratio of T1 is equal to the square root of the resistance ratio, that turns ratio is equal to $\sqrt{2000/411}$, or 2.2:1. That, of course, is also the voltage ratio. So if 0.76 volt must be across the secondary of T1 for T2 to deliver $\frac{1}{2}$ watt, $2.2 \times 0.76 = 1.67$ volts must be across the primary of T1. As the cartridge delivers 0.25 volt, the gain of the Q1 amplifier must be at least $1.67/0.25 = 6.68$. Hence $R_{E1} + r_e$ of Q1 must be equal to or less than $2000/6.68 = 300$ ohms. We can now plot a load line for Q1, and find its quiescent operating point, as shown in Fig. 2.

When we worked around Q2, we did not decide just what the AC load resistance should be. Now, we know it—it is 2000 ohms plus the 300-ohm resistance of the emitter, or a total of 2300 ohms. Thus,

I_{Cmax} is equal to 18-volts/2300-ohms, or 7.8 mA. The quiescent operating point, I_{CQ} is one half of that for Q1, or 3.9 mA. That makes r_e for the Q1 stage equal to $26/3.9$, or 6.7 ohms. Resistor R_{E1} must then be a maximum of 300 ohms less 6.7 ohms, or less than 293.3 ohms. Use a standard 270-ohm resistor for that.

Since the idling collector current is 3.9 mA and β is 100, the base current is $3.9/100$, or 0.039 mA. Using that, R_B can be determined from the base current circuit; i.e., $(V_{CC} - V_{BE})/(R_B + \beta R_E) = (9 - 0.6)/[R_B + 100(270)]$. Solving for R_B , it is equal to 188,000 ohms and a standard 180,000-ohm resistor can be used.

Transformer coupling is used quite often in radios where one IF stage is coupled to the other in a circuit similar to that shown in Fig. 1. In those circuits, however, a capacitor shunts the coil windings to form a resonant circuit so that only the IF frequency will pass through the amplifier circuit and all other frequencies will be rejected.

Resistor-capacitor coupling

Transformers are fine when you are working with RF frequencies. But at AF frequencies, transformers can get quite expensive and bulky. That's why R-C coupling is now being used for many applications, although transformer-coupling is still common in RF circuits.

Let's look at the circuit shown in Fig. 3. This time, let us say that the input

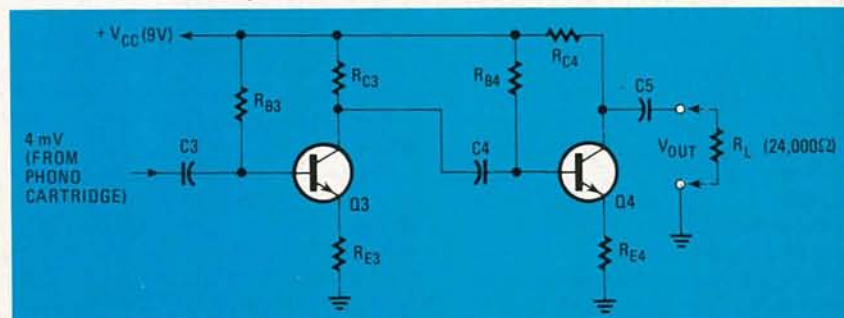


FIG. 3—AMONG THE ADVANTAGES of using R-C coupled circuits such as this one are the elimination of bulky and expensive transformers, and increased voltage gain.

signal comes from a low-impedance magnetic-type phonograph cartridge with an output of 4 millivolts. Assume we want to build that voltage up to a sufficient level to drive the power amplifier of Fig. 1. Thus the circuit in Fig. 1 would be driven by an amplifier circuit rather than by a ceramic cartridge. In this case, 0.25 volt is required at the output of the R-C circuit of Fig. 3 and that output must drive a load equal to the input impedance of the transformer amplifier, or 180,000 ohms in parallel with β multiplied by 276.7 ohms (the emitter resistance of Q1), or 24,000 ohms. The overall voltage gain of the amplifier must then be $0.25\text{-volt}/4\text{-millivolt} = 62.5$. Theoretically, the voltage gains of the two stages can be made identical. If that were done, the gain of each stage would be $\sqrt{62.5}$, or 7.9.

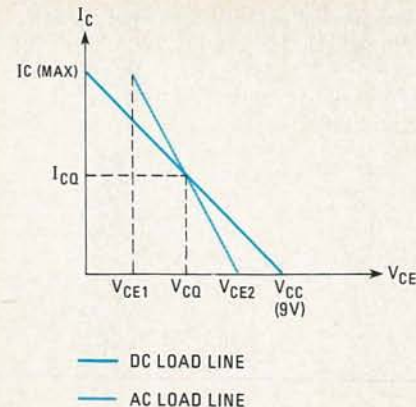


FIG. 4—THE LOAD LINES for the transistors in the circuit shown in Fig. 3.

Usually, however, the voltage gain of the second stage is made lower than that of the first stage because the output load of the second stage generally has a lower impedance.

Let's start the design by considering the second stage. Resistor R_{C4} should be chosen so that its value is between 10% to 20% of R_L ; let's make it 3300 ohms. Now the DC load on Q4 is 3300 ohms and the AC load is 3300 ohms in parallel with the 24,000 ohm load, or 2933 ohms. To have a gain of somewhat more than 5, make R_{E4} equal to about 20% of 2933 ohms, or about 470 ohms.

The DC load line (see Fig. 4) extends from V_{CC} to a point on the I_C axis equal to I_{Cmax} . From the load-line equation, $V_{CC} = I_{Cmax}(R_{C4} + R_{E4}) + V_{CE4}$, we find

that $I_{Cmax} = 2.4$ mA when $V_{CE4} = 0$ and $V_{CC} = 9$. And, as I_C must swing between 2.4 mA and 0, $I_{CQ} = 1.2$ mA and $V_{CQ} = 4.5$ volts.

The AC load line must pass through that point. We can determine V_{CE1} from the negative of the equation for the slope of the AC load line, or $(V_{CE2} - V_{CQ})/(I_{CQ} - 0) = (2933 + 470) = 3403$ ohms. Substituting $V_{CQ} = 4.5$ volts and $I_{CQ} = 1.2$ mA into the equation, $V_{CE2} = 8.6$ volts. As V_{CE1} is the same distance below V_{CQ} as V_{CE2} is above it, and because $(V_{CE2} - V_{CQ}) = (8.6 - 4.5) = 4.1$ volts, then $V_{CE1} = (4.5 - 4.1) = 0.4$ volts when I_C is equal to 2.4 mA.

The output must be able to swing the full 0.25 volt RMS, or $2\sqrt{0.25} = 0.707$ volt peak-to-peak in order to be able to drive the amplifier in Fig. 1 suf-

ficiently. The maximum peak-to-peak swing of the transistor is $(V_{CE2} - V_{CE1}) = (8.6 - 0.4) = 8.2$ volts. Thus a 0.707-volt swing is well within the capabilities of this amplifier stage.

The actual gain of this stage can be determined after we find just what r_{e4} is in the Q4 circuit. It is equal to $26/1.2 = 21.7$ ohms. Thus, the total resistance in the emitter circuit is $470 + 21.7 = 491.7$ ohms, and the gain is just about equal to $2933/491.7 = 6$. Because the total gain must be at least equal to 62.5, the gain of the first stage, Q3, must be greater than 62.5/6, or 10.42.

We will now determine R_{B4} in a manner similar to that used for determining R_{B2} in the circuit of Fig. 1. Use the equation $V_{CC} = I_B R_{B4} + V_{BE} + \beta I_B R_{E4}$. Note that $\beta = 100$ and $I_B = I_{CQ}/\beta = 1.2/100 = 1.2 \times 10^{-5}$ amps. After substituting into the equation, we can calculate that R_{B4} is equal to 653,000 ohms and a standard 620,000-ohm resistor can be used.

The load on the output of the Q3 circuit is 620,000 ohms in parallel with $\beta(R_{E4} + r_e)$, or 45,291 ohms. As R_{C3} should be less than 10% of that, use a 4300-ohm resistor for that. Now the DC load on the transistor is 4300 ohms and the AC load is that resistor in parallel with 45,291 ohms, or 3927 ohms.

Following the procedure used to find R_{E4} , we can determine that R_{E3} should be 330 ohms if we are to get a gain of more than the required minimum of 10.42. To check the gain, we must first determine I_{Cmax} . It is $(9 - 0.6)/(3927 + 330) = 1.97$ mA, and I_{CQ} is one-half of I_{Cmax} , or about 1 mA. That makes r_e equal to 26/1, or 26 ohms. The total emitter resistance is $330 + 26$, or 356 ohms. Consequently, the actual gain of the Q3 circuit is about $3927/356$, or 11. As that is more than the required minimum of 10.42, the circuit fulfills our requirements.

The value of R_{B3} can be found from the equation $V_{CC} = I_B R_{B3} + V_{BE} + \beta I_B R_{E3}$. When we note that $I_B = \beta I_{CQ}$, we can determine that R_{B3} is equal to about 897,000 ohms, and you can use a standard 910,000-ohm resistor.

The resistance the magnetic cartridge sees is 35,600 ohms (the resistor in the emitter circuit of Q3, multiplied by the beta of that transistor) in parallel with 910,000 ohms, or just about 35,000 ohms. That is pretty close to the resistance most cartridges must see. If the resistance the cartridge must see is specified to be less than 35,000 ohms by the manufacturer, an additional shunting resistor must be wired across the cartridge. The shunting resistor can be calculated by using the equation for resistors in parallel and substituting 35,000 ohms and the resistance that the cartridge must see into the equation. Place that calculated resistor directly across the cartridge, not on the circuit side of C3. The reason it must be placed that way is to keep it from upset-

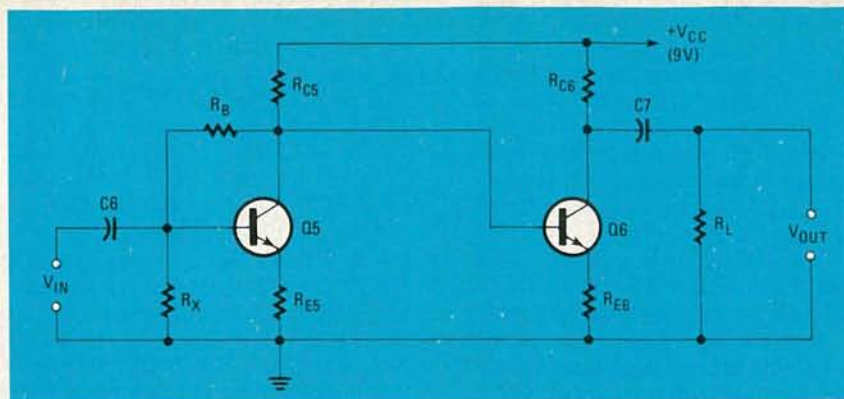


FIG. 5—DIRECT-COUPLED CIRCUIT. Those circuits have been developed to overcome the drawbacks of using capacitors as coupling devices.

ting the bias voltage established by the other components in the circuit. The function of the isolating capacitor has been detailed in the previous article on audio-amplifier circuits.

Direct-coupled amplifiers

Using capacitors to couple transistor amplifiers does have some important drawbacks. For one thing, capacitors do not pass low frequencies very readily. Besides, capacitor-resistor circuits introduce phase shifts, are not linear, and add distortion. As a result, circuits have been developed that do not use coupling capacitors. An example of one such circuit is shown in Fig. 5. While that circuit does not appear to be much different than the one you see in Fig. 3, the differences that do exist are significant. Note primarily that the coupling capacitor between the collector of Q5 and the base of Q6 has been omitted. The overall gain is still the product of the gains of the two transistor circuits.

The circuit around Q6 can be designed in a fashion similar to that used for the design of the circuit around Q4 in Fig. 3. But here, we must also consider that the base voltage of Q6 is the same as the collector voltage of Q5. Voltage at the collector of Q5 must not be upset if Q6's base voltage is low, as it tends to be. To compensate for that, a parallel R-C network may be wired in series with R_{E6} . Considering the DC current that flows through the series resistor circuit, the DC voltage developed across the combination is much greater than it would have been had only R_{E6} been in the circuit. Voltage at the base of Q6 is the voltage across the series combination of resistors added to the V_{BE} of the transistor. The capacitor across the added series resistor is there to bypass the AC signal so that the new resistor will not affect the AC gain of Q6. The reactance of that bypass capacitor should be less than 10% of the value of the resistor it is bypassing at the lowest frequency to be amplified.

Note the R_B and R_X resistors in the base circuit of Q5. If you look back at the bias circuits described in a previous article, you will find that the circuit involving

those resistors is essential for stabilizing transistor collector-current to withstand variations of temperature. It is used here because the effect of any increase in leakage current in Q5 is multiplied in magnitude due to the DC amplification of the direct-coupled circuit. So if the leakage current in Q5 is multiplied in that transistor by a factor of 10 due to the increase in temperature, it can be multiplied further by the beta of Q6 before it appears in its collector circuit. Excellent stability is an important consideration when designing direct-coupled amplifier circuits.

More than just the collector current of Q5 flows through R_{C5} . When determining the voltage drop across that resistor, be sure to add other currents that flow through that resistor to the I_C normally expected. Bias current for Q5 flows through R_{C5} (and R_B) along with the base bias-current for Q6. The sum of those currents may, at times, be a substantial factor in determining the quiescent conditions around the circuit, along with the expected I_C .

Special direct-coupled circuits

Last time, we described the emitter-follower or common-collector circuit. Its characteristics were a high input impedance and a low output impedance; its power gain was equal to beta. The Darlington circuit goes one step farther and improves even on those excellent characteristics by a factor of beta.

Darlington amplifiers consist of two cascaded emitter followers, as shown in Fig. 6. If the beta of Q1 is β_1 , and the beta of Q2 is β_2 , the current and power gains of the overall circuit are $\beta_1 \times \beta_2$. But the voltage gain still remains slightly less than 1, as it did for the emitter follower.

The input impedance is R_E multiplied by the product of the betas, in parallel with R_B . The value of R_B can be quite large in that circuit as there is only a small amount of base current in Q1. That Q1 base current must equal $(V_{CC} - 2V_{BE})/(\beta_1 \beta_2 R_E + R_B)$ for a substantial amount of emitter current to flow in Q2.

The output impedance of that circuit is $R_B/\beta_1 \beta_2$ in parallel with R_E . The values

continued on page 102

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COMPUTER CORNER

Selecting an accounting package

LES SPINDLE*

OFFICE AUTOMATION PERFORMS A VARIETY of useful functions that can greatly increase the cost-effectiveness of each employee. In previous columns, we have examined such sophisticated business-software innovations as database management systems (DBMS) and financial-planning programs (like *VisiCalc*). Those state-of-the-art programs offer creative solutions to everyday office problems. Perhaps less ostentatious—but no less useful—are the software packages available for everyday bookkeeping tasks, one of which is shown in Fig. 1.



FIG. 1

General ledger, accounts payable, accounts receivable, inventory, and payroll form what are commonly referred to as the "Big 5" accounting programs. The bookkeeper who works with quality programs in each of those categories—or opts for a complete interactive "Big 5" package—will benefit from greatly increased productivity from his accounting staff.

Careful planning and selective shopping—as always—are the keys to finding the software that will be best suited to your purposes. It is important to select software that offers enough features and options to make the conversion of your bookkeeping records relatively simple. At the same time, it is wise to select packages that allow enough flexibility for you to fine-tune the program to your own special requirements. Let's take a look at a few criteria to keep in mind when choosing various accounting programs for your office.

In planning to computerize your general-ledger system, begin by defining your needs. Can that system operate in-

dependently, or is it preferable to have one that interacts with other "Big-5" programs? Will it be restricted to the financial activities of one office, or must the reports from various branch offices or subsidiaries be integrated as well? What is the volume of transactions? It is important that your computer system have the memory capacity to support the necessary functions. How educated is your office staff in computer functions? That is important in choosing between a general-ledger package that uses a step-by-step format to guide the novice user through the program, or a more sophisticated offering.

It is important to chart out all of your requirements before even beginning to shop. The salesman demonstrating the software will only be able to direct you to the package best suited to your purposes if he has a good idea of the quantity of data that has to be processed and the specific requirements of your bookkeeping methods.

A general ledger is essentially a financial history of a specific accounting period that lets management analyze a company's profile to determine appropriate planning strategies. The primary functions are to keep a record of financial transactions and the resulting balances, and to generate regular balance-sheets and profit/loss statements.

It is vital for a general-ledger program to use a double-entry accounting system. (That means that for every dollar of debit, there is a corresponding credit entry of equal amount, and vice versa.) Double-entry bookkeeping is a safeguard to insure that all accounts are in balance.

Another point to look for is the system's ability to set up a proper chart of accounts. That is a crucial part of the general-ledger framework, for it can provide rapid departmental profiles, as well as a perspective on the financial status of the entire company. Since the allowable number of accounts varies from package to package, the prospective buyer should select a package that can handle the current number of accounts, and is flexible enough to expand as the company does.

Accounts payable/receivable

An accounts-payable or -receivable package has five primary functions. It must: create master files for customers

and vendors; record sales and purchases; index all sales/purchase data with the appropriate master files; manipulate the data, and prepare summary reports by master file account, date, or some other reference.

As with general-ledger programs, your particular needs will determine the appropriate checklist of required features that you should prepare.

One of the important criteria is the ability to produce audit trails—especially for companies that require a certified audit. A good program will automatically prepare a transaction listing immediately after the data is input.

Most accounts-payable systems will require more detailed record-keeping. For example, a check may have to be coded to multiple accounts (for example, when a single check pays a bill that covers different accounts).

Job-costing is another important feature, as many manufacturers will need to access accumulated costs by job. An accounts-payable system set up to handle job-costing can accept job codes as well as account numbers.

Among other items to keep in mind are: commission accounting (some companies don't pay commissions to the salesman until after the company is paid) and finance charges. The better accounting packages are able to compute and assign finance charges rapidly and efficiently. Most accounts-receivable systems will print statements and invoices, and good ones will automatically print checks—both functions are useful and time-saving for the accountant.

Inventory control

There are several benefits in managing the flow of merchandise to be gained from an inventory-control program. Overstocks can be reduced and out-of-stock situations can be avoided. The average small business pays 25-50% of the value of its inventory just in inventory carrying-costs.

The two most common approaches to automating inventory control are invoicing and order entry. With the invoicing approach, the program won't accept invoicing data until after the order is filled—generating an invoice only upon completion of the order. That system has the drawback of not being able to list

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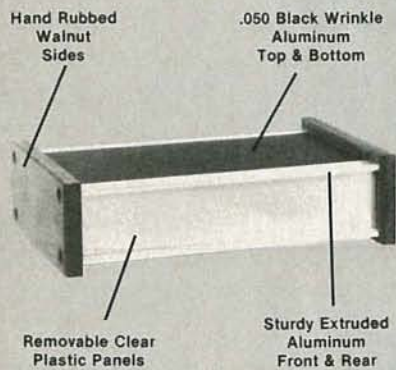
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backorders and open-orders.

An order-entry system will accept the data at any time, separating order-placement and invoicing into two different steps. Also, open-orders may be altered in any way called for, prior to billing. Most businesses will benefit from that approach, as opposed to the more limiting invoicing setup.

An obvious, but critical, consideration in the selection process is knowing the system's ability to handle the current volume of items—as well as knowing its capability for coping with future growth. That often involves hardware decisions such as whether your disk capacity is sufficient to handle the data involved. Most inventory packages on the market will handle at least 1000 items of stock, which should be adequate for most small businesses.

The allowable size of data fields is also very important. Basic specifications for items in stock must be supplemented with other information. Make sure that the package you select allows for that. Such data as product class or code, average cost, previous cost, and product description are often required, so your data fields must be structured with enough space to enter the necessary information.

Payroll

A payroll system is set up primarily to compute and print payroll checks, distrib-

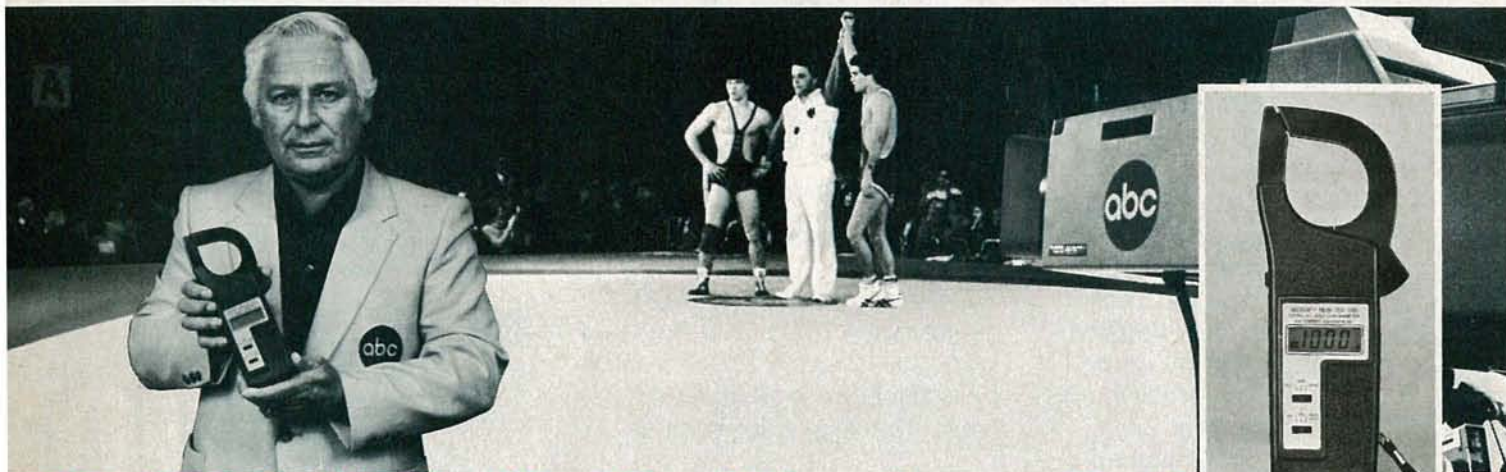
ute labor costs to the proper files, and compile the information required for government reports.

The most common features one can expect in a package are: computing FICA, state unemployment, and federal and state taxes; reporting of employee earnings by category (such as department); generating quarterly and annual government tax filings; and handling of miscellaneous items (non-taxable items, overtime pay, savings and dental plans.)

There are a myriad of other features to be found in various accounting packages. Among them are: reports on hours worked as well as dollar amounts by department and job; union reports; earnings history for each employee; employee files (hire, review, and termination dates), and data on earned-income credits.

More than any other accounting program, a good payroll program can save weeks of employee labor per year. But because of the confidentiality, timeliness, and accuracy required, choosing the right one is an especially crucial step in putting together a computerized bookkeeping system.

Sample the brochures from many different vendors and ask as many questions as you have to satisfy yourself before deciding which package has the features that will work best in your accounting organization. **R-E**



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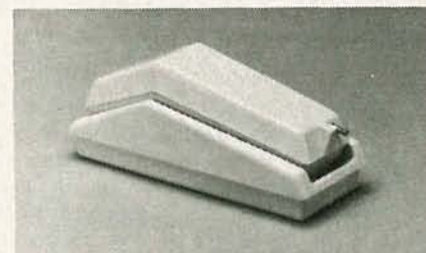
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NEW IDEAS

Low-battery indicator

TODAY MANY HOBBYISTS BUILD BATTERY-operated projects using highly efficient solid-state devices that ensure long battery-life. Even with those circuits, though, a need occasionally arises to make certain that the battery—which may have been in use for some time—is still in good condition. I use the low-battery indicator shown in Fig. 1 in several battery-powered special-effects devices for the theater, where it is crucial that everything operate when it's supposed to.

is selected by the potentiometer; the device can be adjusted most easily by applying to it the voltage at which you want the LED to turn on and adjusting the potentiometer until it just does so.

The indicator uses only six parts: R1 is 27K; R2 is a 100K linear potentiometer; R3 is 1K, and Zener diode D1 is rated at 6.2 volts; IC1 is a 741, and just about any LED will work. The device is easy to build and doesn't present much of a load to the battery it monitors—the version I use draws only

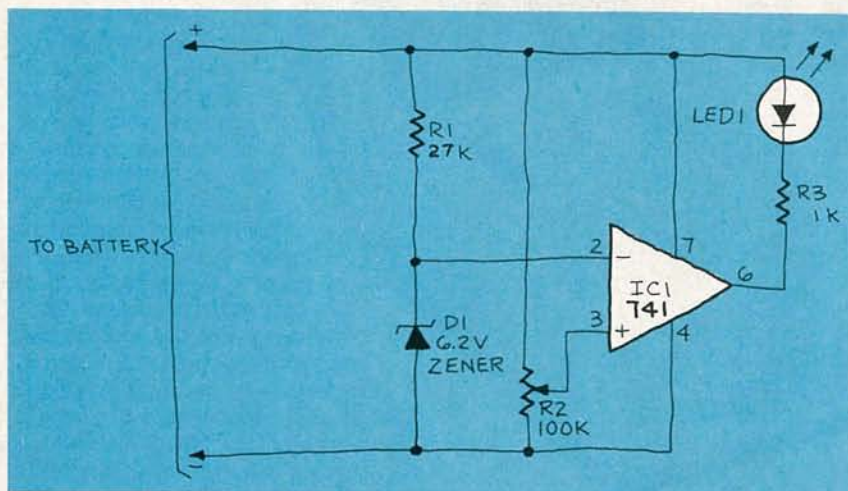


FIG. 1

The low-battery indicator uses an LED to signal when the battery voltage has dropped below a pre-selected level. It is easy to build, reliable, and inexpensive, and can be adapted for a wide range of voltages. The device shown is intended to operate in a 9-volt circuit.

The sensing circuit consists of a 741 op-amp set up as a voltage comparator, using a Zener diode as a voltage reference. The op-amp is inserted as a bridge between two resistance ladders, one containing the Zener reference, and the other a high-value linear potentiometer. The Zener is connected to the inverting input of the op-amp, and the wiper of the potentiometer is connected to the non-inverting input. The top and bottom of the bridge are connected to V_{CC} and ground, respectively.

When the voltage at the wiper of the potentiometer drops below the voltage set by the Zener, the output of the op-amp goes low; that turns on the LED connected between it and V_{CC} . The LED turn-on voltage

about one milliamp when idling, and about 20 milliamps when the LED is lit.

The circuit can be adapted to work with battery-powered circuits requiring between 6 and 18 volts; the only changes needed would be a lower-voltage Zener and smaller current-limiting resistor in the case of voltages below nine volts, and a larger resistor for higher voltages.—Donald F. Ricklies



"But, Mom, I liked the green faces—they were scary."

NEW IDEAS

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TV/VIDEO SYNC, Primer and Product Note Opt. 005-1 is a 16-page primer, with 13 drawings and 3 photographs; the special sections include the United States NTSC broadcast standards and nomenclature, a discussion of PAL and SECAM systems of color transmissions, and a glossary of TV and video terms.

Other sections cover theory of operation, typical setup, specifications, and the features and applications possible when a TV/video

sync option is added to any currently-manufactured HP 1700-series oscilloscope. Applications described are relevant to such industries as medical imaging, consumer video products, and television broadcasting.

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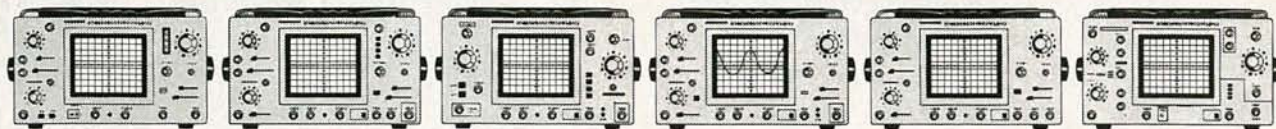
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Measuring relative humidity

ROBERT F. SCOTT, SEMICONDUCTOR EDITOR

A KNOWLEDGE OF THE RELATIVE HUMIDITY at any given time can permit us, or others around us, to make decisions that can affect our health, comfort and personal safety. In addition, the measurement of relative humidity (abbreviated RH) is vital in such areas as food processing, air conditioning, packing, photography, paper and lumber production, and chemical manufacturing. Knowing the RH—along with the temperature—can allow airport operators to predict fog and runway icing. Similarly, farmers and nurserymen can predict dew and frost; and highway safety authorities can forecast dangerous fog and icing on bridges.

Electronic circuits for humidity measurement have rarely appeared in the press because humidity-sensitive transducers have been expensive and seldom available to the electronics constructor and hobbyist. Now, for about

ten-dollars-worth of semiconductors and discrete components, and less than sixty dollars for the humidity sensor, you can build a direct-reading electronic hygrometer. The device, described in National Semiconductor's Application Note AP 256, is designed around the PCRC-55 humidity sensor, from Phys-Chemical Research, and a linear 0 to 10-volt DC meter.

The sensor

The electro-humidity sensor has a hygrometric element whose impedance changes with changes in relative humidity. The hygrometric portion of the sensor is on the surface of a chemically treated styrene copolymer plastic wafer. Water vapor is sorbed or desorbed by means of adsorption (the adhesion of molecules to a surface; not the same as absorption). That

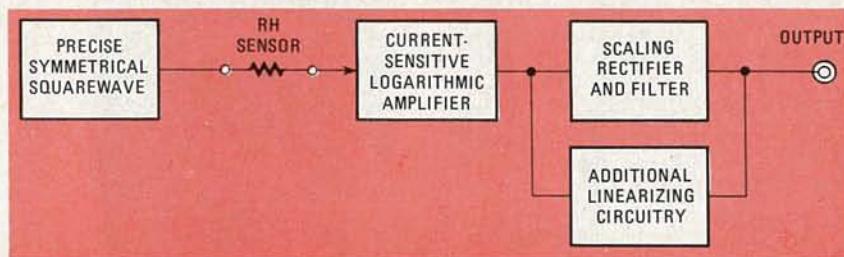


FIG. 1

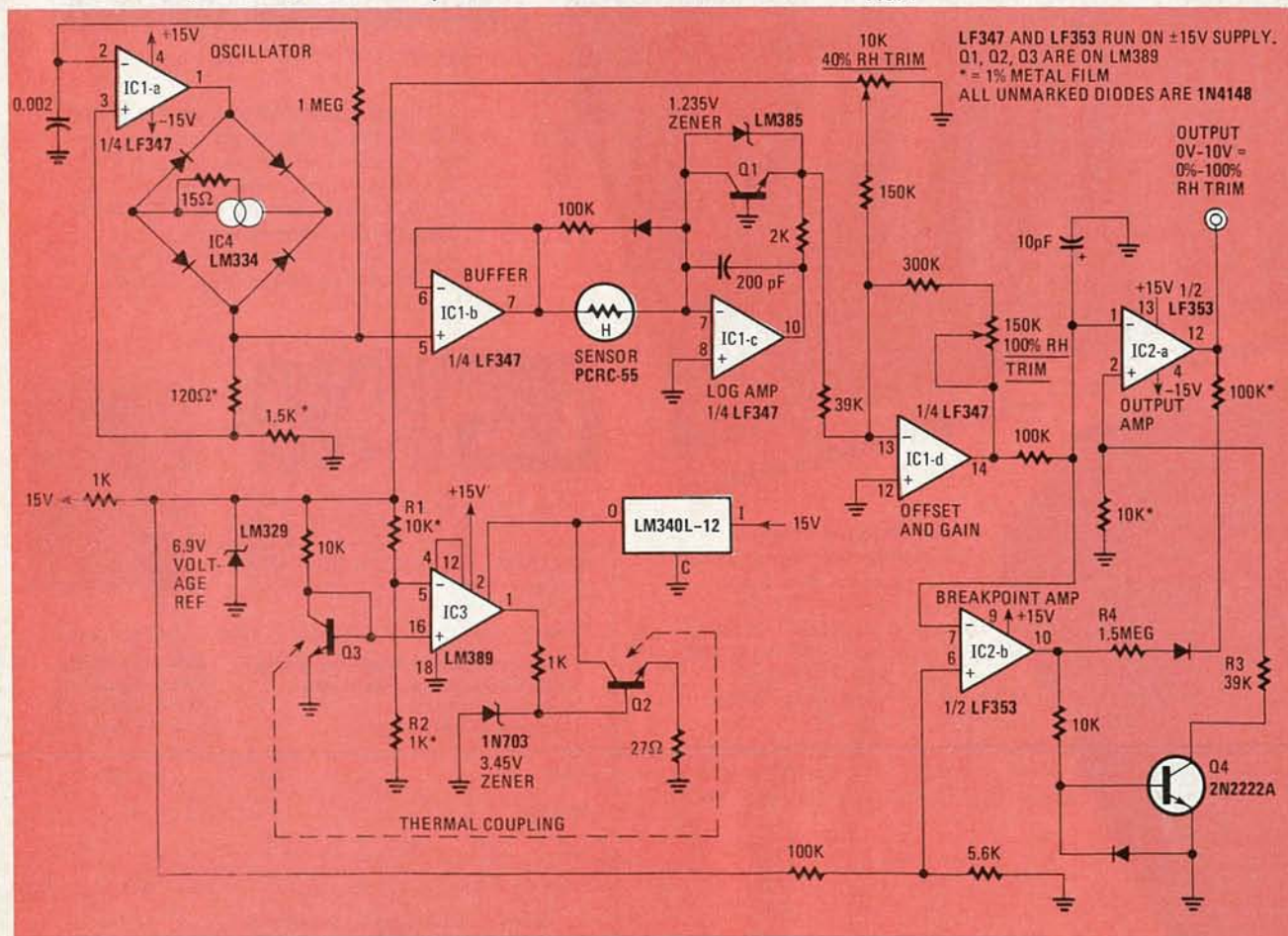


FIG. 2

results in the sensor's having a very rapid response to changes in relative humidity. A single sensor covers the complete range of relative humidity—from 0 to 100%.

The sensor should be excited only by AC voltages (preferably sinusoidal waveforms) of at least 20 Hz with no DC component. Sustained operation with a DC voltage, or AC with a DC component, causes a shift in calibration. The maximum allowable current is one mA.

Figure 1 is a block diagram of the electronic hygrometer. An amplitude-stabilized squarewave drives a precise alternating current through the sensor. The sensor's output current feeds a current-sensitive logarithmic amplifier to linearize the response. The output of the log amplifier is scaled, rectified, and filtered to provide a 0-10-volt DC-output representing RH's from 0 to 100%.

A practical circuit

In Fig. 2, a symmetrical squarewave is generated by IC1-a, an op-amp with positive feedback applied to cause it to oscillate. The combination of constant-current source IC4 and its associated diode-bridge clamps the squarewave output of IC1-a at +8 volts. IC4 has a positive temperature coefficient of 0.033%/°C, which almost completely compensates for the negative 0.036%/°C temperature coefficient of the PCRC-55 sensor. (Mount IC4 close to the sensor so they

will be at the same ambient temperature; that way the temperature coefficient of the complete instrument will approach zero.)

The squarewave is buffered, and then fed through the sensor into the summing junction of IC1-c. On negative-going halves of the squarewave, the V_{BE}/I_C characteristic of Q1, in the IC1-c feedback loop, gives the amplifier a logarithmic amplitude-response. On positive-going half-cycles of the squarewave, feedback through the diode to the summing junction ensures that this point remains at a virtual ground so the sensor always "sees" the required symmetrical drive-waveform.

The output of the log amplifier goes to IC1-d, an op-amp used to sum-in the calibration at the 40% RH point, and to provide adjustable gain for trimming the output to the proper level for a 100% RH reading. The output of IC1-d is filtered to DC by the 100K resistor and 10 μ F capacitor and then fed to IC2-a, the output amplifier.

IC2-b compensates for the sensor's departure from linear response below 40% RH. It does that by altering the gain of the output amplifier when its input drops to about 0.36 volt—corresponding to 40% RH. The inverting input of IC2-b, the breakpoint amplifier, is tied to the non-inverting input of the output amplifier. When the input to those paralleled gates

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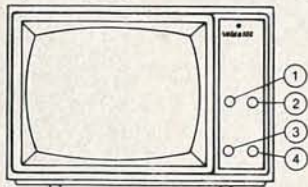
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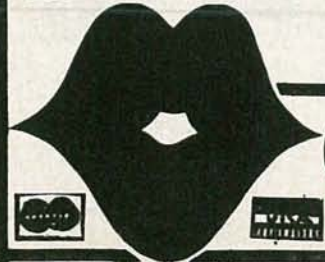
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drops below 0.36 volt, the output of IC2-b swings positive; that turns on transistor Q4 to produce the required change in the output amplifier. Transistor Q4 is turned off for RH values above 40%, so linearity is then determined solely by the log amplifier.

Transistor Q1, in the log-amplifier feedback loop, is extremely temperature sensitive and can adversely affect the performance of the amplifier. To compensate for that, the designers at National Semiconductor came up with a unique circuit application. Transistors Q1, Q2, and Q3 are discrete NPN devices on the same substrate as the audio power-amplifier in IC3, an LM389. Transistor Q3 serves as the on-chip temperature sensor while Q2 is used as the on-chip heater. The audio amplifier senses the temperature-dependent V_{BE} of Q3 and uses it to drive Q2 and heat the chip to a temperature (typically 50°C) set by the reference voltage at the junction of R1 and R2. That circuit stabilizes Q1's temperature and makes it immune to changes in ambient temperature.

To adjust the temperature-stabilizing circuit, ground Q2's base, apply power and then measure Q3's collector voltage. Make a note of the ambient room temperature. Now, calculate what Q3's collector voltage will be at 50°C; allowing $-2.2\text{mV}/^\circ\text{C}$. Adjust the value of R2 to develop a voltage close to the calculated

UNDERSTANDING ATMOSPHERIC MOISTURE

Water covers over three-fourths of the Earth's surface, and there is always some moisture in the Earth's atmosphere—even over the driest desert. At times, as much as 4–5% of a portion of the atmosphere is water vapor or water in a gaseous state. The variation of the amount of moisture in the air (*humidity*) is influenced by geographic location, temperature, and wind currents. The most important of those is *temperature*. Heat causes some of the molecules of the Earth's surface water to escape into the atmosphere.

Saturation occurs when, at some temperature, the rates of evaporation and condensation balance. The space above the liquid contains all the water vapor it can hold. When that happens, the air is *saturated*. The amount of vapor a volume of air can hold before becoming saturated depends on the temperature; the higher the temperature, the more moisture the air can hold. Air can hold four times more moisture at 70°F than it can at 32°F.

Absolute humidity is a measure of the actual amount of moisture in the air at a given temperature. It is expressed either as the number of *grains* (one seven-thousandth of a pound) per cubic foot of air, or in terms of pressure in millibars or in inches of mercury.

Relative humidity is the moisture content of the air *expressed as a percentage*. It is the ratio of the absolute humidity to the greatest amount of moisture the air is capable of holding at the same temperature. As an analogy, consider a quart bottle that is half full of liquid. The ratio of its actual contents (one pint) to its capacity (one quart) is 50%. Similarly, a mass of air that is holding half the moisture possible at a given temperature has a relative humidity (RH) of 50%.

When the relative humidity is high (60% to 85%) we say the weather is muggy. Evaporation of perspiration is slow and we feel overheated. We are uncomfortable as it accumulates on our skin and clothing.

When the relative humidity is too low, perspiration evaporates too rapidly and we are uncomfortably chilly. Prolonged exposure causes our throat and nasal passages to become dry. When it comes to personal comfort, temperature and relative humidity are closely related. We enjoy relative humidities in the range of 45% to 55%.

Indoors in the winter when relative humidity is low, the comfort range is around 68°–74°F. Warm-air heating systems dry out the air and we often use humidifiers to raise the moisture content to a comfortable level. On a muggy summer day, we may use a dehumidifier to make indoor air drier. An air conditioner dehumidifies as it cools.

Dew point is the temperature at which air, as it cools, becomes saturated (RH is 100%) and water droplets condense on cool surfaces such as grass and plants. If the dew point is above 32°F, the condensate is dew. When the dew point is 32°F or lower, the condensate forms ice crystals that we call frost. **R-E**

50°C potential at IC3's inverting (–) input. The range of values for R2 can be fairly wide because the exact chip temperature is not important as long as it is stable. Finally, remove the ground from Q2's base and the chip should reach the predetermined stable operating temperature within 100 ms. You can check temperature stability by reading Q3's collector voltage while blowing on IC3. The measured voltage should remain constant within 100 μV (0.05°C).

Calibration

To calibrate the electronic hygrometer, connect a 35K resistor in place of the sensor and adjust the 150K 100% RH TRIM pot for 10 volts output. Now, substitute an 80-megohm resistor for the sensor and set the 40% RH TRIM pot for 4 volts at the output terminal. Repeat the 100% RH TRIM and 40% RH TRIM adjustments until they no longer interact with each other. Finally, substitute a 60-megohm resistor for the sensor and select a 39K resistor for R3; that will develop a 2.4-volt output corresponding to 24% RH. It may be necessary to select a particular 1.5-megohm resistor for R4 to minimize jitter in the meter reading around the 24% RH point.

As this is written, the sensor is priced at \$57.00, \$54.00, and \$51.50 each, in lots of 1 to 24, for devices with RH deviations of $\pm 1\%$, $\pm 1.5\%$, and $\pm 3\%$, respective-

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ly. To obtain the latest pricing and data on the humidity sensor, write to Phys-Chemical Research Corp., 36 W. 20th St., New York, NY 10011. Be sure to specify the PCRC-555 because the company has other sensors whose characteristics are not suitable for the circuit that has been described here.

Semiconductor catalog

SGS Shortform '82 is a 72-page catalog listing pertinent application information and technical data on the SGS line of small-signal and power transistors; and linear, low-power Schottky TTL, MOS, and CMOS 4000/4500 logic-series IC's. Also included in a list of SGS data books, technical notes, design notes, and the names and addresses of distributors and reps throughout the U.S.—**SGS Semiconductor Corp.**, 7070 E. 3rd Ave., Scottsdale, AZ 85251.

Instrumentation Amplifier IC

The standard three-op-amp instrumentation amplifier with its more than a half-dozen discrete external components has been replaced by National's LM363. A low-cost, high-performance monolithic instrumentation amplifier, the LM363 (and the LM163 military version) comes in an 8-pin TO-5 package. It requires no external resistors for accurate fixed gains of either 10, 100, or 500 (one gain per package). The device has a

super-beta bipolar input stage that yields a very low input-voltage noise, and a high common-mode rejection ratio (CMRR). For added versatility, the internally set gains can be increased by adding external resistors. A 16-pin DIP-packaged device, which should soon be available, will feature pin-strappable gain options of 10, 100, and 1000. The LM363 is priced as low as \$9.60 in 100-piece lots.—**National Semiconductor**, PO Box 70818, Sunnyvale, CA 94086.

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SERVICE CLINIC

A common cause JACK DARR, SERVICE EDITOR

THIS MONTH'S TITLE HAS NOTHING TO DO with any political organization. It refers to a problem in TV sets that can be puzzling if you don't know about it. If you've been with me for a long time, you'll soon realize that I've covered this ground before. However, I've noted of late that I'm getting quite a few letters in the mailbag concerning what is obviously this particular problem so I thought it would be a good idea to discuss it once again.

The problem in question has one characteristic symptom—the set suddenly develops *many* symptoms at the same time! It may have poor sync, weak video, poor- or off-color, and so on. Common sense tells us that it's unlikely for many parts or circuits to all go bad at the same time. (Unlikely, but definitely not impossible!) So, what we should look for is a common cause—something that's common to all stages (and that's where the title comes from, at least in part).

In every TV set ever made, there's one circuit that fits that description—the DC power-supply. The DC supply has two functions; one, of course, is to supply the operating voltages to all the stages of the set. The other is the cause of the problem. The B+ supply also serves as the *common return* path for signals from every stage. Figure 1 will give you an idea of what I mean. The large filter-capacitors in the B+ supply provide a very-low-impedance path to ground or common (there's the rest of the title) for AC only; the impedance is a decimal point followed by about ten zeros and a 1.

If one of the filter capacitors develops a defect such as high power factor, or even opens, that raises the impedance in the common return-path and all the signals returning to ground develop signal voltages across that impedance. Those signals float around the B+ network, and

“back-up” into the other stages, where they immediately cause trouble by beating with the normal signal. That is plain old-fashioned feedback, and you know what that causes—oscillation, regeneration, beats, and other various and sundry problems.

Fortunately, there's a simple way to determine whether a bad B+ capacitor is at the root of your problem(s). Just touch a scope probe to a DC-voltage supply point on any of the circuits that are acting up. The normal response should be absolutely nothing; just a nice straight green line, even with a high vertical-gain. (Incidentally, that is the best possible test for either filter or bypass capacitors.) If you see any sign of a signal on the bypassed end of a circuit, you know that the filter or bypass capacitor is bad.

Try bridging a good capacitor across the suspect one. If the signals disappear and the problem clears up, you've found the cause (and cure). Replace the original capacitor with a good one.

Caution: perform this test with the power on *only in tube sets*. In all solid-state sets, turn the power off, connect the bridging capacitor, and then turn the set back on. The charging surge causes transients that can damage transistors, IC's, and other components.

That “common” problem causes some of the oddball symptoms we've all seen—the dark horizontal bar crawling up through the picture, bending or weaving at the sides of the raster, beats in the picture or color, AGC problems, etc.

One of the more unusual situations I ran across was a set with absolutely no vertical sync; the horizontal sync was steady as a rock. Scoping the composite-sync waveform at the output of the sync separator showed what looked like a normal pattern. A normal pattern will show

the vertical sync as thin lines every so often along the “bar” made up of the horizontal-sync pulses. I finally noticed that they weren't there (I hadn't looked close enough at first)! When I scoped the DC power-supply for the stage, I found that there was a small sharp spike at the vertical frequency.

That spike was feeding back into the sync-separator output, opposite in polarity to the sync pulses, and neatly punching out the vertical sync! The cause turned out to be a bad electrolytic capacitor in DC power supply.

So, there you are. If you find multiple symptoms, start looking for a common cause. Only one thing in the set fits that, and it is the DC power-supply. Solid-state sets, too, can suffer from the problem. In sets with low-voltage DC supplies derived from the flyback, look for pulses and hash on the DC at the horizontal frequency. Each supply has a rectifier diode *and* a filter capacitor. Check to make sure it is not open, leaky, or suffering from a high power-factor.

Your scope is the only instrument that will positively locate and identify the cause of that sort of trouble, so, if you suspect it, grab the scope probe and start looking; it can save you a whole lot of time.

R-E

SERVICE QUESTIONS

LOW VOLTAGE

I've got a raster on this Sharp 3W77, but it is about an inch short on either side. All of the DC voltages (out of the flyback) are too low. There are also three pictures, side-by-side, on the screen.—J.C., West New York, NJ

I think that the last symptom is the key to the entire problem. If your horizontal oscillator is that far off, it will upset the voltage from the flyback, and just about everything else. Fortunately, that is more or less a standard oscillator/AFC circuit. Ground the top of the AFC diode. That will kill the AFC and let the oscillator free-wheel. Now, tune the coil until you get just a single picture, with straight sides, floating back and forth. That indicates that the oscillator is back on fre-

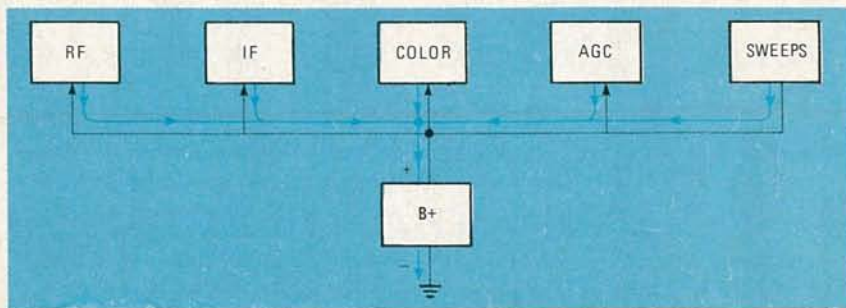
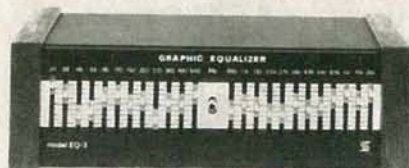


FIG. 1



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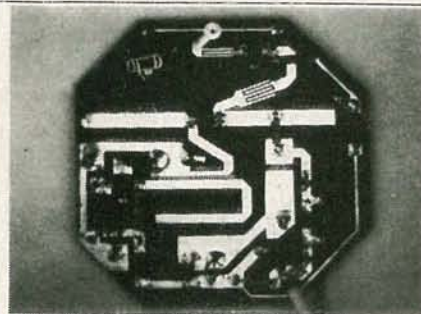
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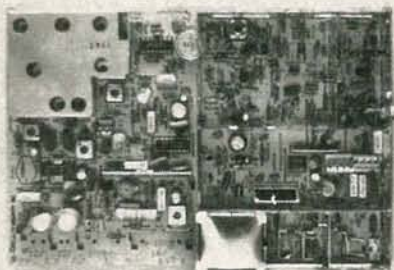
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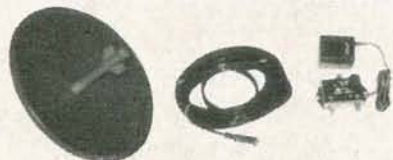
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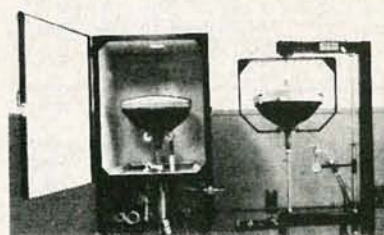
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quency. Remove the short from the AFC diode and the picture should lock in; if not, your AFC diodes may be bad—check and replace them if necessary.

NO COLOR

I wasn't getting any color from a KE-chassis GE set. I made the tests you suggested and found the 2700-ohm plate resistor of the 6BW11 open. Thanks for the help.—G.M., Walled Lake, MI

THERMAL PROBLEM

I've got a thermal problem in a Sampo M-12B. The vertical lock is poor and, when it does lock, retrace lines are visible. I froze C604 and the problem disappeared until it warmed up. I replaced the capacitor, and the same thing happened. I also froze C505 and the problem disappeared...temporarily. A new C505 didn't help; nor did a new vertical IC and sync-separator transistor. Any ideas?—D.B., Selah, WA

You obviously have a thermal problem—the question is where! Try cooling some of the resistors. While I tend to agree that the little low-voltage electrolytics are a likely cause, in your case they seem to be OK (installing new ones isn't a bad idea, though).

Resistors are subject to thermal drift, and that could easily account for the far-off DC voltages you found at the sync separator. As a last resort, check all the solder joints in the area you were cooling; cold-solder joints can be responsible for some weird problems. R-E

JUST A FEW OHMS OFF

This Zenith 23GC45 suffered from poor vertical-height. I subbed the 9-92 module, but that didn't help. It finally turned out that R203—a 4.7-ohm resistor from pin 9 to ground—had increased in value to 6.5 ohms. I replaced the resistor and that took care of the problem. A good low-resistance scale on your meter is a handy tool!—D.H., Vacaville, CA
 Amen!

FULL-WAVE BRIDGE QUESTION

I'm puzzled about something I found while fooling around. If I feed an audio signal into a full-wave bridge rectifier, the resulting sound is badly garbled. If I add a DC bias to the rectifier's AC inputs, the audio sounds normal. Can you tell me what's going on?—J.G., Morgan Hill, CA

I hope so. When you feed AC to a full-wave bridge rectifier, there are always two paths that conduct, and two that are blocked. That gives a DC voltage, with some ripple at the input frequency. That would be the equivalent of very severe clipping in an audio amplifier. The normal audio-signal at the output is AC floating on a DC signal; it can go either positive or negative, adding to or subtracting from the DC voltage.

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your audio signal above the DC level, which let it vary normally, and the audio came through.

MISLEADING SYMPTOM

The raster in this Magnavox T982 took on a trapezoidal shape; usually that means a shorted yoke-winding. If you're wondering whether we tried changing the yoke, yes, we did.

Later, we found that the values of the filter capacitors in the power supply had decreased in value, allowing the vertical circuit to be "pumped" at the horizontal rate. Replacing the capacitors brought things back to normal. Watch out for this—not only in that chassis, but also in the T981 and T987.

Thanks to C.T., of Adelphi, MD, for the warning.

NEGATIVE VOLTAGE

There is no high-voltage, etc., on this J.C. Penney 685-4827. The B+ voltages seem to be OK and the horizontal oscillator seems to be working. When I checked around the output stage, I found a negative voltage on the plate of 40KD6 and the cathode of the 34CE3 damper tube. What is going on?—M.B., Newport News, VA

Try this: Hook a DC voltmeter to the cathode of the damper tube and turn the power on. Watch the meter—if it comes up to full B+, disconnect the meter

quickly. If it shows zero, or, as you found, a small negative voltage, the cathode of the damper is probably open inside the tube. Check the continuity from the damper cathode to the plate of the 40KD6; that checks the primary winding of the flyback. Also, try a new damper tube—that is rarely the problem, but it is possible.

NON-STARTING VERTICAL OSCILLATOR

I've checked all the parts in the vertical circuit of this RCA KCS204A, but the oscillator simply won't start. I don't get it.—C.A.S., Snellville, GA

Oscillators that won't start, even though all the parts check out OK, have been a common problem for years. Just remember that if a circuit won't work, there almost always has to be a bad part! In your case, I'd suspect an open capacitor in the feedback loop—if there's no feedback, there's no oscillation.

Double check: feed a 60-Hz signal to the input of the first stage of the oscillator. If you now get deflection, the amplifier part of the circuit is OK. (All vertical oscillators of this type are really just two-stage amplifiers with very heavy feedback from input to output.) If the amplifier is working, then the problem is in the feedback loop, and very likely to be an open capacitor. Check C609 and the others to see whether they're good. R-E



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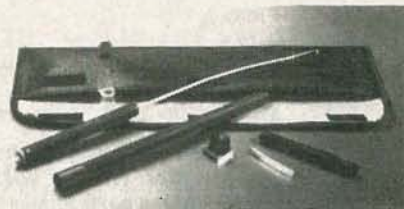
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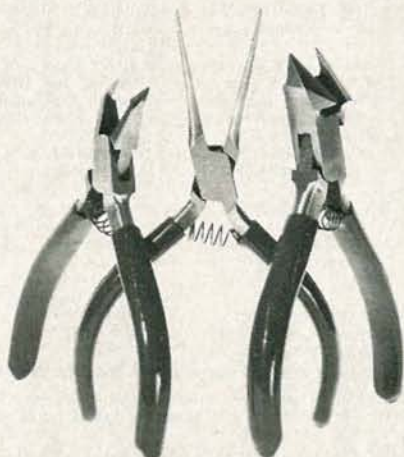
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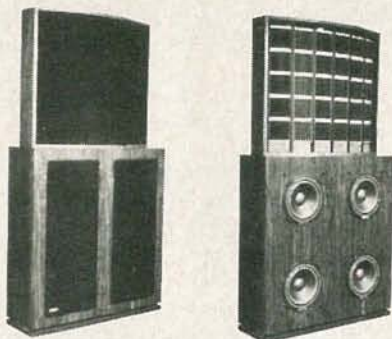
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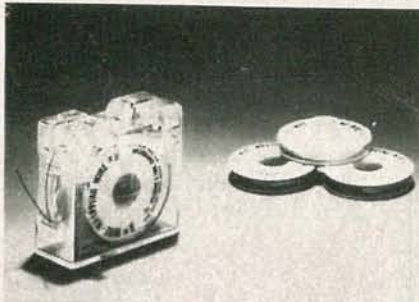
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strip mechanism, the wire is first drawn out to desired length; then a built-in plunger cuts the length free from the roll, while a gentle pull through the stripping blade removes the insulation. The model *WD-30-TRI* is priced at \$9.17; the three-color refill, model *R-30-TRI*, is priced at \$7.01.—**OK Machine and Tool Corporation**, 3455 Conner Street, Bronx, NY 10475. R-E

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HOBBY CORNER

Help from our readers

EARL "DOC" SAVAGE, K4SDS, HOBBY EDITOR

THIS MONTH SEEMS TO BE A GOOD TIME TO pause and thank those of you who have written to offer help to others. Your answers and suggestions are appreciated. It is a good time, too, to pass along some of your helpful hints to those who may have experienced similar problems.

Little shocker

Ernest Worley of Irvine, CA went to the trouble of analyzing the "Little Shocker" circuit that appeared in this column in the August 1981 issue. Among several improvements he suggested are two of special interest.

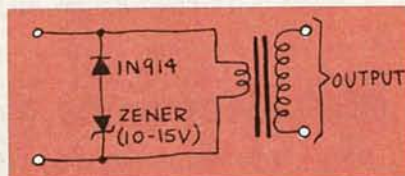


FIG. 1

The most significant change is shown in Fig. 1. If you compare it with the original circuit, you will see that Ernest has added a 12-volt Zener diode in series with the 1N914 across the primary of the output transformer. The Zener allows the transformer to make a more complete

AN INVITATION

To better meet your needs, "Hobby Corner" will undergo a change in direction. It will be changed to a question-and-answer form in the near future. You are invited to send us questions about general electronics and its applications. We'll do what we can to come up with an answer or, at least, suggest where you might find one.

If you need a basic circuit for some purpose, or want to know how or why one works, let us know. We'll print those of greatest interest here in "Hobby Corner." Please keep in mind that we cannot become a circuit-design service for esoteric applications; circuits must be as general and as simple as possible. Please address your correspondence to:

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recovery with each pulse while still protecting the transistor from breakdown.

In addition, Ernest adjusted the values of the capacitor and resistors to increase the frequency to about 350 Hz and to bring the duty cycle to 50%. The end result is to give the shocker a bit more "zing" and, more importantly, to con-

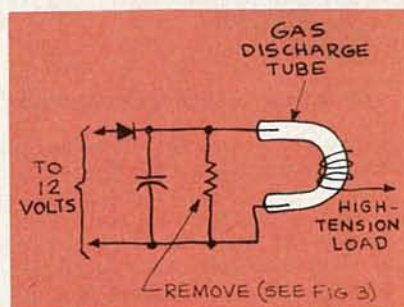


FIG. 2

former. If it is not a circuit like an audio amplifier, where fidelity is important, give the Zener a try. It may improve circuit efficiency and conserve battery power.

Thanks, Ernest, for sharing the battery-saving-Zener with us.

Induction timing-light

David Reading of Marshfield, WI offers a good solution to James McDaniel's desire to change his timing light to the induction pick-up variety (April 1982). Careful selection of the few extra parts required—especially the triggering transformer—will allow you to fit it all into the original light case.

Figure 2 shows a basic high-tension timing-light circuit. Figure 3 shows a circuit with David's added parts (note that the resistor shown in Fig. 2 has been removed). He says any 4-kV trigger

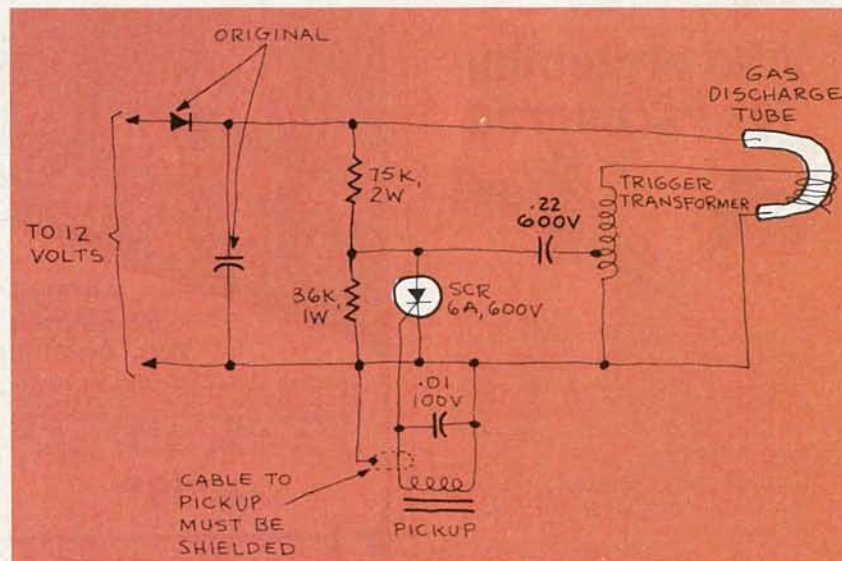


FIG. 3

serve battery life. You may recall that short battery-life was a characteristic of the original circuit, and the changes help get rid of that problem.

I am not calling Ernest's changes to your attention so much because the shocker circuit is especially deserving, but rather because the addition of the Zener diode across the primary of the transformer can be applied to other situations.

Often you will run across designs that couple transistor outputs through a trans-

former will work fine and suggests one sold by Mouser Electronics (11433 Woodside Ave., Santee, CA 92071) as their stock number 42FM401.

David has tried several different styles of induction pickups. All had three to five turns of wire around the core and they all worked fine. He cautions that the SCR must not be rated any lower than six amps at 600 volts.

Many thanks, David. Now all of us can give new life to those old timing lights.

Bullet velocity

Randolph Richter (Schenectady, NY), J.A. Keys (Long Beach, CA) and others were quite helpful with the problem of measuring bullet velocity (December 1981). The solution, of course, is to measure the time it takes the bullet to travel over a known distance as accurately as possible.

The usual approach is to place two wires a given distance apart and then break them with the bullet. I acknowledge that the most difficult part of that procedure, at least to me, would be to hit the two wires! The experts tell me, however, that it isn't too hard to do provided that the wires are close and the rifle (or pistol) is supported by sand bags.

In any case, it is easy enough to measure the distance between the wires. Now, if you only knew the length of time between the breaking of the first wire and the breaking of the second one, finding the velocity would be easy.

Everyone seemed to go about timing the breaks in just about the same way. A counter circuit counts the pulses from a fast clock. The clock is triggered by the breaking of the wires—the first one starts it when broken, and the second one stops it. Knowing the clock rate and the number of counts, it is simple to figure the time that has elapsed between the breaking of the two wires.

After you know both the distance and the time, you can divide the first by the second to get the velocity. Let's now look at the actual techniques our readers used.

J.A. and Randolph use different circuits to measure the time. J.A. uses an internally triggered commercial counter that's enabled by an EXCLUSIVE-OR gate. The two wires are, of course, connected to the gate inputs. When one is broken the counter starts; it stops when the second wire is broken.

Randolph, on the other hand, built his gear from designs in Radio Shack's *Engineer's Notebook*. His circuit also consists of counters and a clock, but the triggering setup he uses is a little different. In that setup, he has the clock running before he fires and uses the breaking of the first wire to start the counter.

Breaking the second wire stops the clock by turning off its power. When the clock stops, there are no more pulses for the counter to count and the display freezes, showing the number of pulses that were counted between the wire breaks. Neat!

If you are thinking of measuring bullet velocities, I should tell you that almost every response to this problem that I received included one piece of advice: OBSERVE FIREARM SAFETY PROCEDURES! That is very important and only a fool thinks it is "silly" to repeatedly go through a safety checklist when conducting tests like this.

Forgetting about guns for the moment,

you can measure the speed of almost anything with the system described here. It will completely eliminate errors due to human reaction-time, as well as errors (real or imagined) that might be due to favoritism.

Certainly, anything slower than a bullet will present no difficulty. All you have to do is to slow down the clock to a rate that is appropriate to the event you are measuring.

Mosquitoes again

Summer is an interesting time of the year. Many of you become involved in warm-weather activities and electronics experimenting receives less attention than it does in other seasons. Your outside activities bring to mind (and body) the problem of mosquitoes, and what can be done about them.

Last summer was no exception. Several readers wrote to ask for information or a circuit for a mosquito repeller. If you have a problem with those infernal pests, you may wish to refer to "Hobby Corner" in the March 1980 and February 1981 issues of **Radio-Electronics**. There you will find a circuit and some comments about its value.

It may be helpful to summarize the comments I have received over the last couple of years. Before you plunge into building a repeller, you should know that the number of readers who think it is great is about equalled by the number who think it is of no value at all! Until there is more information, I wouldn't recommend that you build one except on an experimental basis.

I am beginning to think that the effectiveness of a mosquito repeller is, like beauty, in the eye of the beholder. Certainly, there is no consensus of opinion on the question.

It occurs to me that, in addition to the psychological aspects of your reactions, there is the probability that certain components in the circuitry are critical. The speaker, for instance, could make a considerable difference in the unit's effectiveness. (Has anyone tried using one of those inexpensive piezo-element tweeters?) In addition, I have been led to believe that the frequency of the oscillator is critical.

It would be nice to resolve the question once and for all. If you have any information on the effect of sound on mosquitoes (or other insects), how about sharing it with the rest of us? If you have built a repeller that seems to work, let us know the exact parts you used and the oscillator frequency, as close as you can measure it. If you have done any actual experimenting with sound and the little beasties, tell us about your results.

Because of the continuing interest, I'll start a "mosquito" file and share its contents from time to time. If you have anything to contribute to that file please write in and share it with us.

R-E

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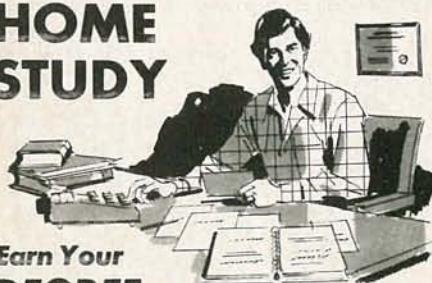
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ANALOG CIRCUIT

continued from page 78

of r_c and r_b are negligible when compared to R_E and R_B and can therefore be ignored. The load across R_E can be very small, but should be no less than ten times the calculated output impedance.

In addition to the direct-coupled Darlington arrangement, transistors can be connected in series or in parallel. When

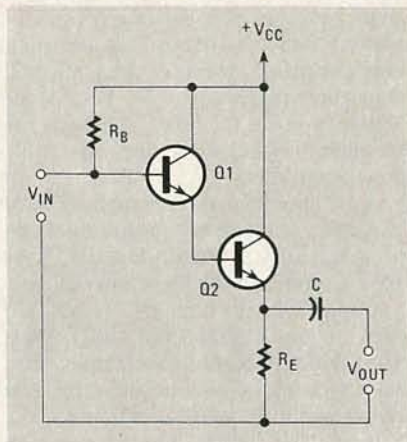


FIG. 6—THE DARLINGTON AMPLIFIER consists of two cascaded emitter-follower circuits.

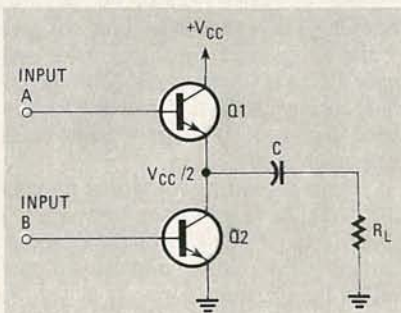


FIG. 7—WHEN TWO TRANSISTORS ARE connected in series, the supply voltage ideally will be split evenly between the two devices.

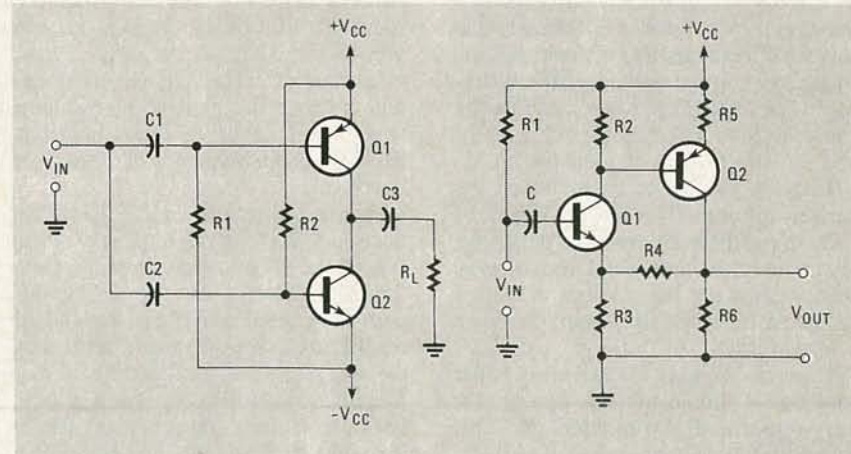


FIG. 8—COMPLEMENTARY TRANSISTOR CIRCUITS. The one shown in *a* is a push-pull audio amplifier. The one shown in *b* can be modified to meet a variety of circuit requirements.

connected in parallel, a two-transistor combination can theoretically deliver twice the current each individual device is capable of supplying. For that to be true,

the current they must pass must divide evenly between the two devices. That of course, never happens. Actually, the current rating of each transistor should be reduced by at least 20% to keep the devices in the circuit from being destroyed. One method of equalizing the currents in the transistors is to place a small resistor in series with the emitters of each one. That, however, reduces the amount of power the circuit can deliver for a given power-supply voltage. As for the series circuit, ideally the supply voltage divides evenly between the two devices, as shown in Fig. 7.

Complementary pairs are used quite frequently in direct-coupled circuits. One common use for that type of arrangement is in a push-pull audio-amplifier circuit, such as the one shown in Fig. 8-a. In that amplifier, the positive half of the input signal's cycle is fed to one transistor, in this case Q2, while the negative half is fed to the other, in this case Q1. The two halves of the signal combine across R_L to reproduce the full cycle. In a practical application, R_L could be a loudspeaker.

The complementary-pair circuit shown in Fig. 8-b can be modified depending on the requirements of the circuit. In the circuit as shown, the current gain is the product of the betas of the two transistors, the voltage gain is $1 + (R_4/R_3)$, and the power gain is the product of the voltage and current gains. The input impedance is approximately equal to R_1 while the output impedance is $R_4R_6/(R_4 + \beta_2R_6)$, where β_2 is the beta of Q2.

Two variations of the circuit shown in Fig. 8-b are frequently used. In one, R_4 is omitted while R_3 and R_5 are shorted. Now, voltage gain drops to 1 and the output impedance becomes equal to R_6 .

In the second variation, R_2 and R_3 are omitted while R_4 and R_5 are shorted. Now the characteristics are the same as for the first variation except that the output impedance drops to near zero.

Another form of direct-coupled circuit is the differential amplifier. That extremely useful circuit will be the subject of the next part of this article. **R-E**

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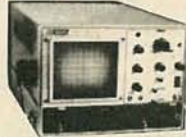
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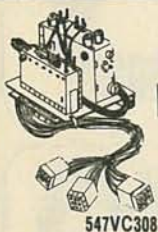


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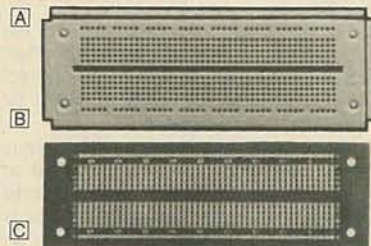
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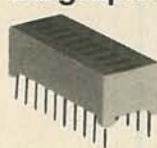


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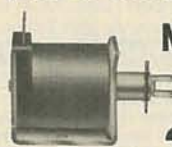
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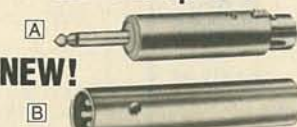
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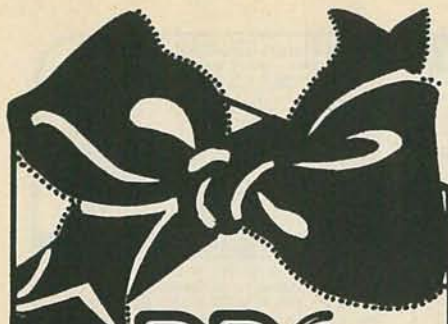
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RADIO-ELECTRONICS

BAYONET Lamp SOCKETS
SNAP-IN SOCKET FITS BAYONET BASE LAMPS (SEE P-2576, BELOW). SNAPS INTO A 5/8" HOLE, HAS 6" LONG WIRE. LEADS: B-2575

15¢ 10/\$1.20 100/\$10.
WT. 2 OZ. WT. 1/2 LB.

DIP SOCKETS

8 PIN 16¢
S-0800

14 PIN 18¢
S-1400

16 PIN 20¢
S-1600

18 PIN 25¢
S-1800

24 PIN 38¢
S-2400

28 PIN 45¢
S-2800

40 PIN 55¢
S-4000

EDGE Connector
* GOLD WW 3-LEVEL PINS *
C-0356 10/\$9. 1 LB
6 LB 100/\$80.
1000/\$750. 60 LB

WOW! BRAND
NEW, HIGH QUALITY CASES AT A FRACTION OF THEIR WORTH

5 x 7 x 12"
6 LB X-0995

INSTRUMENT CASE
LOCAL MFR. NEEDS LEADER SCOPES BUT DOESN'T NEED THE CASES! SO, HE SELLS 'EM TO US!! THE FRONT PANELS ARE PUNCHED, BUT REPLACES WITH 2 SCREWS. BACK HAS BANANA JACKS AND CORD WINDER. TWO TONE BLUE COLOR GREAT FOR HOMEBREW PROJECTS

\$9.95

BAYONET 13V BULBS
MR. GE 1816 330 MA T-3 3/4
LIST PRICE IS 10/92 70

10/75¢ 3/4 LB 100/\$6.50
WT. 2 OZ. 1000/\$55 1/2 LB

BCD Thumbwheel SWITCH
WITH COMPLIMENT 0-9

B-2124 GOLD CONTACTS STD SIZE CAN BE STACKED TO ANY WIDTH

1.25 TO ANY WIDTH COMPLIMENT

6 VOLT 5 AMP-HR RECHARGEABLE SEALED BATTERY

B-5020 3 LB **\$9.50**

LEAD ACID BATTERY
SIZE: 5-3/8" X 3-1/8" X 1-7/8"

PC TRANSFORMER DUAL 13V 350 MA

T-1350 6 OZ

1.50 1-5/8" X 1-1/4" X 1-1/4"

VHF/UHF CHANNEL A-I MIDBAND Varactor TUNER

HI-GAIN, BRAND NEW, FOR UPGRADE OR CONVERTER USE. IF OUTPUT, USES +12V AND +24V SUPPLY. T-1195 WT 1/2 LB

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HEAVY DUTY SPST CAST ALUMINUM CASE

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GOOD FOR A FEW DOZEN AMPS

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TAPESWITCH MODEL CVP 623 IS UL APPROVED MEASURES 6" X 23" OUTPUT IS 2 PIN MOLEX

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FAST & EASY TO USE, MIX 1/2 POUND TO A GALLON OF WATER. (OZ TO 1 PINT AND ETC.) THE WHITE CRYSTALS MAKE A CLEAR SOLUTION THAT TURNS BLUE WHEN FULL OF COPPER.

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(MAKES 2 GAL.) **\$1.95**

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50¢ EA. B-1195 RED B-1194 BLUE

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QUALITY BUILT BY MOTOROLA WITH AUTO TURN ON, END OF TAPE SENSOR, AUTO TRACK CHANGE, 115 VAC CAPACITOR START MOTOR, STEREO R/P HEAD AND +12 VDC PLAY BACK ELECTRONICS. GREAT FOR HOMEBREW PHONE ANSWERERS, ROBOTS, ETC.

3 AG FUSES

F-0102 WT 1 OZ

35 A 32 V MFD BY LITTLE FUSE

F-0101 WT 1 OZ

1-1/2A 250 V MFD BY BUSS

ANY QUANTITY **10¢ EACH** YOUR CHOICE

.068 mfd 250 V Molded CAP

C-1015 4 X 3 X 1"

25/\$1 WT 2 OZ

4 OZ 250/\$8

1000/\$30 1/2 LB

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PO BOX 12095 SARASOTA, FL. 33578

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Lowest Prices Video and Electronic Components

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Select from two inputs to VCR or TV



- A-CS75/75 75 ohm inputs/75 ohm output \$2.75 ea. 10/\$22.00
- A-CS75/75P Professional version of above \$7.50 ea. 10/\$60.00
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Converts 75 ohm to 300 ohm inputs



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Splits one input to multiple outputs



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- A-CMS3 3-way splitter (75 ohm) \$2.25 ea. 10/\$18.00
- A-ACP4 4-way splitter (300 ohm) \$2.25 ea. 10/\$18.00

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- A-ABS-U Back-of-set VHF/FM antenna \$4.25 ea. 10/\$34.00

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- A-VA9 F jack to RCA plug adaptor \$1.49 ea. 10/\$12.00
- A-VA10 RCA jack to F plug adaptor \$1.49 ea. 10/\$12.00
- A-VA8 RCA jack to BNC plug adaptor \$1.69 ea. 10/\$13.50
- A-VA2 Double female BNC adaptor \$1.99 ea. 10/\$16.00
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(Molded heavy-duty construction)



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- A-VA06 Quick F plug to quick F plug (6 ft.) \$2.99 ea. 10/\$24.00
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Each kit in reusable compartmentalized storage cabinet

- SK-CD 5PI to .1Mf (150 pcs)
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- SK-RM50 10 ohms to 10 meg (350 pcs)

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30 watt soldering iron with each kit ordered (\$4.99 value)



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\$50 order receive free auto fm signal booster (\$12.50 value)
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* CHECK * MONEY ORDER
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8202	10.95	825F (AM9617)	7.95
8212	3.25	82C00	15.00
8214	3.80	82C02	6.00
8216	3.80	280A CPU	3.75
8224	3.25	280A SIO	12.95
MM5307	8.95	280A PIO	3.75
8228	2.75	280CTCA	3.75
8228	8.50	1M82827HL	9.95
8155	3.00	8275	15.95
8227	14.00	8910	13.95
3242	6.00	8910	2.50

RAM's

21L02-3	70	2706	2.95
95L422	5.95	2716 - 8V	4.25
2901-1	2.45	2732	8.95
2114L	1.85	2932	7.95
2114L-3	1.85	2706-25	18.95
7M4500	4.95	82533	1.95
MK400P	1.95	825123	1.95
MK427-3	1.95	825129	1.95
7M4500BL	2.95	825131	1.95
MK400-11	1.95	3828A-3	3.00
4116-2	1.80	AM3124C	2.00
4116-2	1.80	6256-4 (P188A)	2.00
4116-15	1.95	745424	6.95
MK400-2-3	10.95		
816-3	6.95		
28104-4	2.50		
816-3	6.95		
8256(486-3)	7.25		

ROM's

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95L422	5.95	2716 - 8V	4.25
2901-1	2.45	2732	8.95
2114L	1.85	2932	7.95
2114L-3	1.85	2706-25	18.95
7M4500	4.95	82533	1.95
MK400P	1.95	825123	1.95
MK427-3	1.95	825129	1.95
7M4500BL	2.95	825131	1.95
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4116-2	1.80	6256-4 (P188A)	2.00
4116-15	1.95	745424	6.95
MK400-2-3	10.95		
816-3	6.95		
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8256(486-3)	7.25		

INTERFACE & DRIVERS

1488	80	MM1402	1.75
1488	80	MM1403	1.75
8130	2.50	MM1404	1.75
8820	2.50	MM1513	2.50
8131	2.50	MM1555	2.50
8833	2.50	MM1556	2.50
8834	2.50	MM1557	2.50
8837	2.50	MM1558	2.50
MM5211	9.00	MM1559	2.50
MM5211	2.50		

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7142B	3.25	COM4017	3.75

RAM's

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50 PIN EDGEBOARD CONN.	3.95
26 PIN EDGEBOARD CONN.	2.50
50 PIN RIGHT ANGLE CONN.	3.95
20 PIN RIGHT ANGLE CONN.	2.25

50 PIN RIBBON CONNECTORS	\$2.25
34 PIN RIBBON CABLE CONN.	\$2.50
34 PIN RIBBON CABLE CONN.	\$2.50
10 PIN RIBBON CABLE CONN.	\$1.50
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4" x 6" DOUBLE SIDED EPOXY BOARD 1/16" THICK	\$1.62 ea.

EPOXY GLASS VECTOR BOARD

1/16" thick with 1/10" spacing
4 1/2" x 6 1/2" \$1.95

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1.000	4.000	6.144	18.000
2.000	5.000	8.000	18.432
3.000	6.000	10.000	20.000
3.579			

7 WATT LD 65 LASER DIODE(R) \$8.95

25 watt Infra Red Pulse (SG 2006 equiv.)
Laser Diode (Spec sheet included) \$24.95

2N3820 P FET	\$.45
2N5457 N FET	\$.45
2N2646 UJT	\$.45
ER 900 TRIGGER DIODES	\$4.91/00
1N 6028 PROS. UJT	\$.65

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1UF 16V	10/ \$1.00	100/ \$8.00
01UF 35V	16/ \$1.00	100/ \$5.00

IN4148 (IN914) 15/1.00

TANTALUM CAPACITORS

22UF 35V	5/ \$1.00	10UF 20V	\$.40
68UF 35V	5/ \$1.00	22UF 10V	\$.30
10UF 20V	5/ \$1.00	15UF 16V	3/ \$1.00
2.2UF 20V	5/ \$1.00	30UF 6V	5/ \$1.00
3.3UF 20V	4/ \$1.00	33UF 10V	\$.40
4.7UF 35V	4/ \$1.00	47UF 20V	\$.85
6.8UF 35V	3/ \$1.00	68UF 10V	\$1.00
		120UF 6V	\$.75
		200UF 20V	\$1.75

DIP SOCKETS

8 PIN	.17	20 PIN	.30
14 PIN	.20	24 PIN	.35
16 PIN	.22	28 PIN	.40
18 PIN	.25	40 PIN	.60

FLAT RIBBON CABLE GRAY, 28 gauge

26 conductor 60/ft 40 conductor 90/ft
50 conductor 1.00

.5V at 800ma SOLAR CELLS 3" diameter \$4.35

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DB 25P male	\$2.75
DB 25S female	3.75
HOODS	1.25

POSTAGE ADD 10% FOR ORDERS UNDER \$25.00

RATES ADD 5% FOR ORDERS BETWEEN \$25.00 AND \$50.00
ADD 3% FOR ORDERS ABOVE \$50.00

C/MOS

4001	35	4028	85	4081	30	74C23	45
4002	35	4029	75	4082	30	74C24	1.75
4006	45	4030	35	4083	80	74C27	1.50
4007	27	4034	75	4089	65	74C28	1.50
4008	70	4036	75	4091	95	74C29	70
4009	45	4040	75	4092	65	74C30	1.40
4010	45	4042	85	4091	85	74C38	80
4011	32	4044	75	4094	140	74C39	80
4012	32	4044	85	4095	180	74C83	80
4013	70	4046	90	4096	70	74C184	2.50
4014	70	4047	25	4098	125	74C185	1.75
4015	70	4048	40	4099	70	74C186	1.30
4016	80	4049	85	4099	85	74C187	1.30
4017	60	4091	80	4095	275	74C183	1.15
4021	35	4092	85	4096	27	74C182	1.30
4019	45	4093	75	4098	85	74C173	1.30
4020	70	4094	90	4099	27	74C175	1.30
4022	80	4099	29	74C104	90	74C90	1.50
4025	35	4072	35	74C106	30	74C98	1.50
4024	45	4071	25	74C120	27	74C94	1.75
4026	35	4078	85	74C124	120	74C91	1.50
4027	80	4077	35	74C124	120	74C91	1.50

DOUBLE SIDED DIP RIBBON CABLE JUMPER ASSEMBLIES

16 PIN 4" LONG	\$ 2.00
14 PIN 12" LONG	\$ 2.00

RS 232 CABLE

10 Conductor, #22 color coded wire, gray PVC outer cover, 3/8" diameter
40 per ft. - 100/ \$30.00
Add 20% postage for orders under 100'
Add 10% postage for orders over 100'.

SPECIALS

CPU's		CRT Controllers	
6502	\$6.00	TMS9927NL	\$ 9.95
8080A	2.75	6845	13.95
2901A	3.75	8275	16.95

ROM's

2732	\$ 7.95
2716	\$ 4.00

RAM's

2114L-4	\$1.35
4116-2	1.60
MK4802-3-2KX8	8.95
93L422 BIPOLAR	5.95
4164-2-64K	7.95
2147 J3	4.95
4118-4	6.95
D765C	25.00
68000LB	\$70.00

DISC Controllers

1791	25.00
1793	35.00
1795	45.00
1797	45.00
2950	25.00

SPECIALS GOOD THRU DEC. 1982

NO. 30 WIRE WRAP WIRE SINGLE STRAND 100' \$1.40

DIP SWITCHES

CTS 206 4	4 POSITION 1.25
CTS 206 7	7 POSITION 1.40
CTS 206 8	8 POSITION 1.50
CTS 206 10	10 POSITION 1.95

TOGGLE SWITCHES

1000 - SPDT	1.10
2060 - DPDT	1.50
206P - DPDT	CENTER OFF \$1.75

SCR's

1.5A	6A	35A	110A
100	45	60	1.40
200	70	80	1.90
400	120	140	2.60
600	180	3.60	15.00

4164 64K DYNAMIC \$625
200 NS

TMM2016 2KX8 STATIC \$415
200 NS

ALL MERCHANDISE 100% GUARANTEED!

CALL US FOR VOLUME QUOTES

STATIC RAMS

2101	256 x 4 (450ns)	1.95
5101	256 x 4 (450ns) (cmos)	3.95
2102-1	1024 x 1 (450ns)	.89
2102L-4	1024 x 1 (450ns) (LP)	1.29
2102L-2	1024 x 1 (250ns) (LP)	1.69
2111	256 x 4 (450ns)	2.99
2112	256 x 4 (450ns)	2.99
2114	1024 x 4 (450ns)	8/14.95
2114L-4	1024 x 4 (450ns) (LP)	8/15.25
2114L-3	1024 x 4 (300ns) (LP)	8/15.45
2114L-2	1024 x 4 (200ns) (LP)	8/15.95
2147	4096 x 1 (55ns)	4.95
TMS4044-4	4096 x 1 (450ns)	3.49
TMS4044-3	4096 x 1 (300ns)	3.99
TMS4044-2	4096 x 1 (200ns)	4.49
MK4118	1024 x 8 (250ns)	9.95
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TMM2016-150	2048 x 8 (150ns)	4.95
TMM2016-100	2048 x 8 (100ns)	6.15
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HM6116-3	2048 x 8 (150ns) (cmos)	5.95
HM6116-2	2048 x 8 (120ns) (cmos)	8.95
HM6116LP-4	2048 x 8 (200ns) (cmos)(LP)	6.95
HM6116LP-3	2048 x 8 (150ns) (cmos)(LP)	8.95
HM6116LP-2	2048 x 8 (120ns) (cmos)(LP)	10.95
Z-6132	4096 x 8 (300ns) (Qstat)	34.95

LP - Low Power Qstat - Quasi-Static

DYNAMIC RAMS

TMS4027	4096 x 1 (250ns)	1.99
MK4108	8192 x 1 (200ns)	1.95
MM5298	8192 x 1 (250ns)	1.85
4116-300	16384 x 1 (300ns)	8/11.75
4116-250	16384 x 1 (250ns)	8/11.95
4116-200	16384 x 1 (200ns)	8/13.95
4116-150	16384 x 1 (150ns)	8/15.95
4116-120	16384 x 1 (120ns)	8/29.95
2118	16384 x 1 (150ns) (5v)	4.95
MK4816	2048 x 8 (300ns) (5v)	24.95
4164-200	65536 x 1 (200ns) (5v)	6.25
4164-150	65536 x 1 (150ns) (5v)	7.25

5V - single 5 volt supply

EPROMS

1702	256 x 8 (1us)	4.50
2708	1024 x 8 (450ns)	3.95
2758	1024 x 8 (450ns) (5v)	5.95
2716	2048 x 8 (450ns) (5v)	3.95
2716-1	2048 x 8 (350ns) (5v)	6.25
TMS2716	2048 x 8 (450ns)	7.95
TMS2532	4096 x 8 (450ns) (5v)	7.95
2732	4096 x 8 (450ns) (5v)	4.95
2732-250	4096 x 8 (250ns) (5v)	12.95
2732-200	4096 x 8 (200ns) (5v)	16.95
2764	8192 x 8 (450ns) (5v)	16.95
2764-250	8192 x 8 (250ns) (5v)	18.95
2764-200	8192 x 8 (200ns) (5v)	19.95
TMS2564	8192 x 8 (450ns) (5v)	24.95
MC68764	8192 x 8 (450ns) (5v)(24 pin)	call

5v - Single 5 Volt Supply

EPROM ERASERS

	Timer	Capacity Chip	Intensity (uW/Cm ²)	
PE-14		6	5,200	83.00
PE-14T	X	6	5,200	119.00
PE-24T	X	9	6,700	175.00
PL-265T	X	20	6,700	255.00
PR-125T	X	16	15,000	349.00
PR-320	X	32	15,000	595.00

DISC CONTROLLERS

1771	16.95
1791	29.95
1793	38.95
1795	54.95
1797	54.95
6843	34.95
8272	39.95
UPD765	39.95
1691	18.95
2143	18.95

INTERFACE

8T26	1.69
8T28	2.49
8T95	.99
8T96	.99
8T97	.99
8T98	.99
DM8131	2.95
DP8304	2.29
DS8835	1.99
DS8836	.99

MISC.

3242	7.95
3341	4.95
MC3470	4.95
MC3480	9.00
11C90	13.95
95H90	7.95
2513-001 UP	9.95
2513-002 LOW	9.95

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76477	3.95
76489	8.95
AY3-8910	12.95
MC3340	1.49

CRT CONTROLLERS

6845	14.95
68B45	35.95
HD46505SP	15.95
6847	12.25
68047	24.95
8275	29.95
7220	99.95
CRT5027	39.95
CRT5037	49.95
TMS9918A	39.95

BIT-RATE GENERATORS

MC14411	11.95
BR1941	11.95
4702	12.95
COM5016	16.95
COM8116	10.95
MM5307	10.95

UARTS

AY3-1014	6.95
AY5-1013	3.95
PT1472	9.95
TR1602	3.95
2350	9.95
2651	8.95
TMS6011	5.95
IM6402	7.95
IM6403	8.95
INS8250	14.95

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AY5-2376	11.95
AY5-3600	11.95
74C922 See 74C00	
74C923 Series Prices	

CLOCK CIRCUITS

MM5314	4.95
MM5369	3.95
MM5375	4.95
MM58167	8.95
MM58174	11.95
MSM5832	6.95

Z-80 2.5 Mhz

Z80-CPU	3.95
Z80-CTC	5.95
Z80-DART	15.25
Z80-DMA	17.50
Z80-PIO	5.75
Z80-SIO/0	18.50
Z80-SIO/1	18.50
Z80-SIO/2	18.50
Z80-SIO/9	16.95

4.0 Mhz

Z80A-CPU	6.00
Z80A-CTC	8.65
Z80A-DART	18.75
Z80A-DMA	27.50
Z80A-PIO	6.00
Z80A-SIO/0	22.50
Z80A-SIO/1	22.50
Z80A-SIO/2	22.50
Z80A-SIO/9	19.95

6.0 Mhz

Z80B-CPU	17.95
Z80B-CTC	15.50
Z80B-PIO	15.50

ZILOG

Z6132	34.95
Z8671	39.95

CRYSTALS

32.768 khz	1.95
1.0 mhz	4.95
1.8432	4.95
2.0	3.95
2.097152	3.95
2.4576	3.95
3.2768	3.95
3.579535	3.95
4.0	3.95
5.0	3.95
5.0688	3.95
5.185	3.95
5.7143	3.95
6.0	3.95
6.144	3.95
6.5536	3.95
8.0	3.95
10.7836	3.95
14.31818	3.95
15.0	3.95
16.0	3.95
18.0	3.95
18.432	3.95
20.0	3.95
22.1184	3.95
32.0	3.95

DATA ACQUISITION

ADC0800	15.55
ADC0804	3.49
ADC0809	4.49
ADC0817	9.95
DAC0800	4.95
DAC0806	1.95
DAC0808	2.95
DAC1020	8.25
DAC1022	5.95
MC1408L6	1.95
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8035	5.95
8039	6.95
INS-8060	17.95
INS-8073	24.95
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8085	5.95
8085A-2	11.95
8086	29.95
8087	CALL
8088	39.95
8089	89.95
8155	7.95
8156	8.95
8185	29.95
8185-2	39.95
8741	39.95
8748	29.95
8755	32.00

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8202	29.95
8203	39.95
8205	3.50
8212	1.80
8214	3.85
8216	1.75
8224	2.25
8226	1.80
8228	3.49
8237	19.95
8238	4.49
8243	4.45
8250	10.95
8251	4.49
8253	6.95
8253-5	7.95
8255	4.49
8255-5	5.25
8257	7.95
8257-5	8.95
8259	6.90
8259-5	7.50
8271	39.95
8272	39.95
8275	29.95
8279	8.95
8279-5	10.00
8282	6.50
8283	6.50
8284	5.50
8286	6.50
8287	6.50
8288	25.00
8289	49.95

6500 1 MHZ

6502	5.95
6504	6.95
6505	8.95
6507	9.95
6520	4.35
6522	8.75
6532	11.25
6545	22.50
6551	11.85

2 MHZ

6502A	9.95
6522A	11.70
6532A	12.40
6545A	28.50
6551A	12.95

3 MHZ

6502B	14.95
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FUNCTION GENERATORS

MC4024	3.95
LM566	1.49
XR2206	3.75
8038	3.95

INTERSIL

ICL7103	9.50
ICL7106	9.95
ICL7107	12.95
ICL8038	3.95
ICM7207A	5.59
ICM7208	15.95

6800

68000	99.95
6800	4.95
6802	7.95
6808	13.90
6809E	19.95
6809	12.95
6810	2.95
6820	4.95
6821	3.25
6828	14.95
6840	12.95
6843	34.95
6844	25.95
6845	14.95
6847	12.25
6850	3.45
6852	5.75
6860	9.95
6862	11.95
6875	6.95
6880	2.25
6883	24.95
68047	24.95
68488	19.95

6800 1MHZ

68B00	10.95
68B02	22.25
68B09E	29.95
68B09	29.95
68B10	7.95
68B21	12.95
68B45	35.95
68B50	12.95

68B00 2MHZ

6500 1 MHZ

6502	5.95
6504	6.95
6505	8.95
6507	9.95
6520	4.35
6522	8.75
6532	11.25
6545	22.50
6551	11.85

2 MHZ

6502A	9.95
6522A	11.70
6532A	12.40
6545A	28.50
6551A	12.95

3 MHZ

6502B	14.95
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EXAR

XR 2206	3.75
XR 2207	3.85
XR 2208	3.90
XR 2211	5.25
XR 2240	3.25

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9316	1.00
9334	2.50
9368	3.95
9401	9.95
9601	.75
960	

74LS00

74LS00	.24	74LS86	.39	74LS169	1.75	74LS323	2.75
74LS01	.25	74LS90	.55	74LS170	1.49	74LS324	1.75
74LS02	.25	74LS91	.89	74LS173	.69	74LS352	1.29
74LS03	.25	74LS92	.55	74LS174	.55	74LS353	1.29
74LS04	.24	74LS93	.55	74LS175	.55	74LS363	1.35
74LS05	.25	74LS95	.75	74LS181	2.15	74LS364	1.95
74LS08	.28	74LS96	.89	74LS189	8.95	74LS365	.49
74LS09	.29	74LS107	.39	74LS190	.89	74LS366	.49
74LS10	.25	74LS109	.39	74LS191	.89	74LS367	.45
74LS11	.35	74LS112	.39	74LS192	.79	74LS368	.45
74LS12	.35	74LS113	.39	74LS194	.79	74LS377	.99
74LS13	.45	74LS114	.39	74LS195	.69	74LS379	1.39
74LS14	.59	74LS122	.45	74LS196	.69	74LS378	1.18
74LS15	.35	74LS123	.79	74LS197	.79	74LS375	1.35
74LS20	.25	74LS124	2.90	74LS221	.89	74LS385	1.90
74LS21	.29	74LS125	.49	74LS240	.95	74LS386	.45
74LS22	.25	74LS126	.49	74LS241	.99	74LS390	1.19
74LS26	.29	74LS132	.59	74LS242	.99	74LS393	1.19
74LS27	.29	74LS133	.59	74LS243	.99	74LS395	1.19
74LS28	.35	74LS136	.39	74LS244	.99	74LS399	1.49
74LS30	.25	74LS137	.99	74LS245	1.49	74LS424	2.95
74LS32	.29	74LS138	.55	74LS247	.75	74LS447	.37
74LS33	.55	74LS139	.55	74LS248	.99	74LS490	1.95
74LS37	.35	74LS145	1.20	74LS249	.99	74LS624	3.99
74LS38	.35	74LS147	2.49	74LS251	.59	74LS668	1.69
74LS40	.25	74LS148	1.35	74LS253	.59	74LS669	1.89
74LS42	.49	74LS151	.55	74LS255	.59	74LS670	1.49
74LS47	.75	74LS153	.55	74LS258	.59	74LS674	9.65
74LS48	.75	74LS154	1.90	74LS259	2.75	74LS682	3.20
74LS49	.75	74LS155	.69	74LS260	.59	74LS683	3.20
74LS51	.25	74LS156	.69	74LS266	.55	74LS684	3.20
74LS54	.29	74LS157	.65	74LS273	1.49	74LS685	3.20
74LS55	.29	74LS158	.59	74LS275	3.35	74LS688	2.40
74LS63	1.25	74LS160	.69	74LS279	.49	74LS689	3.20
74LS73	.39	74LS161	.65	74LS280	1.98	74LS783	24.95
74LS74	.35	74LS162	.69	74LS283	.69	81LS95	1.49
74LS75	.39	74LS163	.65	74LS290	.89	81LS96	1.49
74LS76	.39	74LS164	.69	74LS293	.89	81LS97	1.49
74LS78	.49	74LS165	.95	74LS295	.99	81LS98	1.49
74LS83	.60	74LS166	1.95	25LS2521	2.80		
74LS85	.69	74LS168	1.75	25LS2569	4.25		

IC SOCKETS

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8 pin ST	.13	.11
14 pin ST	.15	.12
16 pin ST	.17	.13
18 pin ST	.20	.18
20 pin ST	.29	.27
22 pin ST	.30	.27
24 pin ST	.30	.27
28 pin ST	.40	.32
40 pin ST	.49	.39
ST SOLDER TAIL		
8 pin WW	.59	.49
14 pin WW	.69	.52
16 pin WW	.69	.58
18 pin WW	.99	.90
20 pin WW	1.09	.98
22 pin WW	1.39	1.28
24 pin WW	1.49	1.35
28 pin WW	1.69	1.49
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7400	.19	74132	.45
7401	.19	74136	.50
7402	.19	74141	.65
7403	.19	74142	2.95
7404	.19	74143	2.95
7405	.25	74145	.60
7406	.29	74147	1.75
7407	.29	74148	1.20
7408	.24	74150	1.35
7409	.19	74151	.55
7410	.19	74152	.65
7411	.25	74153	.55
7412	.30	74154	1.25
7413	.35	74155	.75
7414	.49	74156	.65
7416	.25	74157	.55
7417	.25	74159	1.65
7420	.19	74160	.85
7421	.35	74161	.69
7422	.35	74162	.85
7423	.29	74163	.69
7425	.29	74164	.85
7426	.29	74165	.85
7427	.29	74166	1.00
7428	.45	74167	2.95
7430	.19	74170	1.65
7432	.29	74172	5.95
7433	.45	74173	.75
7437	.29	74174	.89
7438	.29	74175	.89
7440	.19	74176	.89
7442	.49	74177	.75
7443	.65	74178	1.15
7444	.69	74179	1.75
7445	.69	74180	.75
7446	.69	74181	2.25
7447	.69	74182	.75
7448	.69	74184	2.00
7450	.19	74185	2.00
7451	.23	74186	18.50
7453	.23	74190	1.15
7454	.23	74191	1.15
7460	.23	74192	.79
7470	.35	74193	.79
7472	.29	74194	.85
7473	.34	74195	.85
7474	.33	74196	.79
7475	.45	74197	.75
7476	.35	74198	1.35
7480	.59	74199	1.35
7481	1.10	74221	1.35
7482	.95	74246	1.35
7483	.50	74247	1.25
7485	.59	74248	1.85
7486	.35	74249	1.95
7489	2.15	74251	.75
7490	.35	74259	2.25
7491	.40	74265	1.35
7492	.50	74273	1.95
7493	.35	74276	1.25
7494	.65	74279	.75
7495	.55	74283	2.00
7496	.70	74284	3.75
7497	2.75	74285	3.75
74100	1.75	74290	.95
74107	.30	74293	.75
74109	.45	74298	.85
74110	.45	74351	2.25
74111	.55	74365	.65
74116	1.55	74366	.65
74120	1.20	74367	.65
74121	.29	74368	.65
74122	.45	74376	2.20
74123	.49	74390	1.75
74125	.45	74393	1.35
74126	.45	74425	3.15
74128	.55	74426	.85
		74490	2.55

CMOS

4000	.29	4528	1.19
4001	.25	4531	.95
4002	.25	4532	1.95
4006	.89	4538	1.95
4007	.29	4539	1.95
4008	.95	4543	1.19
4009	.39	4555	.95
4010	.45	4556	.95
4011	.25	4581	1.95
4012	.25	4582	1.95
4013	.38	4584	.75
4014	.79	4585	.75
4015	.39	4702	12.95
4016	.39	74C00	.35
4017	.69	74C02	.35
4018	.79	74C04	.35
4019	.39	74C08	.35
4020	.75	74C10	.35
4021	.79	74C14	.59
4022	.79	74C20	.35
4023	.29	74C30	.35
4024	.65	74C32	.39
4025	.29	74C42	1.29
4026	1.65	74C48	1.99
4027	.45	74C73	.65
4028	.69	74C74	.65
4029	.79	74C76	.80
4030	.39	74C83	1.95
4034	1.95	74C85	1.95
4035	.85	74C86	.39
4040	.75	74C89	4.50
4041	.75	74C90	1.19
4042	.69	74C93	1.75
4043	.85	74C95	.99
4044	.79	74C107	.89
4046	.85	74C150	5.75
4047	.95	74C151	2.25
4049	.35	74C154	3.25
4050	.35	74C157	1.75
4051	.79	74C160	1.19
4053	.79	74C161	1.19
4060	.89	74C163	1.19
4066	.39	74C164	1.39
4068	.39	74C164	1.39
4069	.29	74C165	2.00
4070	.35	74C173	.79
4071	.29	74C174	1.19
4072	.29	74C175	1.19
4073	.29	74C192	1.49
4075	.29	74C193	1.49
4076	.79	74C195	1.39
4078	.29	74C200	5.75
4081	.29	74C221	1.75
4082	.29	74C373	2.45
4085	.95	74C374	2.45
4086	.95	74C901	.39
4093	.49	74C902	.85
4098	2.49	74C903	.85
4099	1.95	74C905	10.95
14409	12.95	74C906	.95
14410	12.95	74C907	1.00
14411	11.95	74C908	2.00
14412	12.95	74C909	2.75
14419	7.95	74C910	9.95
4502	.95	74C911	8.95
4503	.65	74C912	8.95
4508	1.95	74C914	1.95
4510	.85	74C915	1.19
4511	.85	74C918	2.75
4512	.85	74C920	17.95
4514	1.25	74C921	15.95
4515	1.79	74C922	4.49
4516	1.55	74C923	4.95
4518	.89	74C925	5.95
4519	.39	74C926	7.95
4520	.79	74C927	7.95
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4526	1.25	74C929	19.95
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74S05	.35	74S175	.95
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74S20	.35	74S195	1.49
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74S85	1.99	74S258	.95
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74S132	1.24	74S288	1.90
74S133	.45	74S289	6.89
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74S138	.85	74S374	2.45
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74S140	.55	74S387	1.95
7			

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LM301	.34	LM339	.99	NE564	2.95	LM1496	.85
LM301H	.79	LM340 (see 7800)		LM565	.99	LM1558H	3.10
LM307	.45	LM348	.99	LM566	1.49	LM1800	2.37
LM308	.69	LM350K	4.95	LM567	.89	LM1812	8.25
LM308H	1.15	LM350T	4.60	NE570	3.95	LM1830	3.50
LM309H	1.95	LM358	.69	NE571	2.95	LM1871	5.49
LM309K	1.25	LM359	1.79	NE592	2.75	LM1872	5.49
LM310	1.75	LM376	3.75	LM703	.89	LM1877	3.25
LM311	.64	LM377	1.95	LM709	.59	LM1889	1.95
LM311H	.89	LM378	2.50	LM710	.75	LM1896	1.75
LM312H	1.75	LM379	4.50	LM711	.79	LM2877	2.05
LM317K	3.95	LM380	.89	LM723	.49	LM2878	2.25
LM317T	1.19	LM380N-8	1.10	LM723H	.55	LM2900	.85
LM318	1.49	LM381	1.60	LM733	.98	LM2901	1.00
LM318H	1.59	LM382	1.60	LM741N-8	.35	LM3900	.59
LM319H	1.25	LM383	1.95	LM741N-14	.35	LM3905	1.25
LM319	1.25	LM384	1.95	LM741H	.40	LM3909	.98
LM320 (see 7900)		LM386	.89	LM747	.69	LM3911	2.25
LM322	1.65	LM387	1.40	LM748	.59	LM3914	3.95
LM323K	4.95	LM389	1.35	LM1014	1.19	LM3915	3.95
LM324	.59	LM390	1.95	LM1303	1.95	LM3916	3.95
LM329	.65	LM392	.69	LM1310	1.49	MC4024	3.95
LM331	3.95	LM394H	4.60	MC1330	1.69	MC4044	4.50
LM334	1.19	LM399H	5.00	MC1349	1.89	RC4136	1.25
LM335	1.40	NE531	2.95	MC1350	1.19	RC4151	3.95
LM336	1.75	NE536	6.00	MC1358	1.69	LM4250	1.75
LM337K	3.95	NE555	.34	LM1414	1.59	LM4500	3.25
LM337T	1.95	NE556	.65	LM1458	.59	LM13080	1.29
LM338K	6.95	NE558	1.50	LM1488	.69	LM13600	1.49
		NE561	19.95	LM1489	.69	LM13700	1.49

H = TO-5 CAN

T = TO-220

K = TO-3

RCA

CA 3023	2.75	CA 3082	1.65
CA 3039	1.29	CA 3083	1.55
CA 3046	1.25	CA 3086	.80
CA 3059	2.90	CA 3089	2.99
CA 3060	2.90	CA 3096	3.49
CA 3065	1.75	CA 3130	1.30
CA 3080	1.10	CA 3140	1.15
CA 3081	1.65	CA 3146	1.85
		CA 3160	1.19

TI

TL494	4.20	75365	1.95
TL496	1.65	75450	.59
TL497	3.25	75451	.39
75107	1.49	75452	.39
75110	1.95	75453	.39
75150	1.95	75454	.39
75154	1.95	75491	.79
75188	1.25	75492	.79
75189	1.25	75493	.89
		75494	.89

BI FET

TL071	.79	TL084	2.19
TL072	1.19	LF347	2.19
TL074	2.19	LF351	.60
TL081	.79	LF353	1.00
TL082	1.19	LF355	1.10
TL083	1.19	LF356	1.10
		LF357	1.40

VOLTAGE REGULATORS

7805T	.89	7905T	.99
7808T	.89	7908T	.99
7812T	.89	7912T	.99
7815T	.89	7915T	.99
7824T	.89	7924T	.99
7805K	1.39	7905K	1.49
7812K	1.39	7912K	1.49
7815K	1.39	7915K	1.49
7824K	1.39	7924K	1.49
78L05	.69	79L05	.79
78L12	.69	79L12	.79
78L15	.69	79L15	.79
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L = TO-92

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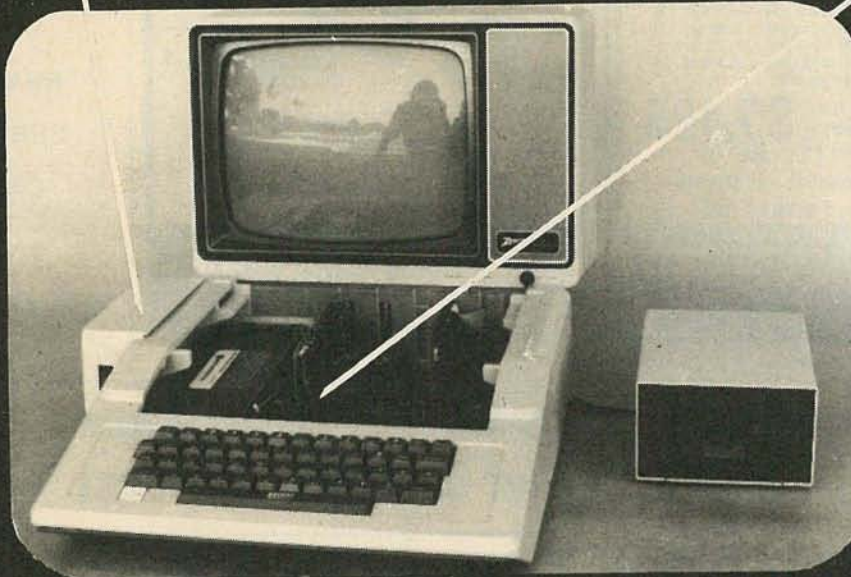
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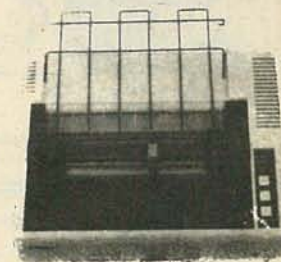
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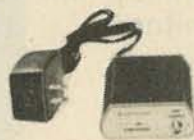
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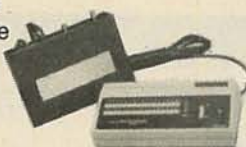
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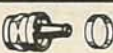
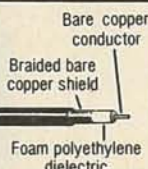
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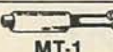
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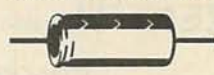


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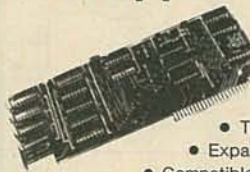
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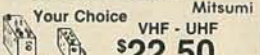


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For the year ending December 31, 1982

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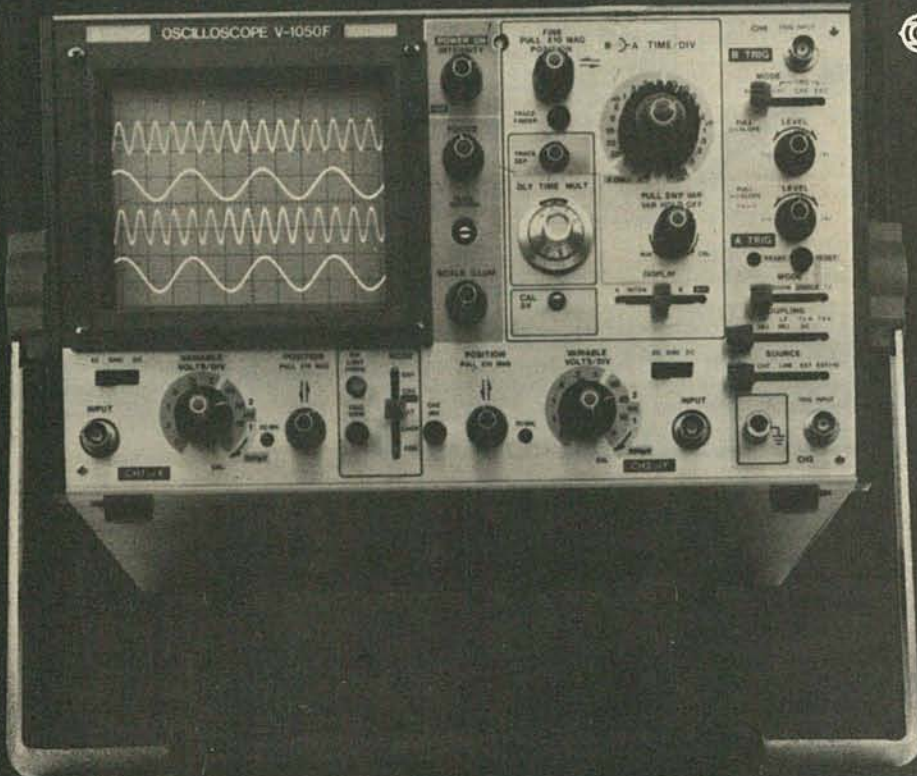
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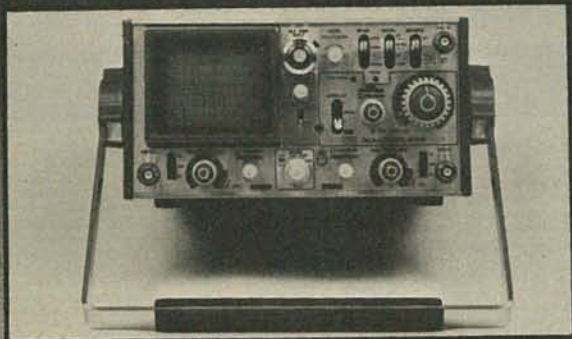
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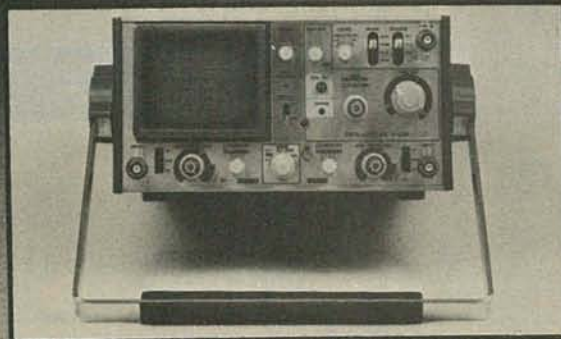
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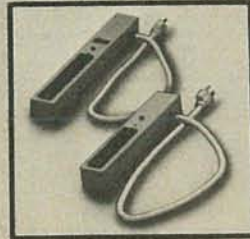
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MODEL TA-2000
\$83.95 per kit

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MONITOR

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This Booster is specially designed for UHF Channels (14-83). After installing between the antenna input cable and the UHF tuner, this unit will provide a minimum of 10dB gain, that is approximately 2 times better than you are seeing now. Ideal for those who live in apartments that can not put up an outdoor antenna. Small in size, only 2" x 1 1/2" x 1". Supply voltage is 15 VDC. Back to Stock.



MODEL 001-0076
\$12.50 ea.

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From 7 WATTS to 100 WATTS
Typical ratings: Operating case temp. 85°C. T.H.D. = 0.5% @ 20Hz-20KHz. Input resistance P=0.1W30KΩ. Power band width 20Hz-20KHz. Freq. response 10-100KHz. Output resistance = 8Ω. With built-in protection circuit. All units come with data sheet.

WHISTLE ACTIVATED SWITCH BOARD

All boards are pre-assembled and tested. Your whistle to its FET condenser microphone from a distance, as far as 30 feet away (sensitivity can be easily adjusted), will turn the switch on and if you whistle again, it will turn off. Ideal for remote control toys, electrical appliance such as lights, coffee pots, TV, Hi-Fi, radio or other projects. Unit works on 9VDC.

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MODEL TR88A
0-15VDC @ 2A
MODEL TR88B
0-30VDC @ 1A

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Tuning voltage +1-128VDC. Input impedance 75Ω. IF band width 7-16MHz. Noise figure 11.5dB Max. Size 2 1/4" x 1 1/4" x 3/4". Supply voltage 15VDC.

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This new designed circuit uses high FREQ FET transistors with 2 stage pre-amp. Transmits FM range (88-120MHz) up to 2 blocks away and with the ultra sensitive condenser microphone that comes with the kit allows you to pick up any sound within 15 ft. away. Kit includes all electronic parts. OSC coils and PC Board. Power supply 9VDC.

FMC-105

\$11.50 Per Kit

FLUORESCENT AUDIO LEVEL MONITOR

This is the kind of VU monitor that is being used by most amplifier manufacturers. IC's are used to simplify circuit layout. Easy to assemble and can be used with all power level amplifiers. Power requirement 12VDC.

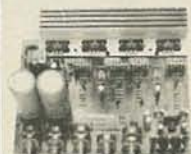


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120W PURE DC POWER STEREO AMP KIT

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★ SPECIAL ★ EXCELLENT PRICE!
MODEL 001-0034
\$29.50 Per Kit
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TA-323 60 WATTS TOTAL 30W + 30W STEREO AMP KIT

This is a solid state all transistor circuitry with on board stereo pre-amp for most microphone or phone input. Power output employs 2 pairs matching Darlingtons Transistors driven by the popular 2N3053 Driver Transistors. Four built on board controls for, volume, balance, treble and bass. Power supply requires 48VCT 2.5A transformer. THD of less than 0.1% between 100Hz-10KHz at full power. (30 Watts + 30 Watts loaded into 8Ω).

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All parts are pre-assembled on a mini PC Board. Supply Voltage 6-9VDC
SPECIAL PRICE \$1.95

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The TA-2900 provides a visual presentation of the changing spectrum thru 100 red LED displays, so you can actually see proof of the equalized sound you've achieved. The TA-2900 kit comes with all the electronic components, IC's, pre-drilled PC board, the instructions and a 19" Rack Mount type metal cabinet with professional silk-screen printed front panel.

- Input Sensitivity Tape Monitor/10mV - 18mV 50K Ω. Speaker Terminal/0.2W - 100W 8Ω
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- Delay Time (1KHz) Fast/18dB/s Slow/6dB/s
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- Power Consumption 36W
- Dimensions 482(W) x 102(H) x 250(D) mm.

\$99.50 per kit



TA-1000 KIT

\$51.95
Power Transformer \$24.00 ea.



100W CLASS A POWER AMP KIT

Dynamic Bias Class "A" circuit design makes this unit unique in its class. Crystal clear, 100 watts power output will satisfy the most picky fans. A perfect combination with the TA-1020 low TIM stereo pre-amp.
Specifications: • Output power 100W RMS into 8Ω, 125W RMS into 4Ω • Frequency response 10Hz-100KHz • THD less than 0.01% • S/N ratio better than 80dB • Input sensitivity 1V max. • Power supply ±40V @ 5A.

REGULATED DUAL VOLTAGE SUPPLY KIT ±10-30 VDC @ 250 ma adjustable, fully regulated. Kit includes all electronic parts, filter capacitors, IC's, heat sinks and PC Board.

\$12.50 per kit

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MARK IV KIT

\$31.50



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CONDENSER TYPE. Touch On — Touch Off. Uses 7473 IC and 12V relay
\$5.50

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12V DC Powered... Lights up 8-15 Watt Fluorescent Light Tubes. Ideal for camper, outdoor, auto or boat. Kit includes high voltage coil, power transistor, heat sink, all other electronic parts and PC Board. Light tube not included.
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Incorporates brand-new DC design that gives a frequency response from 0-100KHz ±0.5dB. Added features like tone defeat and loudness control let you tailor your own frequency supplies to eliminate power fluctuation!
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Only \$44.50
X-former \$4.50 ea.



"FISHER" 30 WATT STEREO AMP

MAIN AMP (15W x 2). Kit Includes 2 pcs. Fisher PA 301 Hybrid IC, all electronic parts with PC Board. Power supply ±16VDC (not included). Power band with KF 1%+3dB). Voltage gain 33dB. 20Hz-20KHz.

Super Buy
Only \$18.50

ULTRASONIC SWITCH KIT

Kit includes the Ultra Sonic Transducers, 2 PC Boards for transmitter and receiver, all electronic parts and instructions. Easy to build and a lot of uses such as remote control for TV, garage door, alarm system or counter. Unit operates by 9-12VDC.
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- ★ Speaker volume control on the back
- ★ Expansion slots

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Easy to assemble! All components are clearly silk screened on the circuit board. Kit includes pre-drilled double sided PC Board, all integrated circuits, sockets, professional high-impact plastic casing, keyboards, connectors and switching power supply. Dealer inquiries invited. No C.O.D. orders

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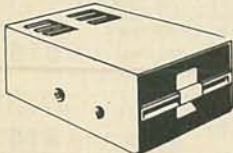
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The system allows you to program both pay TV (decoder required) and standard cable channels for taping on any VCR — while you are watching a different channel on your TV. Simple to install and operate on any make of TV with UHF.

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TOP QUALITY NO SECONDS

TYPE	DESCRIPTION	1 to 10	10 UP
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NE-561	500WV	1.29	1.15
NE-564N	Phase Locked Loop	3.50	3.15
LM-5655N	Phase Locked Loop	1.49	1.29
LM-733N	Video Amp	1.69	1.44
MC-1330	Video Detector	2.49	1.89
MC-1349	Video Amp	2.06	1.85
MC-1350	Video If Amp	1.75	1.55
MC-1352	Video If Amp AGC	2.69	2.09
MC-1358	Audio If Amp	1.95	1.79
MC-1374P	R.F. Modulator	3.19	2.87
MC-1459	Dual Conn. Op Amp	.86	.79
MC-1460N	Balanced Mod/Demodulator	1.79	1.59
LM-1800	PLL Stereo Decoder	4.49	3.95
LM-1889	Video Modulator	2.95	2.49
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LM-7805	5 Volt Positive Volt. Reg.	1.39	1.25
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LM-7815	15 Volt Positive Volt. Reg.	1.39	1.25
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12 db GAIN

CHANNELS 44 - 52

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40-900 MHz

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Stock No.	No. Pins	1-24	25-50
11055	24	5.45	\$3.90
11056	28	4.50	4.05
11057	40	5.95	5.35
11058	64	10.50	9.45

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Stock No. 82501 10 ea. of 10-12-15-18-22-27-33-39-47-56 OHM **\$3.95 each** for any assortment

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Stock No. 82504 10 ea. of 3.3K-3.9K-4.7K-5.6K-6.8K-8.2K-10K-12K-15K-18K OHM

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Stock No. 82507 10 ea. of 1M-1.2M-1.5M-1.8M-2.2M-2.7M-3.3M-3.9M-4.7M-5.6M OHM

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Touch switch capsule. Operating motion is .005" without the use of a levered arm. Extremely fast on and off with low noise. Normally open - rated 115 VAC, 1.6 amp-30 milliohm resistance - .615 radius by 160 thick.

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50076	062	25 4	2.39
50077	062	50 8	4.25
50078	032	33 1.5	1.31
50079	032	85 5	4.47
50080	032	175 8	2.57

TI WIRE WRAP SOCKETS

tin plated phosphor bronze/corr. xct pins with gas tight seal

Stock No.	No. Pins	1-24	25-100
11301	8	\$.45	\$.40
11302	14	.66	.54
11303	16	.72	.64
11304	18	.82	.73
11305	20	1.11	.99
11306	22	1.26	1.12
11307	24	1.41	1.25
11308	28	1.71	1.52
11309	40	2.31	2.05

TI LOW PROFILE SOCKETS

tin plated phosphor bronze/corr. xct pins with gas tight seal

Stock No.	No. Pins	1-24	25-100
11201	8	\$ 1.15	\$ 1.13
11202	14	1.18	1.14
11203	16	2.1	1.16
11204	18	2.4	2.1
11205	20	2.7	2.4
11206	22	3.0	2.6
11207	24	3.3	3.0
11208	28	3.8	3.4
11209	40	5.3	4.5

ELPAC POWER SUPPLIES — DC/DC CONVERTERS

SINTEC Stock No.	ELPAC Part No.	Input Voltage (VDC)	Output Voltage (VDC)	Output Current (MA)	Dimensions (HxWxD) in inches	Price
13827	CB3801	3-0-7.0	12-0-6	0-25	48x51x3.05	7.95
13828	CB3811	3-0-7.0	12-0-6	0-25	48x51x3.05	7.95
13827	CB3802	3-0-7.0	15-0-7	0-20	48x51x3.05	7.95
13828	CB3812	3-0-7.0	15-0-7	0-20	48x51x3.05	7.95
13829	CB3804	3-0-7.0	28-10-7	0-10	48x51x3.05	7.95
13830	CB3814	3-0-7.0	28-10-7	0-10	48x51x3.05	7.95

1.5 W TYPE

Stock No.	CL3801	4-0-7.0	12-0-6	125	65x112x1.77	\$24.95
13832	CL3811 <td>4-0-7.0 <td>12-0-6 <td>125 <td>65x112x1.77 <td>24.95</td> </td></td></td></td>	4-0-7.0 <td>12-0-6 <td>125 <td>65x112x1.77 <td>24.95</td> </td></td></td>	12-0-6 <td>125 <td>65x112x1.77 <td>24.95</td> </td></td>	125 <td>65x112x1.77 <td>24.95</td> </td>	65x112x1.77 <td>24.95</td>	24.95
13833	CL3802	4-0-7.0 <td>15-0-7</td> <td>100 <td>65x112x1.77 <td>24.95</td> </td></td>	15-0-7	100 <td>65x112x1.77 <td>24.95</td> </td>	65x112x1.77 <td>24.95</td>	24.95
13834	CL3812	4-0-7.0 <td>15-0-7</td> <td>100 <td>65x112x1.77 <td>24.95</td> </td></td>	15-0-7	100 <td>65x112x1.77 <td>24.95</td> </td>	65x112x1.77 <td>24.95</td>	24.95
13835	CL3804	4-0-7.0 <td>28-11.4</td> <td>50 <td>65x112x1.77 <td>24.95</td> </td></td>	28-11.4	50 <td>65x112x1.77 <td>24.95</td> </td>	65x112x1.77 <td>24.95</td>	24.95
13836	CL3814	4-0-7.0 <td>28-11.4</td> <td>50 <td>65x112x1.77 <td>24.95</td> </td></td>	28-11.4	50 <td>65x112x1.77 <td>24.95</td> </td>	65x112x1.77 <td>24.95</td>	24.95

13801-1 Data Sheet for 13801 \$ 25

13825-1 DATA SHEET FOR DC/DC CONVERTERS 25

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SINTEC Stock No.	ELPAC Part No.	Output Voltage (VDC)	Output Current (MA)	Dimensions (HxWxD) in inches	OVP	Price
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13803	SOLV15-12	12	4.7/16x42	Fixed included	39.95	
13804	SOLV15-15	15	4.7/16x42	Fixed included	39.95	
13806	SOLV15-24	24	6.75A 4.7/16x42	Fixed included	39.95	
13808	SOLV30-5	5.0A	5.8/84 7/8x3-1/8	OVP-4	59.95	
13809	SOLV30-12	12	5.8/84 7/8x3-1/8	OVP-4	59.95	
13810	SOLV30-15	15	5.8/84 7/8x3-1/8	OVP-4	59.95	
13812	SOLV30-24	24	2.0A 5.8/84 7/8x3-1/8	OVP-4	59.95	

13802-1 Data Sheet for SOLV Series 25

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puts IC's on their true row to row spacing. One side is for 300 centers, flip tool over for devices on 600 centers. Put device in tool and squeeze.

ONE TOOL DOES 8 Thru 40 PINS!
Stock No. 11059 **\$12.95**

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13295	14 pin	16 pin	18 pin	20 pin	22 pin	24 pin	28 pin	30 pin	32 pin	36 pin	40 pin	44 pin

\$1.82 per pack of 10

IC EXTRACTOR

One-piece, spring steel construction. Will extract all LSI, MSI and SSI devices with 8 to 24 pins.
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110/220 volts

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Transformer 110 V.A.C. Input
2/12 V.@ 1 A out P.C. Mount **\$2⁹⁵**

5 to 25 PF Trim Cap 12/2.50



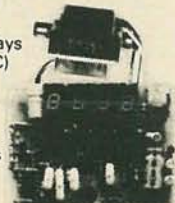
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CIRCLE 74 ON FREE INFORMATION CARD

QUALITY parts at DISCOUNT PRICES

<p>MINI BSR</p> <p>AUTOMATIC RECORD CHANGER</p> <p>* B.S.R. MODEL C136R/C/3 * PLAYS 33/45/78 RECORDS * MINI SIZE: 8 1/4" X 12"</p> <p>INCLUDES DUST COVER AND PLASTIC CASE (NOT PICTURED) WITH FRONT CUT OUT TO FIT STEREO UNIT (NOT INCLUDED).</p> <p>\$17.50 each</p>	<p>COMPUTER GRADE CAPACITOR</p> <p>1700 mfd. 150VDC \$2.00 2 1/2" DIA X 4 3/4" HIGH 3,600 mfd. 40VDC \$1.00 1 3/8" DIA. X 3" HI 6,400 mfd. 60VDC \$2.50 1 3/8" dia X 4 1/4" 12,000 mfd. 40VDC \$3.00 2" DIA X 4 1/4" HIGH 18,000 mfd. 75VDC \$4.00 2 1/2" DIA X 4 1/2" HIGH 22,000 mfd. 15VDC 2" dia X 2 1/2" HIGH \$2.00 22,000 mfd. 40VDC 2" DIA. X 6" HIGH \$3.00 24,000 mfd. 30VDC 1 5/8" DIA X 4" HI \$3.50 31,000 mfd. 15VDC 1 5/8" DIA X 4" HI \$2.50 72,000 mfd. 15VDC 2" DIA. X 4" HIGH \$3.50</p>	<p>TRANSFORMERS</p> <p>120 volt primaries</p> <p>5.6 VOLTS at 750 MA \$3.00 6 VOLTS at 150 MA \$1.25 16.5 V. at 3 AMPS \$6.50 18 VOLTS at 350 MA \$2.00 18 VOLTS at 1 AMP \$4.50 18 V.C.T. at 2 AMP \$5.50 24 VOLTS at 250 ma \$2.50 28 V.C.T. at 3.5 AMP \$6.50 35 V.C.T. at 1 AMP \$3.50 42 V.C.T. at 1.2 AMP \$4.90 65 V.C.T. at 2 AMP \$5.50</p>
<p>JOYSTICK</p> <p>PRECISION DEVICE... CONTAINS 4 50K CENTER TAPPED ALPS POTS. \$4.75 ea</p>	<p>BLACK PLASTIC CASE</p> <p>PAC-TEC SERIES C</p> <p>BLACK PLASTIC ENCLOSURE ADJUSTABLE HEIGHT FROM 1.63" TO 2.93"; WIDTH 6.85"; DEPTH 8". BUILT-IN STAND OFFS FOR P.C. BOARDS. FRONT AND BACK PANELS NOT INCLUDED.. \$5.25 PER CASE.</p>	<p>L.E.D.'s</p> <p>STANDARD JUMBO DIFFUSED</p> <p>RED 10 FOR \$1.50 GREEN 10 FOR \$2.00 YELLOW 10 FOR \$2.00</p> <p>FLASHER LED</p> <p>5 VOLT OPERATION RED JUMBO SIZE 10 FOR \$1.00</p> <p>BI POLAR LED</p> <p>2 FOR \$1.70</p> <p>SUB MINI LED</p> <p>RED .079" X .098" 10 FOR \$1.00 20 mA at 1.75 V \$1.00 200 FOR \$18.00</p>
<p>VARACTOR DIODES</p> <p>BB-103 3 FOR \$1.00 100 FOR \$30.00 MV2205 3 FOR \$1.00 100 FOR \$30.00</p>	<p>MITSUBI MODEL UES-A55 VARACTOR UHF TUNER</p> <p>FREQ. RANGE 470 - 889 MHz ANTENNA INPUT 500 OHMS \$25.00 each 10 for \$220.00</p>	<p>LED HOLDERS</p> <p>TWO PIECE HOLDER FOR JUMBO LED 10 FOR 65¢ 200 FOR \$10.00</p>
<p>SLIDE POTS</p> <p>500K linear taper 2 7/8" LG. 75¢ EACH 1 3/4" TRAVEL 75¢ EACH</p>	<p>35 MFD 330 VOLT</p> <p>1" X 5/8" DIA. 45¢ EACH... 10 FOR \$4.00</p> <p>170 MFD 330 VOLT</p> <p>1 1/2" X 7/8" 2 FOR \$1.50/10 FOR \$7.00</p> <p>600 MFD 360 VOLT</p> <p>3 3/4" HIGH X 1" DIA. \$1.00 EA. 10 FOR \$9.00</p> <p>750 MFD 330 VOLT</p> <p>2" HIGH X 1 3/4" \$1.25 EACH 10 FOR \$11.00</p>	<p>DC WALL TRANSFORMER</p> <p>ALL ARE 115 VAC PLUG IN</p> <p>4 VDC at 70 MA \$2.50 5.8 VDC at 125 MA \$2.50 9 VDC at 100 MA \$2.00 9 VDC at 225 MA \$3.00 22 VDC at 60 MA \$2.50</p>

FREE! SEND FOR OUR NEW 1983 40 PAGE CATALOG FREE!

<p>KEYBOARD w/ case</p> <p>16 KEY KEYBOARD MATRIX ENCODED</p> <p>TERMINATES TO FLEXIBLE CABLE WITH CONTACTS ON 100 CENTERS. EDGE CONNECTOR INCLUDED... \$4.50 PER KEYBOARD, CASE, AND CONNECTOR.....</p>	<p>4PDT RELAY</p> <p>14 pin style 3 amp contacts 24 volt d.c. at 150 mA a.c. coil Used but fully tested \$1.70 EACH specify coil voltage LARGE QUANTITIES AVAILABLE SOCKETS FOR RELAY 50¢ each</p>	<p>4 WATT STEREO AMP</p> <p>ASSEMBLY INCLUDES VOLUME, BALANCE AND TONE CONTROLS, CERAMIC PHONE INPUT, LINE INPUT AND 8 OHM OUTPUT FOR SPEAKER OR HEADPHONES. OPERATES ON 12 VAC @ 500MA. OUR ABOVE 12 VOLT TRANSFORMER IS IDEAL FOR THIS APPLICATION \$4.50 EACH</p>
<p>EDGE CONNECTOR</p> <p>ALL ARE .156" SPACING</p> <p>15/30 GOLD SOLDER EYELET \$2.00 EACH</p> <p>18/36 GOLD SOLDER EYELET \$2.00 EACH</p> <p>22/44 GOLD SOLDERTAIL (P.C. STYLE) \$2.50 EA 10 FOR \$22.50</p> <p>22/44 TIN SOLDERTAIL (P.C. STYLE) \$1.35 EA 10 FOR \$12.50</p> <p>22/44 GOLD SOLDER EYELET \$2.50 EACH</p>	<p>6 VDC RELAY</p> <p>MINIATURE D.P.D.T. 3 AMP CONTACTS FUJITSU # FBR321D006 \$1.75 EA 10 / 16.00</p>	<p>13 VDC RELAY</p> <p>CONTACT: S.P.N.C. 10 AMP @ 120 VAC ENERGIZE COIL TO OPEN CONTACT... COIL: 13 VDC 650 OHMS SPECIAL PRICE \$1.00 EACH</p>
<p>2 CHANNEL LIGHT ORGAN</p> <p>EASILY HOOKS INTO STEREO SPEAKERS AND ALLOWS 110 VAC LIGHTS TO DANCE WITH MUSIC. TWO SEPARATE 110 VAC OUTPUTS FOR HIGH AND LOW FREQUENCY AUDIO SIGNALS. USE TWO ORGANS FOR STEREO... \$6.50 PER UNIT COLOR LIGHT STRING AVAILABLE \$1.75 EA</p>	<p>FOOT CONTROL WITH SIDE SWITCH</p> <p>MEASURES 4 3/4" X 9 3/4" WITH 100 K LINEAR POT AND MOMENTARY S.P.D.T. SWITCH. \$6.50 EACH</p>	<p>MINI-PUSH BUTTON</p> <p>S.P.S.T. MOMENTARY NORMALLY OPEN 1/4" BUSHING 25¢ EACH 10 FOR \$2.25 100 FOR \$20.00 SPECIFY COLOR: RED, BLACK, WHITE, GREEN, YELLOW.</p>
<p>6 volt 9amp/hr RECHARGEABLE</p> <p>ELPOWER # EP690 SOLID GEL CELL 5 1/2" X 4 1/4" X 2 3/4"..... \$15.00 EACH</p>	<p>CANNON XLRA-3-13 CONNECTOR</p> <p>3 PRONG CHASSIS MOUNT CONNECTOR \$2.00 EACH 10 for \$19.00</p>	<p>4 CHANNEL 8 TRACK HOME UNIT</p> <p>BRAND NEW UNITS... ASSEMBLY INCLUDES; TAPE HEAD, MOTOR BELT, 110VAC MOTOR, PRE-AMP, LIGHTS, SWITCHES, SOLENOID AND OTHER USEFUL PARTS... AN EXCEPTIONAL BUY! \$7.25 PER ASSEMBLY</p>
<p>BLACK LIGHT (ULTRAVIOLET)</p> <p>G.E. # F6T5BL \$2.50 each</p>	<p>CRYSTALS</p> <p>CASE STYLE HC33/U 2 MHZ: 3579, 545 KC \$3.50 EACH \$1.00 EACH</p>	<p>METAL OXIDE VARISTOR</p> <p>G.E. # V822A12 82 VOLTS, NOMINAL D.C. VOLTAGE. 5/8" DIAMETER. 2 FOR \$1.50</p>

EQUIPMENT SLIDES CHASSIS-TRAK MODEL C3005
3 SECTION. LENGTH 22" CLOSED.
HOLDS TO 85 LBS, EXTENDS 23"
\$5.00 PER PAIR SOME HARDWARE INCLUDED

ALL ELECTRONICS CORP.

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ALASKA, HAWAII, CALIF
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CIRCLE 58 ON FREE INFORMATION CARD

DECEMBER 1982

RADIO-ELECTRONICS CORPORATION

In Mn., Ak., HI. Call 218-681-6674

QUALITY - Name brand products from nationally recognized manufacturers. **SERVICE** - Computerized order processing and inventory control. **SAVINGS** - Volume Discounts • OEM Quantity Pricing • Toll Free 800 Number

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AVOID • CHEM-WIK • BISHOP GRAPHICS •
OK • MACHINERY • UNICOR • GC CHEMICALS •
NATIONAL SEMICONDUCTOR • PANASONIC

11T SILICON RECTIFIERS
3 AMP SILICON RECTIFIER

Part No. Price
11T100 1.00 1.00
11T150 1.00 1.00
11T200 1.00 1.00
11T250 1.00 1.00
11T300 1.00 1.00
11T350 1.00 1.00
11T400 1.00 1.00
11T450 1.00 1.00
11T500 1.00 1.00
11T550 1.00 1.00
11T600 1.00 1.00
11T650 1.00 1.00
11T700 1.00 1.00
11T750 1.00 1.00
11T800 1.00 1.00
11T850 1.00 1.00
11T900 1.00 1.00
11T950 1.00 1.00

National Semiconductor

MA1026 Time Temp. Clock Module
Clock: 2 1/2 Inch 4-Digit LED Displays
• 16 and 24 Hour Drives
• 60 Hz line frequency operation
• 12 or 24 Hour Drives

Part No. Price
MA1026 1.00 1.00
MA1027 1.00 1.00
MA1028 1.00 1.00
MA1029 1.00 1.00
MA1030 1.00 1.00
MA1031 1.00 1.00
MA1032 1.00 1.00
MA1033 1.00 1.00
MA1034 1.00 1.00
MA1035 1.00 1.00
MA1036 1.00 1.00
MA1037 1.00 1.00
MA1038 1.00 1.00
MA1039 1.00 1.00
MA1040 1.00 1.00
MA1041 1.00 1.00
MA1042 1.00 1.00
MA1043 1.00 1.00
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MA1045 1.00 1.00
MA1046 1.00 1.00
MA1047 1.00 1.00
MA1048 1.00 1.00
MA1049 1.00 1.00
MA1050 1.00 1.00

TERMINATION JUMPERS B-SOCKET CONNECTOR TERMINATION P-FPC CONNECTOR TERMINATION C-CARD EDGE CONNECTOR TERMINATION Part No. Price RJ100 1.00 1.00 RJ101 1.00 1.00 RJ102 1.00 1.00 RJ103 1.00 1.00 RJ104 1.00 1.00 RJ105 1.00 1.00 RJ106 1.00 1.00 RJ107 1.00 1.00 RJ108 1.00 1.00 RJ109 1.00 1.00 RJ110 1.00 1.00 RJ111 1.00 1.00 RJ112 1.00 1.00 RJ113 1.00 1.00 RJ114 1.00 1.00 RJ115 1.00 1.00 RJ116 1.00 1.00 RJ117 1.00 1.00 RJ118 1.00 1.00 RJ119 1.00 1.00 RJ120 1.00 1.00 RJ121 1.00 1.00 RJ122 1.00 1.00 RJ123 1.00 1.00 RJ124 1.00 1.00 RJ125 1.00 1.00 RJ126 1.00 1.00 RJ127 1.00 1.00 RJ128 1.00 1.00 RJ129 1.00 1.00 RJ130 1.00 1.00 RJ131 1.00 1.00 RJ132 1.00 1.00 RJ133 1.00 1.00 RJ134 1.00 1.00 RJ135 1.00 1.00 RJ136 1.00 1.00 RJ137 1.00 1.00 RJ138 1.00 1.00 RJ139 1.00 1.00 RJ140 1.00 1.00 RJ141 1.00 1.00 RJ142 1.00 1.00 RJ143 1.00 1.00 RJ144 1.00 1.00 RJ145 1.00 1.00 RJ146 1.00 1.00 RJ147 1.00 1.00 RJ148 1.00 1.00 RJ149 1.00 1.00 RJ150 1.00 1.00

PCB TRANSISTORS

Part No. Price
MJE1300 1.00 1.00
MJE1301 1.00 1.00
MJE1302 1.00 1.00
MJE1303 1.00 1.00
MJE1304 1.00 1.00
MJE1305 1.00 1.00
MJE1306 1.00 1.00
MJE1307 1.00 1.00
MJE1308 1.00 1.00
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MJE1347 1.00 1.00
MJE1348 1.00 1.00
MJE1349 1.00 1.00
MJE1350 1.00 1.00

alpha wire cable tubing

Minimum Wire Diameter: PVC Insulated
• 1/8" to 1/2" Wire Size
• 1/8" to 1/2" Tube Size

Part No. Price
AW100 1.00 1.00
AW101 1.00 1.00
AW102 1.00 1.00
AW103 1.00 1.00
AW104 1.00 1.00
AW105 1.00 1.00
AW106 1.00 1.00
AW107 1.00 1.00
AW108 1.00 1.00
AW109 1.00 1.00
AW110 1.00 1.00
AW111 1.00 1.00
AW112 1.00 1.00
AW113 1.00 1.00
AW114 1.00 1.00
AW115 1.00 1.00
AW116 1.00 1.00
AW117 1.00 1.00
AW118 1.00 1.00
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AW145 1.00 1.00
AW146 1.00 1.00
AW147 1.00 1.00
AW148 1.00 1.00
AW149 1.00 1.00
AW150 1.00 1.00

11T SILICON ZENER DIODES

Part No. Price
11T100 1.00 1.00
11T150 1.00 1.00
11T200 1.00 1.00
11T250 1.00 1.00
11T300 1.00 1.00
11T350 1.00 1.00
11T400 1.00 1.00
11T450 1.00 1.00
11T500 1.00 1.00
11T550 1.00 1.00
11T600 1.00 1.00
11T650 1.00 1.00
11T700 1.00 1.00
11T750 1.00 1.00
11T800 1.00 1.00
11T850 1.00 1.00
11T900 1.00 1.00
11T950 1.00 1.00

Programmable Clock Modules

6 7/8" or 8 1/4" Red 4-Digit LED Displays
• 12 or 24 Hour Drives
• 60 Hz line frequency operation
• Display hours and minutes, and seconds, alarm, and sleep time
• Battery back up in case of power failure
• Typical applications include: clock radio, time alarm, auto clock, instrument panel clock, alarm clock

Part No. Price
MA1026 1.00 1.00
MA1027 1.00 1.00
MA1028 1.00 1.00
MA1029 1.00 1.00
MA1030 1.00 1.00
MA1031 1.00 1.00
MA1032 1.00 1.00
MA1033 1.00 1.00
MA1034 1.00 1.00
MA1035 1.00 1.00
MA1036 1.00 1.00
MA1037 1.00 1.00
MA1038 1.00 1.00
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MA1046 1.00 1.00
MA1047 1.00 1.00
MA1048 1.00 1.00
MA1049 1.00 1.00
MA1050 1.00 1.00

PCB TRANSISTORS

Part No. Price
MJE1300 1.00 1.00
MJE1301 1.00 1.00
MJE1302 1.00 1.00
MJE1303 1.00 1.00
MJE1304 1.00 1.00
MJE1305 1.00 1.00
MJE1306 1.00 1.00
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MJE1346 1.00 1.00
MJE1347 1.00 1.00
MJE1348 1.00 1.00
MJE1349 1.00 1.00
MJE1350 1.00 1.00

DIP Jumper Cable Assemblies

Standard Assembly Styles

Part No. Price
DJ100 1.00 1.00
DJ101 1.00 1.00
DJ102 1.00 1.00
DJ103 1.00 1.00
DJ104 1.00 1.00
DJ105 1.00 1.00
DJ106 1.00 1.00
DJ107 1.00 1.00
DJ108 1.00 1.00
DJ109 1.00 1.00
DJ110 1.00 1.00
DJ111 1.00 1.00
DJ112 1.00 1.00
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DJ142 1.00 1.00
DJ143 1.00 1.00
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DJ145 1.00 1.00
DJ146 1.00 1.00
DJ147 1.00 1.00
DJ148 1.00 1.00
DJ149 1.00 1.00
DJ150 1.00 1.00

SHRINKING TUBING

Part No. Price
ST100 1.00 1.00
ST101 1.00 1.00
ST102 1.00 1.00
ST103 1.00 1.00
ST104 1.00 1.00
ST105 1.00 1.00
ST106 1.00 1.00
ST107 1.00 1.00
ST108 1.00 1.00
ST109 1.00 1.00
ST110 1.00 1.00
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ST146 1.00 1.00
ST147 1.00 1.00
ST148 1.00 1.00
ST149 1.00 1.00
ST150 1.00 1.00

JAG "NORMAL-BLO" FUSES

Part No. Price
F100 1.00 1.00
F101 1.00 1.00
F102 1.00 1.00
F103 1.00 1.00
F104 1.00 1.00
F105 1.00 1.00
F106 1.00 1.00
F107 1.00 1.00
F108 1.00 1.00
F109 1.00 1.00
F110 1.00 1.00
F111 1.00 1.00
F112 1.00 1.00
F113 1.00 1.00
F114 1.00 1.00
F115 1.00 1.00
F116 1.00 1.00
F117 1.00 1.00
F118 1.00 1.00
F119 1.00 1.00
F120 1.00 1.00
F121 1.00 1.00
F122 1.00 1.00
F123 1.00 1.00
F124 1.00 1.00
F125 1.00 1.00
F126 1.00 1.00
F127 1.00 1.00
F128 1.00 1.00
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F136 1.00 1.00
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F138 1.00 1.00
F139 1.00 1.00
F140 1.00 1.00
F141 1.00 1.00
F142 1.00 1.00
F143 1.00 1.00
F144 1.00 1.00
F145 1.00 1.00
F146 1.00 1.00
F147 1.00 1.00
F148 1.00 1.00
F149 1.00 1.00
F150 1.00 1.00

5 1/2" 4-Digit LED Clock

Part No. Price
MA1026 1.00 1.00
MA1027 1.00 1.00
MA1028 1.00 1.00
MA1029 1.00 1.00
MA1030 1.00 1.00
MA1031 1.00 1.00
MA1032 1.00 1.00
MA1033 1.00 1.00
MA1034 1.00 1.00
MA1035 1.00 1.00
MA1036 1.00 1.00
MA1037 1.00 1.00
MA1038 1.00 1.00
MA1039 1.00 1.00
MA1040 1.00 1.00
MA1041 1.00 1.00
MA1042 1.00 1.00
MA1043 1.00 1.00
MA1044 1.00 1.00
MA1045 1.00 1.00
MA1046 1.00 1.00
MA1047 1.00 1.00
MA1048 1.00 1.00
MA1049 1.00 1.00
MA1050 1.00 1.00

PCB TRANSISTORS

Part No. Price
MJE1300 1.00 1.00
MJE1301 1.00 1.00
MJE1302 1.00 1.00
MJE1303 1.00 1.00
MJE1304 1.00 1.00
MJE1305 1.00 1.00
MJE1306 1.00 1.00
MJE1307 1.00 1.00
MJE1308 1.00 1.00
MJE1309 1.00 1.00
MJE1310 1.00 1.00
MJE1311 1.00 1.00
MJE1312 1.00 1.00
MJE1313 1.00 1.00
MJE1314 1.00 1.00
MJE1315 1.00 1.00
MJE1316 1.00 1.00
MJE1317 1.00 1.00
MJE1318 1.00 1.00
MJE1319 1.00 1.00
MJE1320 1.00 1.00
MJE1321 1.00 1.00
MJE1322 1.00 1.00
MJE1323 1.00 1.00
MJE1324 1.00 1.00
MJE1325 1.00 1.00
MJE1326 1.00 1.00
MJE1327 1.00 1.00
MJE1328 1.00 1.00
MJE1329 1.00 1.00
MJE1330 1.00 1.00
MJE1331 1.00 1.00
MJE1332 1.00 1.00
MJE1333 1.00 1.00
MJE1334 1.00 1.00
MJE1335 1.00 1.00
MJE1336 1.00 1.00
MJE1337 1.00 1.00
MJE1338 1.00 1.00
MJE1339 1.00 1.00
MJE1340 1.00 1.00
MJE1341 1.00 1.00
MJE1342 1.00 1.00
MJE1343 1.00 1.00
MJE1344 1.00 1.00
MJE1345 1.00 1.00
MJE1346 1.00 1.00
MJE1347 1.00 1.00
MJE1348 1.00 1.00
MJE1349 1.00 1.00
MJE1350 1.00 1.00

DIP Jumper Cable Assemblies

Part No. Price
DJ100 1.00 1.00
DJ101 1.00 1.00
DJ102 1.00 1.00
DJ103 1.00 1.00
DJ104 1.00 1.00
DJ105 1.00 1.00
DJ106 1.00 1.00
DJ107 1.00 1.00
DJ108 1.00 1.00
DJ109 1.00 1.00
DJ110 1.00 1.00
DJ111 1.00 1.00
DJ112 1.00 1.00
DJ113 1.00 1.00
DJ114 1.00 1.00
DJ115 1.00 1.00
DJ116 1.00 1.00
DJ117 1.00 1.00
DJ118 1.00 1.00
DJ119 1.00 1.00
DJ120 1.00 1.00
DJ121 1.00 1.00
DJ122 1.00 1.00
DJ123 1.00 1.00
DJ124 1.00 1.00
DJ125 1.00 1.00
DJ126 1.00 1.00
DJ127 1.00 1.00
DJ128 1.00 1.00
DJ129 1.00 1.00
DJ130 1.00 1.00
DJ131 1.00 1.00
DJ132 1.00 1.00
DJ133 1.00 1.00
DJ134 1.00 1.00
DJ135 1.00 1.00
DJ136 1.00 1.00
DJ137 1.00 1.00
DJ138 1.00 1.00
DJ139 1.00 1.00
DJ140 1.00 1.00
DJ141 1.00 1.00
DJ142 1.00 1.00
DJ143 1.00 1.00
DJ144 1.00 1.00
DJ145 1.00 1.00
DJ146 1.00 1.00
DJ147 1.00 1.00
DJ148 1.00 1.00
DJ149 1.00 1.00
DJ150 1.00 1.00

SHRINKING TUBING

Part No. Price
ST100 1.00 1.00
ST101 1.00 1.00
ST102 1.00 1.00
ST103 1.00 1.00
ST104 1.00 1.00
ST105 1.00 1.00
ST106 1.00 1.00
ST107 1.00 1.00
ST108 1.00 1.00
ST109 1.00 1.00
ST110 1.00 1.00
ST111 1.00 1.00
ST112 1.00 1.00
ST113 1.00 1.00
ST114 1.00 1.00
ST115 1.00 1.00
ST116 1.00 1.00
ST117 1.00 1.00
ST118 1.00 1.00
ST119 1.00 1.00
ST120 1.00 1.00
ST121 1.00 1.00
ST122 1.00 1.00
ST123 1.00 1.00
ST124 1.00 1.00
ST125 1.00 1.00
ST126 1.00 1.00
ST127 1.00 1.00
ST128 1.00 1.00
ST129 1.00 1.00
ST130 1.00 1.00
ST131 1.00 1.00
ST132 1.00 1.00
ST133 1.00 1.00
ST134 1.00 1.00
ST135 1.00 1.00
ST136 1.00 1.00
ST137 1.00 1.00
ST138 1.00 1.00
ST139 1.00 1.00
ST140 1.00 1.00
ST141 1.00 1.00
ST142 1.00 1.00
ST143 1.00 1.00
ST144 1.00 1.00
ST145 1.00 1.00
ST146 1.00 1.00
ST147 1.00 1.00
ST148 1.00 1.00
ST149 1.00 1.00
ST150 1.00 1.00

JAG "BLO" FUSES

Part No. Price
F100 1.00 1.00
F101 1.00 1.00
F102 1.00 1.00
F103 1.00 1.00
F104 1.00 1.00
F105 1.00 1.00
F106 1.00 1.00
F107 1.00 1.00
F108 1.00 1.00
F109 1.00 1.00
F110 1.00 1.00
F111 1.00 1.00
F112 1.00 1.00
F113 1.00 1.00
F114 1.00 1.00
F115 1.00 1.00
F116 1.00 1.00
F117 1.00 1.00
F118 1.00 1.00
F119 1.00 1.00
F120 1.00 1.00
F121 1.00 1.00
F122 1.00 1.00
F123 1.00 1.00
F124 1.00 1.00
F125 1.00 1.00
F126 1.00 1.00
F127 1.00 1.00
F128 1.00 1.00
F129 1.00 1.00
F130 1.00 1.00
F131 1.00 1.00
F132 1.00 1.00
F133 1.00 1.00
F134

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9 DIGITS 600 MHz \$129⁹⁵ WIRED

PRICES:
 CT-90 wired, 1 year warranty \$129.95
 CT-90 Kit, 90 day parts warranty 109.95
 AC-1 AC adapter 3.95
 BP-1 Nicad pack + AC Adapter/Charger 12.95
 OV-1 Micro-power Oven time base 49.95
 External time base input 14.95

The CT-90 is the most versatile, feature packed counter available for less than \$300.00! Advanced design features include; three selectable gate times, nine digits, gate indicator and a unique display hold function which holds the displayed count after the input signal is removed. Also, a 10MHz TCXO time base is used which enables easy zero beat calibration checks against WWV. Optionally, an internal nicad battery pack, external time base input and Micro-power high stability crystal oven time base are available. The CT-90, performance you can count on!

SPECIFICATIONS:
 Range: 20 Hz to 600 MHz
 Sensitivity: Less than 10 MV to 150 MHz
 Less than 50 MV to 500 MHz
 Resolution: 0.1 Hz (10 MHz range)
 1.0 Hz (60 MHz range)
 10.0 Hz (600 MHz range)
 Display: 9 digits 0.4" LED
 Time base: Standard-10,000 mHz, 1.0 ppm 20-40°C.
 Optional Micro-power oven-0.1 ppm 20-40°C
 Power: 8-15 VAC @ 250 ma

7 DIGITS 525 MHz \$99⁹⁵ WIRED

SPECIFICATIONS:

Range: 20 Hz to 525 MHz
 Sensitivity: Less than 50 MV to 150 MHz
 Less than 150 MV to 500 MHz
 Resolution: 1.0 Hz (5 MHz range)
 10.0 Hz (50 MHz range)
 100.0 Hz (500 MHz range)
 Display: 7 digits 0.4" LED
 Time base: 1.0 ppm TCXO 20-40°C
 Power: 12 VAC @ 250 ma

The CT-70 breaks the price barrier on lab quality frequency counters. Deluxe features such as, three frequency ranges - each with pre-amplification, dual selectable gate times, and gate activity indication make measurements a snap. The wide frequency range enables you to accurately measure signals from audio thru UHF with 1.0 ppm accuracy - that's .0001%! The CT-70 is the answer to all your measurement needs, in the field, lab or ham shack.

PRICES:

CT-70 wired, 1 year warranty \$99.95
 CT-70 Kit, 90 day parts warranty 84.95
 AC-1 AC adapter 3.95
 BP-1 Nicad pack + AC adapter/charger 12.95



7 DIGITS 500 MHz \$79⁹⁵ WIRED

PRICES:

MINI-100 wired, 1 year warranty \$79.95
 AC-Z Ac adapter for MINI-100 3.95
 BP-Z Nicad pack and AC adapter/charger 12.95

Here's a handy, general purpose counter that provides most counter functions at an unbelievable price. The MINI-100 doesn't have the full frequency range or input impedance qualities found in higher price units, but for basic RF signal measurements, it can't be beat! Accurate measurements can be made from 1 MHz all the way up to 500 MHz with excellent sensitivity throughout the range, and the two gate times let you select the resolution desired. Add the nicad pack option and the MINI-100 makes an ideal addition to your tool box for "in-the-field" frequency checks and repairs.

SPECIFICATIONS:

Range: 1 MHz to 500 MHz
 Sensitivity: Less than 25 MV
 Resolution: 100 Hz (slow gate)
 1.0 KHz (fast gate)
 Display: 7 digits, 0.4" LED
 Time base: 2.0 ppm 20-40°C
 Power: 5 VDC @ 200 ma

8 DIGITS 600 MHz \$159⁹⁵ WIRED



SPECIFICATIONS:

Range: 20 Hz to 600 MHz
 Sensitivity: Less than 25 mv to 150 MHz
 Less than 150 mv to 600 MHz
 Resolution: 1.0 Hz (60 MHz range)
 10.0 Hz (600 MHz range)
 Display: 8 digits 0.4" LED
 Time base: 2.0 ppm 20-40°C
 Power: 110 VAC or 12 VDC

The CT-50 is a versatile lab bench counter that will measure up to 600 MHz with 8 digit precision. And, one of its best features is the Receive Frequency Adapter, which turns the CT-50 into a digital readout for any receiver. The adapter is easily programmed for any receiver and a simple connection to the receiver's VFO is all that is required for use. Adding the receiver adapter in no way limits the operation of the CT-50, the adapter can be conveniently switched on or off. The CT-50, a counter that can work double-duty!

PRICES:

CT-50 wired, 1 year warranty \$159.95
 CT-50 Kit, 90 day parts warranty 119.95
 RA-1, receiver adapter kit 14.95
 RA-1 wired and pre-programmed (send copy of receiver schematic) 29.95



DIGITAL MULTIMETER \$99⁹⁵ WIRED



PRICES:

DM-700 wired, 1 year warranty \$99.95
 DM-700 Kit, 90 day parts warranty 79.95
 AC-1, AC adaptor 3.95
 BP-3, Nicad pack + AC adapter/charger 19.95
 MP-1, Probe kit 2.95

The DM-700 offers professional quality performance at a hobbyist price. Features include; 26 different ranges and 5 functions, all arranged in a convenient, easy to use format. Measurements are displayed on a large 3 1/2 digit, 1/2 inch LED readout with automatic decimal placement, automatic polarity, overrange indication and overload protection up to 1250 volts on all ranges, making it virtually goof-proof! The DM-700 looks great, a handsome, jet black, rugged ABS case with convenient retractable tilt bail makes it an ideal addition to any shop.

SPECIFICATIONS:

DC/AC volts: 100uV to 1 KV, 5 ranges
 DC/AC current: 0.1uA to 2.0 Amps, 5 ranges
 Resistance: 0.1 ohms to 20 Megohms, 6 ranges
 Input impedance: 10 Megohms, DC/AC volts
 Accuracy: 0.1% basic DC volts
 Power: 4 'C' cells

AUDIO SCALER

For high resolution audio measurements, multiplies UP in frequency.

- Great for PL tones
- Multiplies by 10 or 100
- 0.01 Hz resolution!

\$29.95 Kit \$39.95 Wired

ACCESSORIES

- Telescopic whip antenna - BNC plug..... \$ 7.95
- High impedance probe, light loading..... 15.95
- Low pass probe, for audio measurements..... 15.95
- Direct probe, general purpose usage..... 12.95
- Tilt bail, for CT 70, 90, MINI-100..... 3.95
- Color burst calibration unit, calibrates counter against color TV signal..... 14.95

COUNTER PREAMP

For measuring extremely weak signals from 10 to 1,000 MHz. Small size, powered by plug transformer-included.

- Flat 25 db gain
- BNC Connectors
- Great for sniffing RF with pick-up loop

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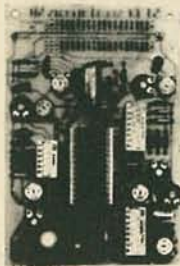
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Sound Effects Kit \$18.50



The SE-01 is a complete kit that contains all the parts to build a programmable sound effects generator. Designed around the new Texas Instruments SN76747 Sound Chip, the board provides banks of MINI DIP switches and pots to program the various combinations of the SLP Oscillator, VCO, Noise, One Shot, and Envelope Controls. A Quad Op Amp IC is used to implement an Adjustable Pulse Generator, Level Comparator and Multiplier. Oscillator for even more versatility. The 3 1/2 x 5 PC Board features a prototype area to allow for user added circuitry. Easily programmed to duplicate Explosions, Phaser Guns, Steam Trains, or almost an infinite number of other sounds. The unit has a multiple of applications. The low price includes all parts, assembly manual, programming charts, and detailed 76477 chip specifications. It runs on a 9V battery (not included). On board 100mV amp will drive a small speaker directly, or the unit can be connected to your stereo with incredible results. (Speaker not included). 76477 is included. Available separately for \$5.15 each.

7 Watt Audio Amp Kit \$6.95

SMALL, SINGLE HYBRID IC AND COMPONENTS FIT ON A 2" x 3" PC BOARD (INCLUDED). RUNS ON 12VDC. GREAT FOR ANY PROJECT THAT NEEDS AN INEXPENSIVE AMP. LESS THAN 3% THD @ 5 WATTS. COMPATIBLE WITH SE-01 SOUND KIT.

The Super Music Maker

REVISION 2 \$24.95 (Basic Kit)

SEE SPECIAL OFFER BELOW



- Comes with 25 pre-programmed tunes.
- On board 7 watt amplifier drives 8 ohm speaker directly.
- Works on 12VAC or 12 VDC.
- Uses either 2708 or 2716 EPROM for expanded tune playing capability. Listing available pre-programmed ROM's.

A true electronic music maker based on a microprocessor chip. The Super Music Maker is the only kit that allows easy addition of pre-programmed tunes by plugging in one memory chip (ROM). Over 20 different ROM's with over 500 tunes are available. Use the kit for a Car Horn, Doorbell, Door Announcer, etc. If you have an EPROM programmer our manual tells you how to program your own tunes. Kit includes quality plated and drilled PC Board and all components.

OPTIONAL ACCESSORIES

DIP SWITCHES One 8 pos. one 5 pos. for tune address 2.00/set
WALL-PLUG TRANSFORMER For operation on 117VAC house current 3.00
INJECTION MOLDED PLASTIC CASE w/custom front & rear panels, hardware and 2 five pos rotary switches (replaces DIP switches) 8.00
HORN SPEAKER 8 watt, 8 Ohm with mounting bracket 6.25

SPECIAL OFFER

Buy a Super Music Maker kit for \$24.95 and get FREE, a 2708 ROM pre-programmed with 35 popular tunes. This offer gives you over 60 songs to choose from!

Doomsday Alarm Kit \$9.95

If you have trouble sleeping and you would like the rest of the neighborhood to share your misery then this little kit will be for you! There is no way to accurately describe the unearthly howls, screams and tones that come out of this kit. Four separate tone oscillators are mixed, cancelled and stepped at a varying rate. 10 Watts of crazy sounds. A great fun kit or a practical burglar alarm. Complete with PC board and all necessary components less speaker. For 6-12 VDC. ORDER DA-02.

- UNIT IS COMPLETELY ASSEMBLED!
- Fused primary and DC sections
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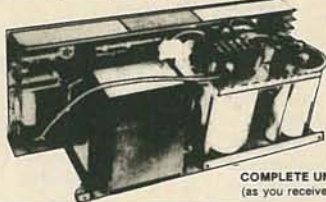
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